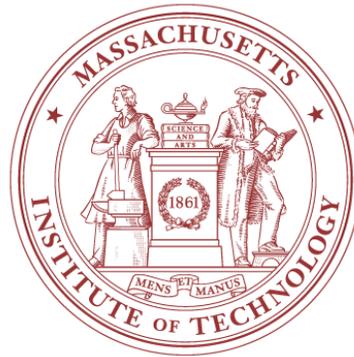


Intro to Practical Digital Communications

Lecture 2

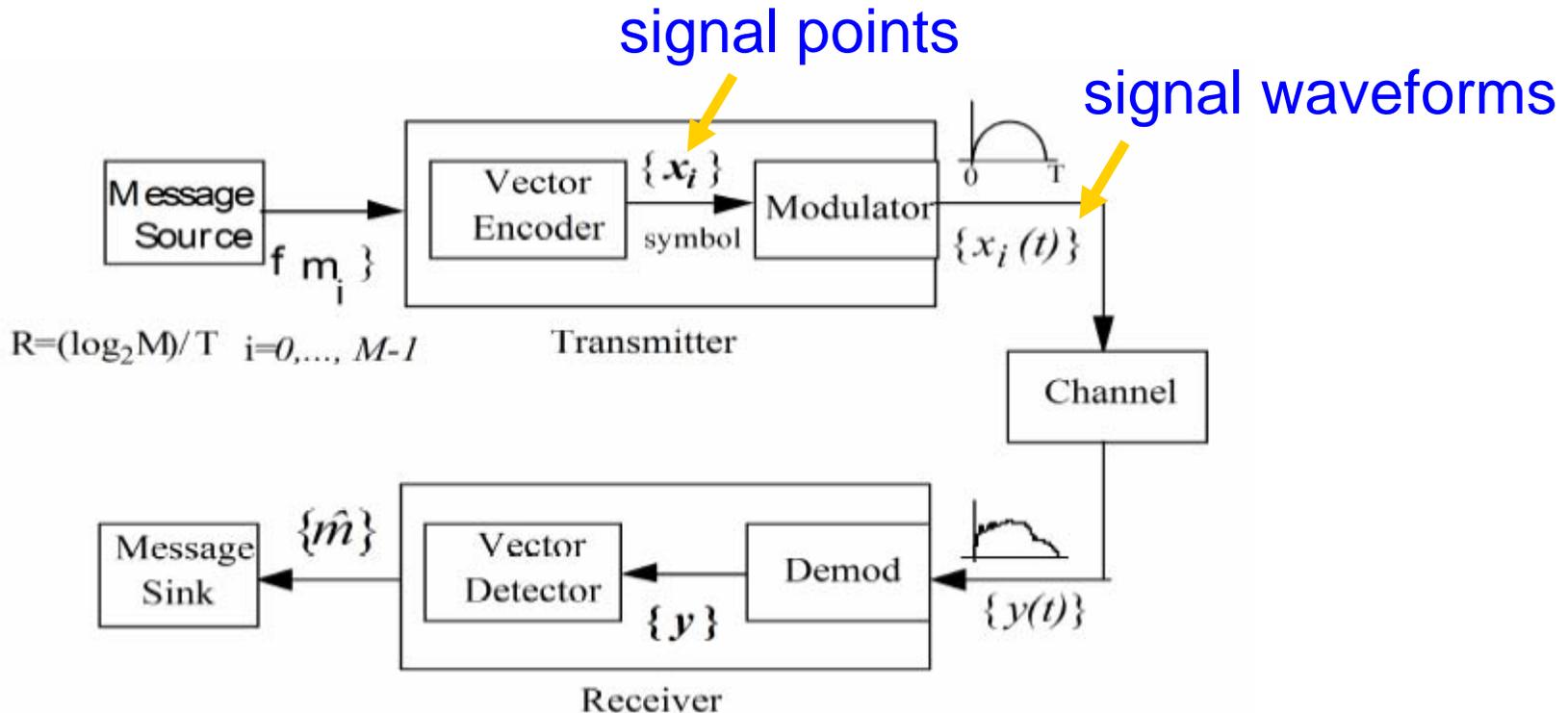
Vladimir Stojanović



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Discrete data transmission

- Messages are encoded into signal points

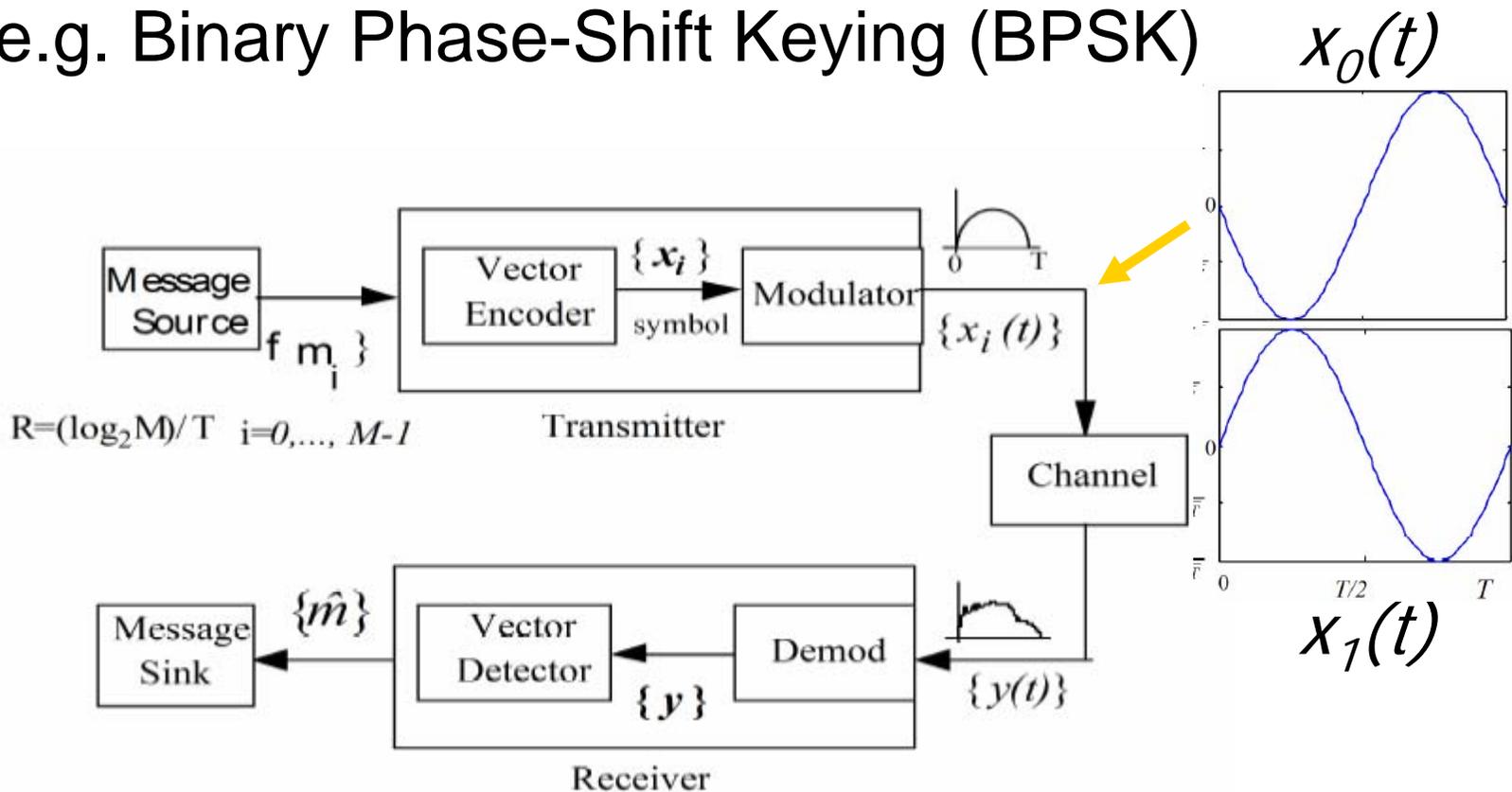


- Signal points are mapped to signal waveforms
 - Modulation

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Modulation and de-modulation

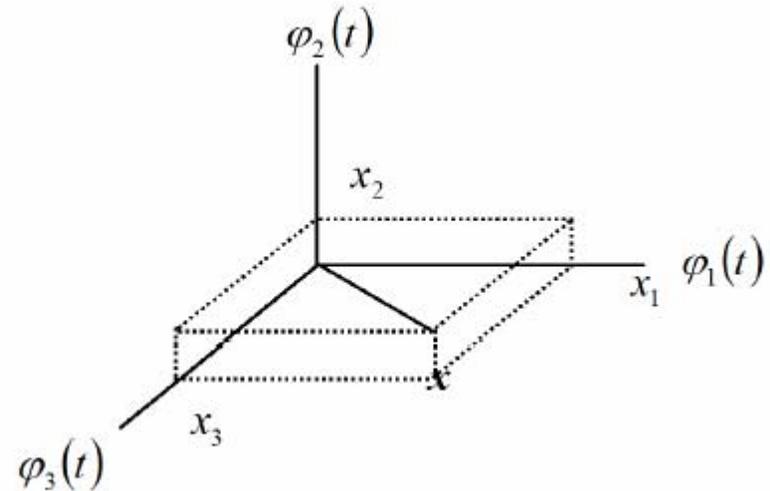
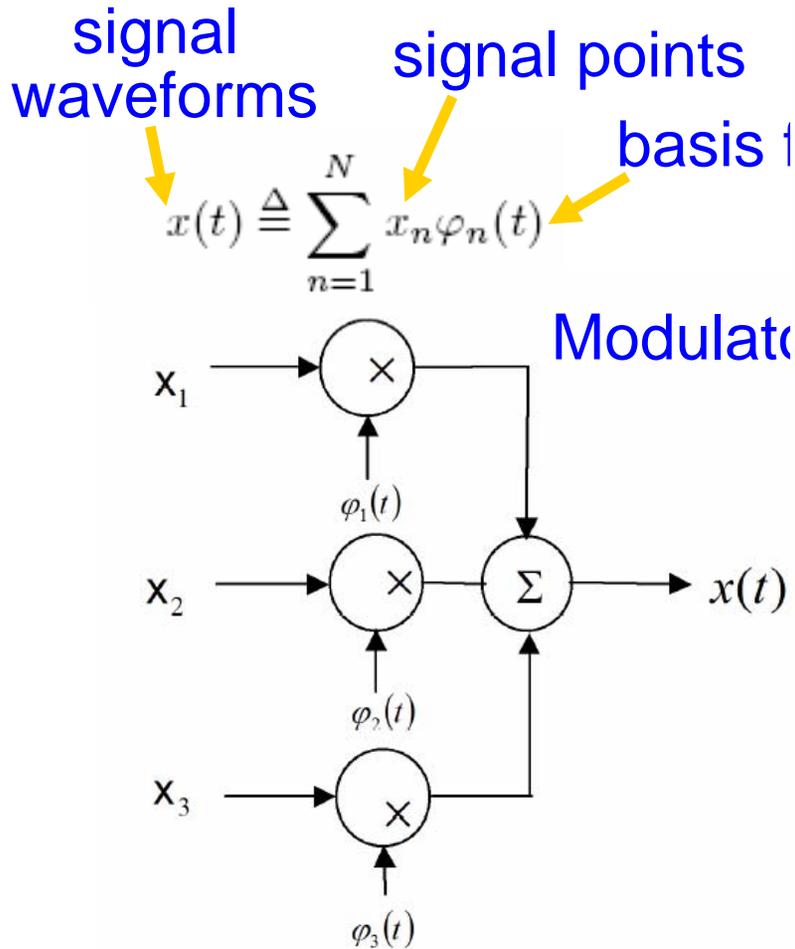
- e.g. Binary Phase-Shift Keying (BPSK)



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Vector signal representation

- Maps continuous signals to discrete vectors
 - Significantly simplifies system analysis



$$P_x \triangleq \frac{\mathcal{E}_x}{T}$$

$$\mathcal{E}_x \triangleq E [||x||^2] = \sum_{i=0}^{M-1} ||x_i||^2 p_x(i)$$

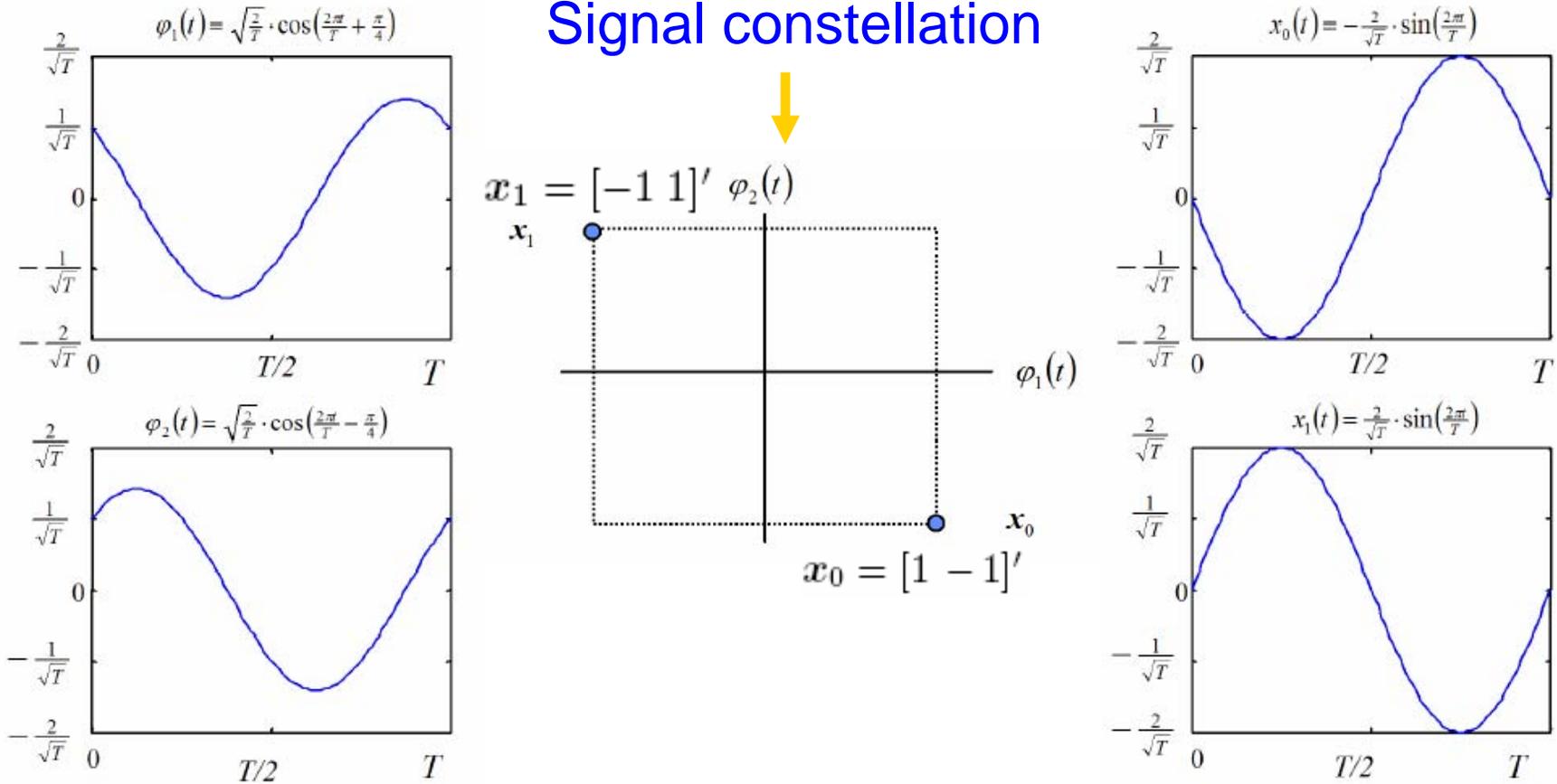
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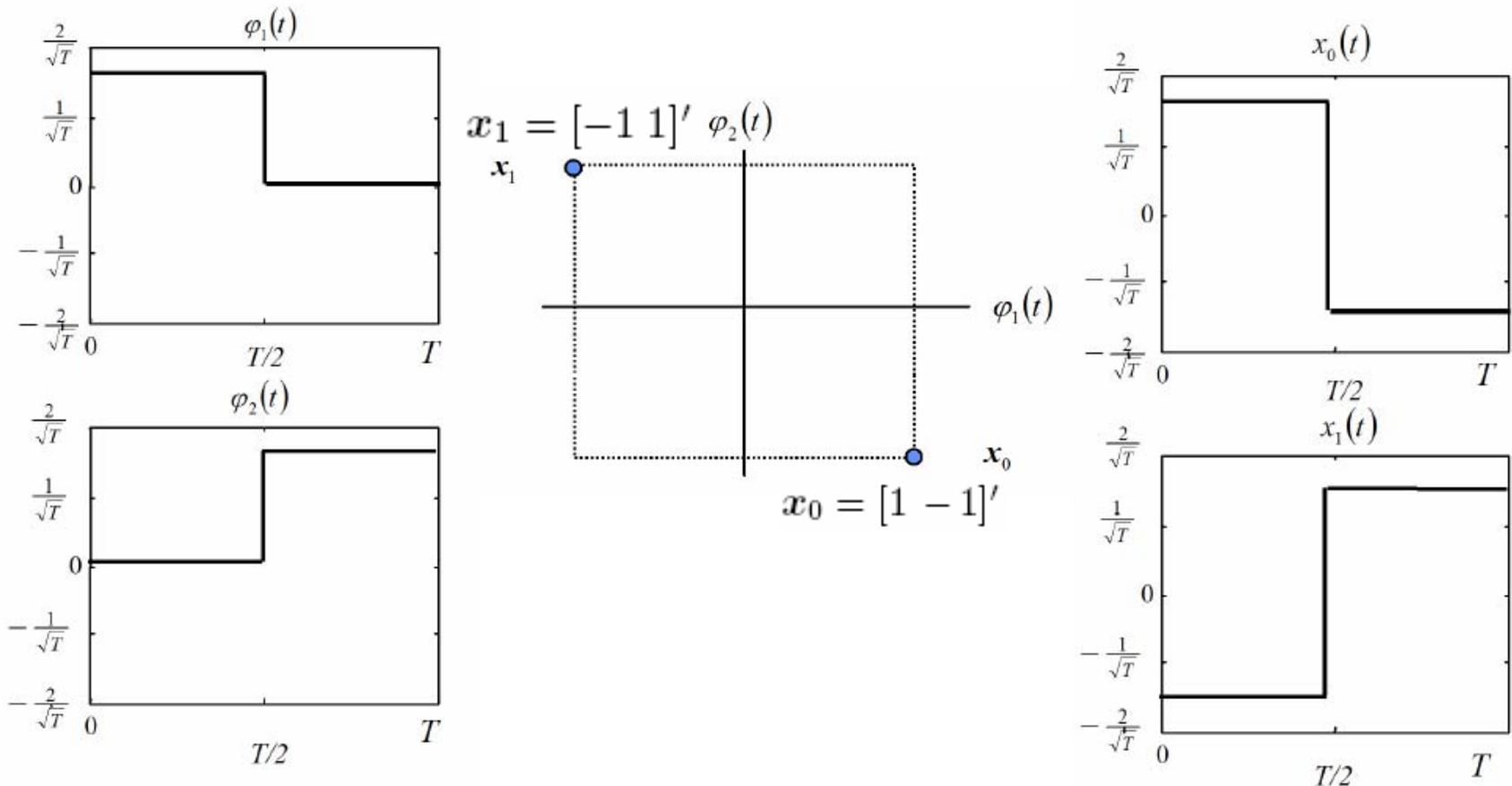
BPSK example

Signal constellation



- What is the information rate (R) of this modulation?

Manchester modulation example (Ethernet)



- Different waveforms can have same vector representations

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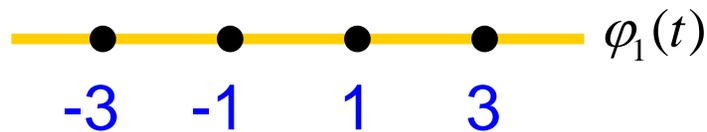
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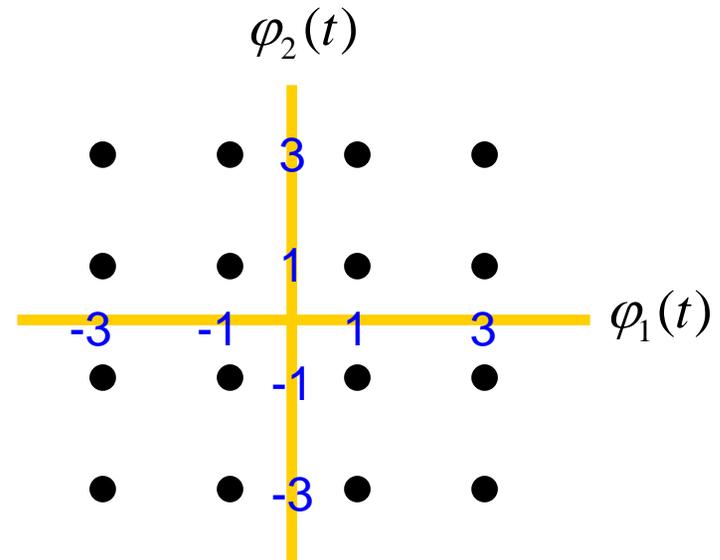
More constellations

Quadrature Amplitude Modulation (QAM)

Pulse Amplitude Modulation (PAM)



e.g. PAM4



e.g. 16-QAM

PAM and QAM have pulses as basis functions

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How do we choose basis functions?

- Need to be orthonormal – (b/c of demodulation)

$$\int_{-\infty}^{\infty} \varphi_m(t)\varphi_n(t)dt = \delta_{mn} = \begin{cases} 1 & m = n \\ 0 & m \neq n \end{cases}$$

- Inner products

- Continuous

$$\langle u(t), v(t) \rangle \triangleq \int_{-\infty}^{\infty} u(t)v(t)dt$$

- Discrete

$$\langle \mathbf{u}, \mathbf{v} \rangle \triangleq \mathbf{u}^* \mathbf{v} = \sum_{n=1}^N u_n v_n$$

- Invariant to choice of basis functions

$$u(t) = \sum_{n=1}^N u_n \varphi_n(t) \text{ and } v(t) = \sum_{n=1}^N v_n \varphi_n(t)$$

$$\langle u(t), v(t) \rangle = \int_{-\infty}^{\infty} u(t)v(t)dt = \int_{-\infty}^{\infty} \sum_{n=1}^N \sum_{m=1}^N u_n v_m \varphi_n(t)\varphi_m(t)dt$$

$$= \sum_{n=1}^N \sum_{m=1}^N u_n v_m \int_{-\infty}^{\infty} \varphi_n(t)\varphi_m(t)dt = \sum_{m=1}^N \sum_{n=1}^N u_n v_m \delta_{nm} = \sum_{n=1}^N u_n v_n$$

$$= \langle \mathbf{u}, \mathbf{v} \rangle \text{ QED.}$$

- Average energy of the constellation
 - Invariant to the choice of basis functions

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Constellation energy

- Implications of the inner product invariance to basis functions

$$\langle u(t), v(t) \rangle = \langle u, v \rangle$$

- If energy is a signal, it is the same regardless of the mod waveform used
 - As long as basis functions are orthogonal
 - Parseval's identity

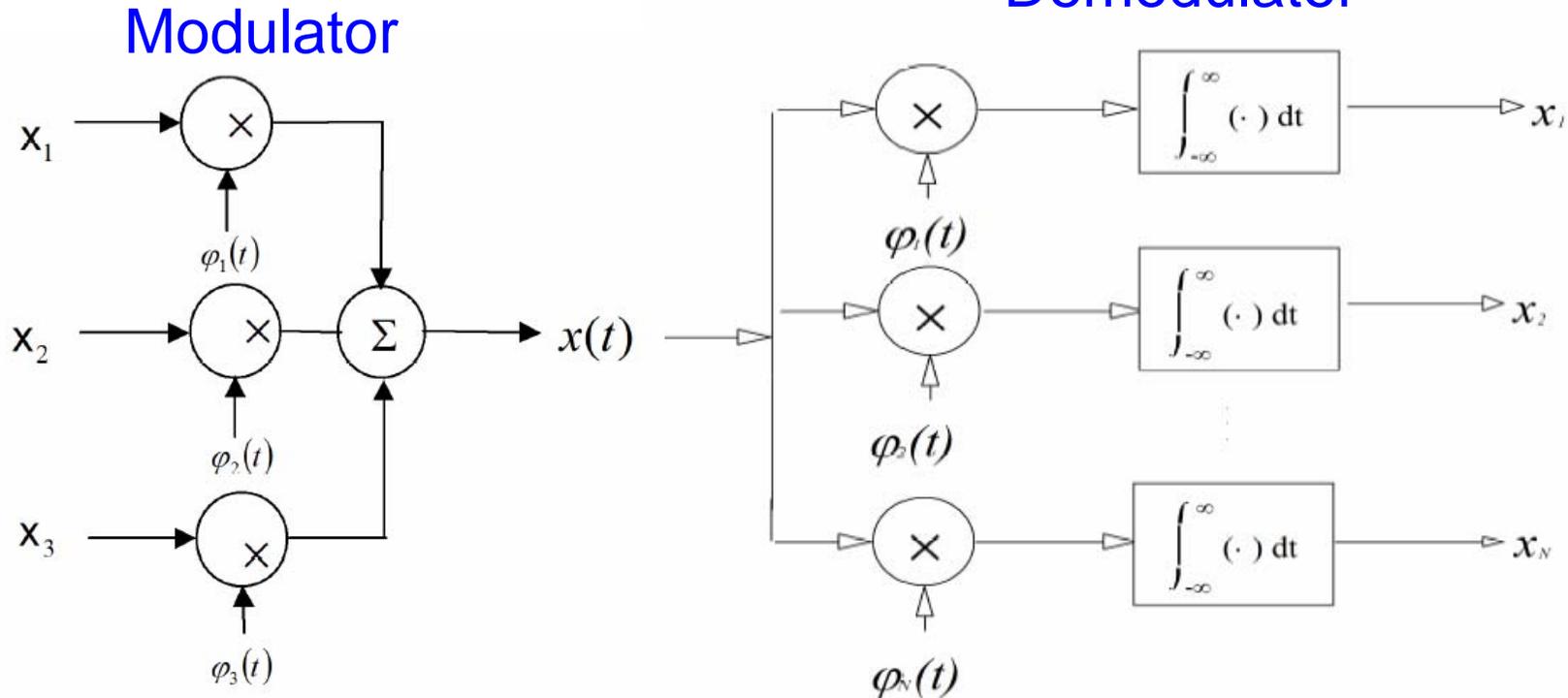
$$\mathcal{E}_x = E [\|\mathbf{x}\|^2] = E \left[\int_{-\infty}^{\infty} x^2(t) dt \right]$$

$$\begin{aligned} E [\langle u(t), v(t) \rangle] &= E [\langle \mathbf{x}, \mathbf{x} \rangle] \\ &= E \left[\sum_{n=1}^N x_n x_n \right] \\ &= E [\|\mathbf{x}\|^2] \\ &= \mathcal{E}_x \quad \text{QED.} \end{aligned}$$

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Correlative demodulator

$$\int_{-\infty}^{\infty} \varphi_m(t)\varphi_n(t)dt = \delta_{mn} = \begin{cases} 1 & m = n \\ 0 & m \neq n \end{cases}$$

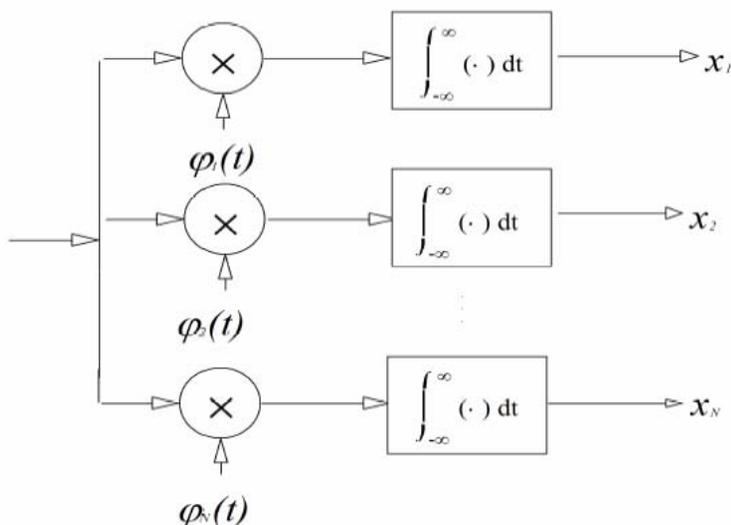


- ❑ **Straightforward demodulator implementation**
 - Use the fact that basis functions are orthogonal
 - Collect the signal energy
 - Hard to build in practice

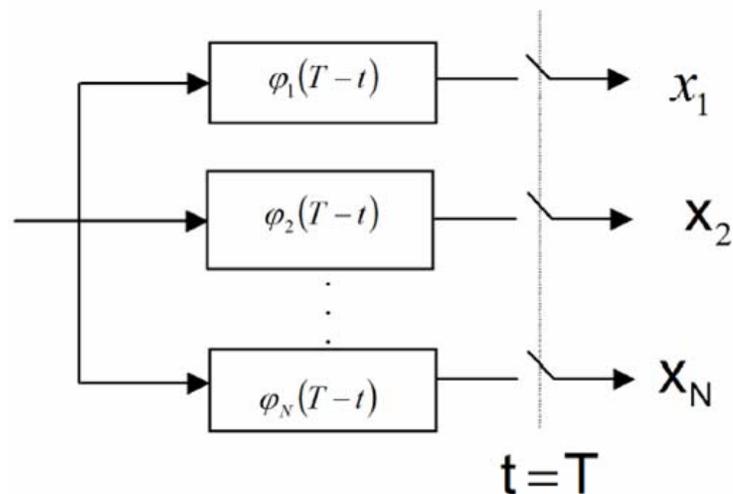
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Practical implementation

Correlative demodulator



Matched-filter demodulator



- Note $x_n = \int_0^T x(t)\varphi_n(t)dt$ equivalent to $x(t) * \varphi_n(T-t)|_{t=T}$
- Can implement with an “integrate-and-dump”

Summary

- ❑ In this course you'll be able to learn
 - Practical digital communication techniques
 - Hands-on, little math
 - Hardware implementations
 - Algorithmic transformations
 - Micro-architectures
 - ASIC flow and behavioral modeling
- ❑ In other words, everything you'll need to start building cutting-edge digital communication systems
- ❑ Started intro to digital communications
 - Modulation – signal constellation, basis functions
 - Demodulation – basis function invariance, matched-filter
 - Next – basics of detection, signalling on band-limited channels

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Sources

- ❑ VppSim/CppSim is a tool developed by prof. Michael Perrott
- ❑ Digital communications material is adapted from prof. John Cioffi's Stanford Course readers
 - <http://www.stanford.edu/class/ee379a,b,c/>