

# 6.302 Feedback Systems

Recitation 3: Block Diagrams (continued)

Prof. Joel L. Dawson

Quick review of Taylor Series Approximation. Suppose we have a complicated function  $f(\cdot)$ , whose value  $f(x)$  we happen to know. If the function  $f(\cdot)$  is well-behaved, we can write the equality (not merely approximation):

$$\begin{aligned} f(x + \Delta x) &= f(x) + \frac{1}{1!} \frac{df}{dx} \Delta x + \frac{1}{2!} \frac{d^2f}{dx^2} (\Delta x)^2 + \dots \\ &= f(x) + \sum_{k=1}^{\infty} \frac{1}{k!} \frac{d^k}{dx^k} (\Delta x)^k \end{aligned}$$

When we linearize, we approximate this equality as:

$$f(x + \Delta x) = f(x) + \Delta f \approx f(x) + \frac{df}{dx} \Delta x$$

So we identify  $\Delta f$  as  $\frac{df}{dx} \Delta x$

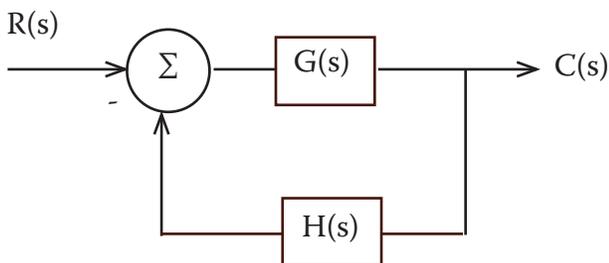
Similarly, suppose we have a function of three variables,  $f(x, y, z)$ . We still write:

$$\Delta f = \frac{\partial f}{\partial x} \Delta x + \frac{\partial f}{\partial y} \Delta y + \frac{\partial f}{\partial z} \Delta z$$

We use this result over and over again when we're working with nonlinear systems.

Point of emphasis from lecture: "Trading Gain for Desensitivity"

Let's go over this slowly...



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So how sensitive is the overall transfer function to changes in  $G$ ? Using our Taylor Series Approximation, we say

$$\begin{aligned}\Delta\left(\frac{C}{R}\right) &= \frac{d}{dG}\left(\frac{G}{1+GH}\right) \cdot \Delta G \\ &= \left[G \cdot \left(-\frac{H}{(1+GH)^2}\right) + \frac{1}{1+GH}\right] \cdot \Delta G \\ &= \left[-\frac{GH}{(1+GH)^2} + \frac{1}{1+GH}\right] \cdot \Delta G \\ &= \frac{1}{(1+GH)^2} \Delta G\end{aligned}$$

More useful to us is knowing the fractional change in the transfer function given a fractional change in  $G$ . This allows us to say things like, "A 10% change in  $G$  leads to only a 0.1% change in the overall transfer characteristic." So we divide both sides of our result by  $C/R$ :

$$\frac{\Delta(C/R)}{C/R} = \frac{\frac{1}{(1+GH)^2} \Delta G}{C/R}$$

$$\text{But } C/R = \frac{G}{1+GH} \dots$$

$$\frac{\Delta(C/R)}{C/R} = \frac{\frac{1}{(1+GH)^2} \Delta G}{\frac{1}{1+GH}}$$

$$\frac{\Delta(C/R)}{C/R} = \frac{1}{1+GH} \frac{\Delta G}{G}$$

For large  $GH$  ( $GH \gg 1$ ), we get

$$\frac{\Delta(C/R)}{C/R} \approx \frac{1}{GH} \frac{\Delta G}{G}$$

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What do we now mean by “trading gain for desensitivity?” For large values of  $GH$ , the closed-loop TF is approximately  $1/H$ . So, in some sense, the only gain that we really “need” could be provided by an open-loop amplifier with gain  $1/H$ . But the forward path gain that we have is  $G$ . The amount of gain, then, that we’re “throwing away” is

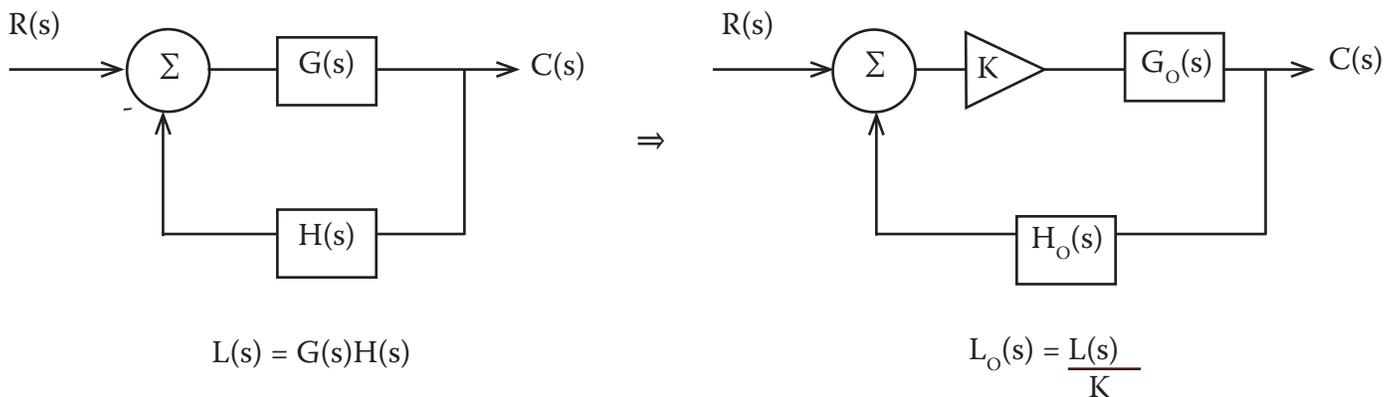
$$\frac{\text{Original forward path gain}}{\text{Gain that we need}} = \frac{G}{1/H} = GH$$

We can call  $GH$  the excess gain, or the gain that we’ve thrown away. But look: the gain that we’ve thrown away shows up again as a reduction of sensitivity to variations in  $G$ .

$$\frac{\Delta/(C/R)}{C/R} \approx \frac{1}{GH} \frac{\Delta G}{G}$$

Spent some time thinking about this.

There is often an implicit assumption in our analyses that our block diagrams can be drawn in the following way:



$L_o(s)$  is the normalized loop transmission.

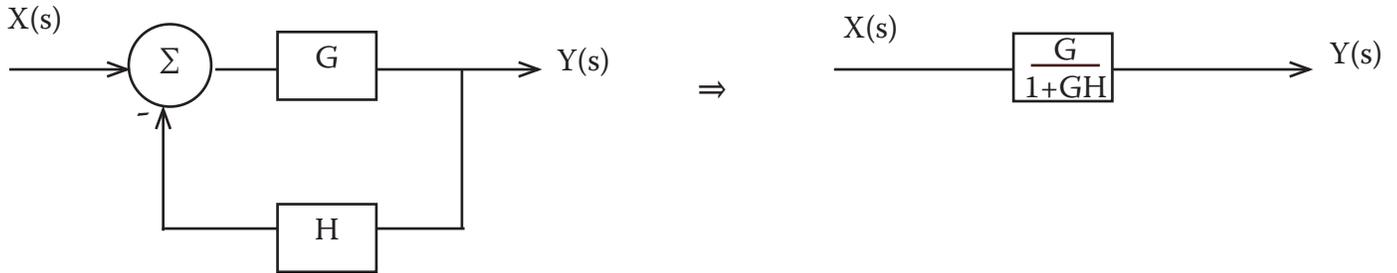
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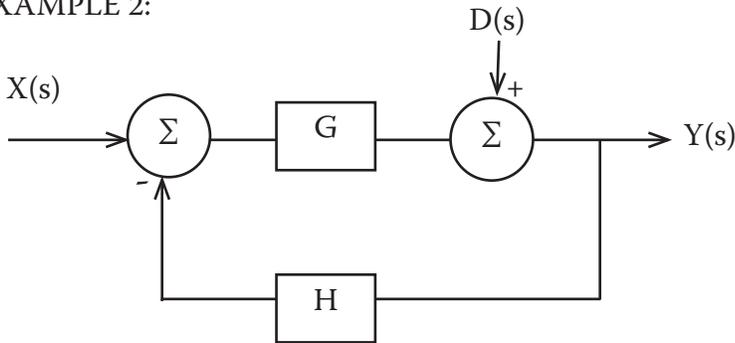
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Now we move on to an important skill, which is manipulating block diagrams. One of the most useful tricks you will use is “collapsing” a feedback loop using Black’s Formula:

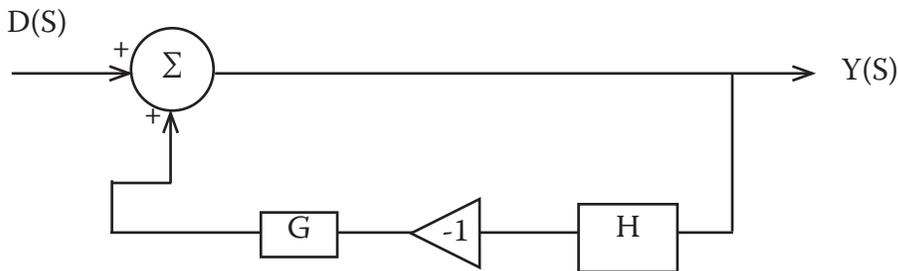
EXAMPLE 1:



EXAMPLE 2:



What is the transfer function  $Y(s)/D(s)$ ? Redraw:



Easy now!

$$\frac{Y(S)}{D(S)} = \frac{1}{1+GH}$$

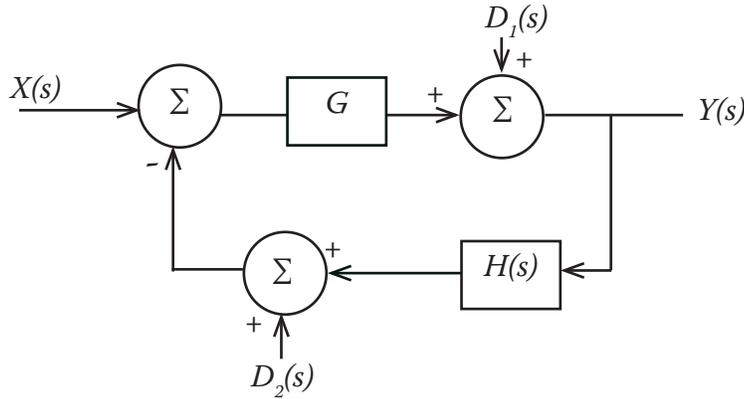
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## CLASS EXERCISE

In the following diagram, which disturbance bothers us the most? (HINT: Block diagram manipulation & Black's Formula is one, but not the only, way to reason this out.)



Other useful results: (graphical expressions of math)

① Block in series:  $X(s) \rightarrow [G_1] \rightarrow [G_2] \rightarrow Y(s) = X(s) \rightarrow [G_1 G_2] \rightarrow Y(s)$

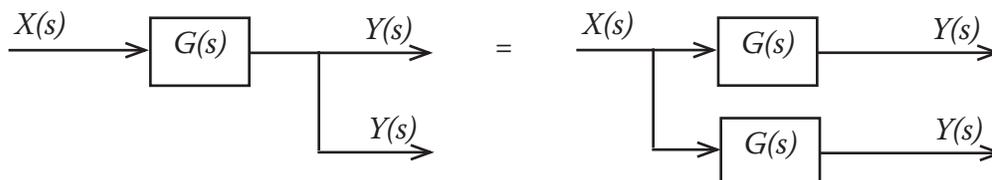
$y(t) = x(t) * g_1(t) * g_2(t) \xrightarrow{\text{LAPLACE TRANSFORM}} Y(S) = G_1(s)G_2(s)X(s)$

② Blocks in parallel:  $X(s) \rightarrow \left[ \begin{array}{c} [G] \\ [G] \end{array} \right] \rightarrow \Sigma \rightarrow Y(s) = X(s) \rightarrow [G_1 + G_2] \rightarrow Y(s)$

$Y(s) = G_1(s)X(s) + G_2(s)X(s)$

$= (G_1(s) + G_2(s))X(s)$

③ Blocks moved through a junction:

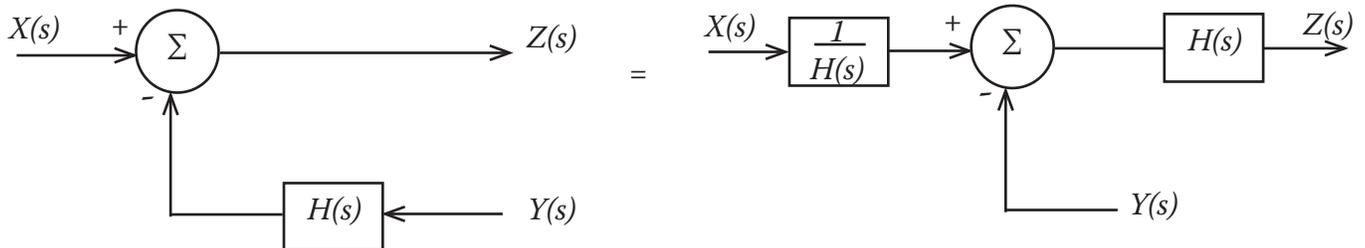


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④ Moving blocks through a summer:



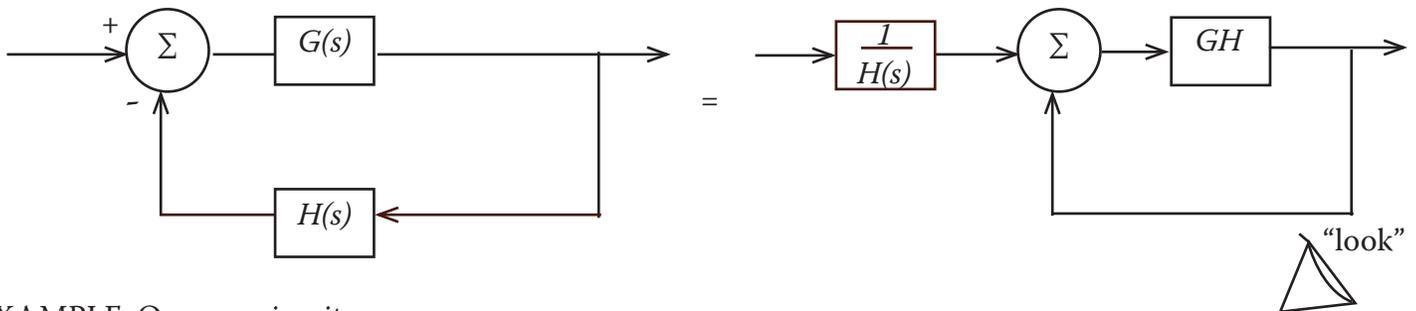
VERY USEFUL. Why does it work? Start with

$$(X - HY) = Z$$

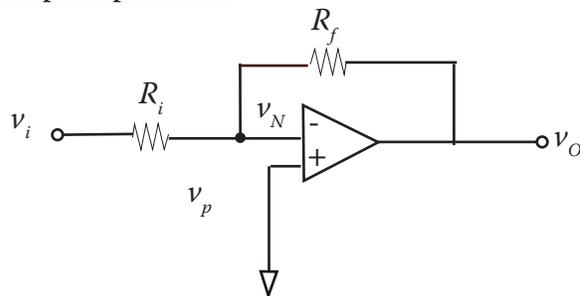
$$H \cdot 1/H (X - HY) = Z$$

$$H \left( \frac{1}{H} X - Y \right) = Z \quad \text{DONE.}$$

Useful because it lets us draw unity feedback diagrams quickly:



EXAMPLE: Op-amp circuits



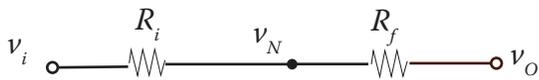
$$v_o = A(S) \cdot [v_p - v_N] = -A(S)v_N$$

must figure out  $v_N$ :  $\Longrightarrow$

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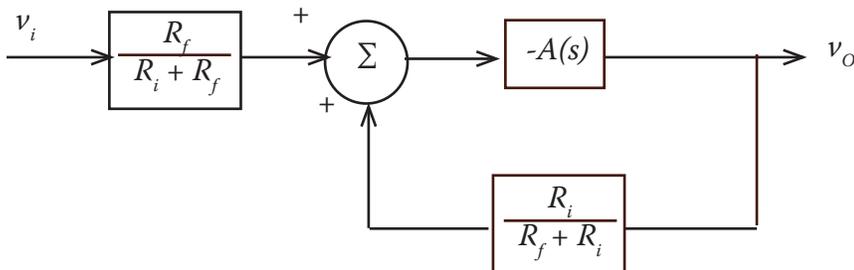
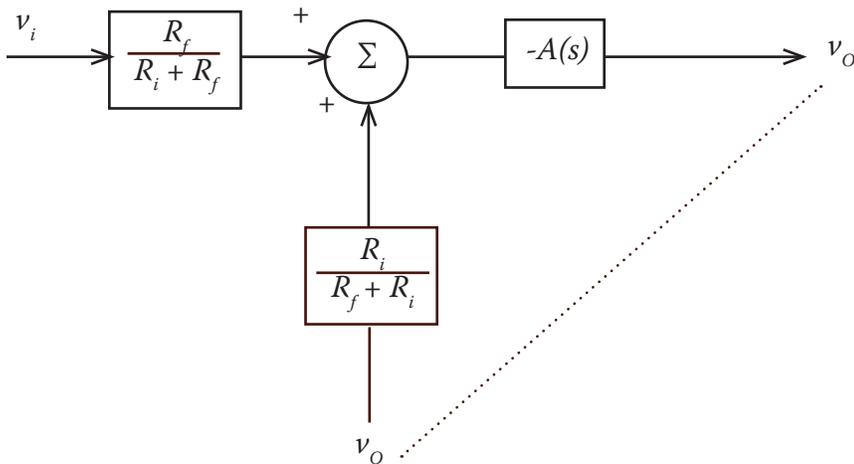
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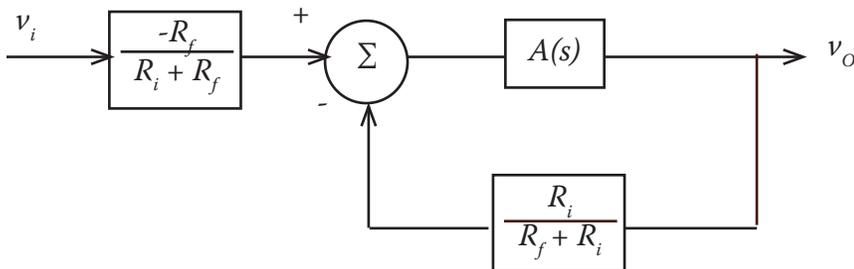
using superposition:

$$v_N = \frac{R_i}{R_i + R_f} v_o + \frac{R_f}{R_i + R_f} v_i$$

Block diagram; start with a summing junction =>



Now move minus sign through summing junction:



and draw as unity

feedback: =>

