

Massachusetts Institute of Technology

Department of Electrical Engineering and Computer Science  
6.245: MULTIVARIABLE CONTROL SYSTEMS

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## Problem Set 10 (due May 12, 2004) <sup>1</sup>

### Problem 10.1

For a cone  $\Delta = \{\Delta\}$  of complex  $n$ -by- $m$  matrices, and for a complex  $m$ -by- $n$  matrix  $M$ , the quantity  $\mu_{\Delta}(M)$  is defined by

$$\mu_{\Delta}(M) = (\inf\{\|\Delta\| : \Delta \in \Delta, \det(I - M\Delta) = 0\})^{-1}$$

(in particular,  $\mu_{\Delta}(M) = 0$  if  $I - M\Delta$  is invertible for all  $\Delta \in \Delta$ ). Such quantity, called *structured singular value* of  $M$  (where  $\Delta$  is what defines the “structure”), plays an important role in analysing robust stability.

When  $\Delta$  is the cone of *all* matrices,  $\mu_{\Delta}(M)$  equals the usual largest singular number of  $M$ . When  $\Delta$  is the set of all diagonal matrices with complex entries,  $\mu_{\Delta}(M) = \mu_{\mathbf{C}}(M)$  is called the *complex structured singular value*. When  $\Delta$  is the set of all diagonal matrices with real entries,  $\mu_{\Delta}(M) = \mu_{\mathbf{R}}(M)$  is called the *real structured singular value*.

Let  $\Delta$  be the cone of diagonal matrices with complex entries  $z_i$  such that  $\operatorname{Re}(z_i) \geq |\operatorname{Im}(z_i)|$ . Our objective is to produce a method for estimating  $\mu_{\Delta}(M)$ , based on semidefinite programming.

- (a) Describe the set of all quadratic constraints which are satisfied for the relation between two complex numbers  $w$  and  $v$  satisfying  $w = zv$ , where  $\operatorname{Re}(z) \geq |\operatorname{Im}(z)|$ .
- (b) Use the result of (a) to develop an LMI optimization algorithm for calculating an upper bound  $\hat{\mu}_{\Delta}(M)$  of  $\mu_{\Delta}(M)$  for an arbitrary  $n$ -by- $n$  complex matrix  $M$ .

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<sup>1</sup>Version of May 6, 2004

- (c) Test the upper bound on a set of randomly generated 3-by-3 and 10-by-10 complex matrices  $M$ . Compare  $\hat{\mu}_{\Delta}(M)$  with  $\mu_{\mathbf{C}}(M)$ , which can be estimated using MATLAB's

```
bounds=mu(M)
```

(the two components of output `bounds` will be an upper and a lower bound of  $\mu_{\mathbf{C}}(M)$ ).

### Problem 10.2

In the design setup shown on Figure 10.1,  $r$  is the reference signal,  $y$  is measured plant output,  $u$  is control action, and  $e = y - r$  is tracking error. Transfer functions  $W_1$  (reference

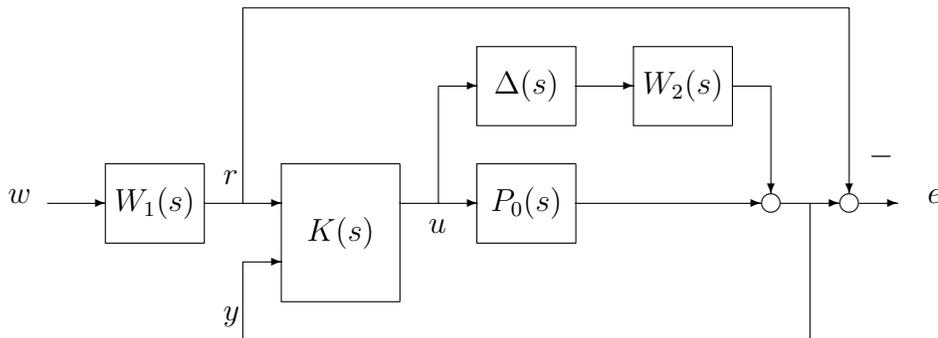


Figure 10.1: Design setup for Problem 10.2

signal shaping filter),  $P_0$  (nominal plant model), and  $W_2$  (uncertainty weight) are given:

$$W_1(s) = \frac{1}{1 + 20s}, \quad W_2(s) = r \frac{s + 1}{s + 10}, \quad P_0(s) = \frac{s - 2}{s^2 - 1},$$

where  $r > 0$  is a parameter.  $\Delta = \Delta(s)$  is the normalized uncertainty, ranging over the set of all stable transfer functions with  $\|\Delta\|_{\infty} \leq 1$ . The objective is to design an LTI controller  $K = K(s)$  of order not larger than 8, which stabilizes the feedback system for all possible  $\Delta$  (“robust stabilization”), while trying to make the worst (again, over all possible  $\Delta$ ) closed loop H-Infinity norm from  $w$  to  $e$  (“robust performance”) as small as possible.

- (a) Find the maximal value  $r_0$  of those  $r > 0$  for which robust stabilization is possible.

- (b) For  $r = 0.1r_0$ , use D-K iterations of H-Infinity optimization and semidefinite programming to minimize robust performance.