

Topics for today:

Handshaking

'Concurrent' and 'Sequential' statements

(Another example: a counter)

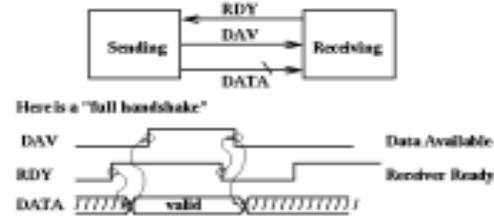
Yet another example: a small ALU

Brief discussion of resource usage

Handshaking

Required when multiple lines of input are involved

This is a 'full handshake' Note that both positive going and negative going transitions are important in both directions



Receiver indicates ready to receive data by setting RDY

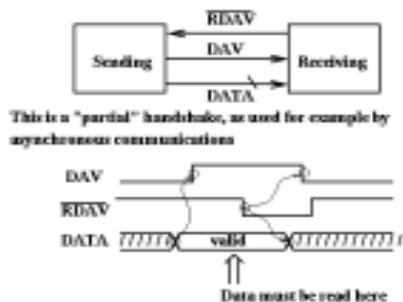
Sender sets data valid then sets DAV

Receiver reads data then clears RDY

Sender acknowledges by clearing DAV

A Less Elaborate handshake

This is often used in things like UARTs which must deal with asynchronous data streams that they do not control



Sender stabilizes data and sets DAV

Receiver reads data and clears RDAV

Sender de-asserts data and clears DAV

Typically, sender does not wait for /RDAV before setting new data. This can be used for detecting 'overrun' errors.

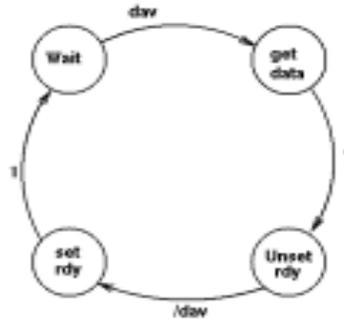
We should be able to describe the sending and receiving agents as simple finite state machines. Here is the FSM at the Sending end: (Full handshake)



```
library ieee;
use ieee.std_logic_1164.all;

entity fullsend is
generic (size: integer := 4);
port (rdy, clk : in std_logic;
      datin : in std_logic_vector(size-1 downto 0);
      dav : out std_logic;
      datout : out std_logic_vector(size - 1 downto 0));
end fullsend;
```

And here is the FSM for the receiving end:



```

library ieee;
use ieee.std_logic_1164.all;

entity fullrcv is
  generic (size: integer := 4);
  port (dav, rclk : in std_logic;
        datin : in std_logic_vector(size-1 downto 0);
        rdy : out std_logic;
        datout : out std_logic_vector(size - 1 downto 0));
end fullrcv;
  
```

```

architecture behavioral of fullsend is
  type StateType is (wt, dat, d_av, r_dy);
  attribute enum_encoding of StateType: type is "00 01 11 10";
  signal state : StateType;
begin
  dav <= '1' when (state = d_av) or (state = r_dy) else '0';
  handshake : process(clk)
  begin
    if rising_edge(clk) then
      case state is
        when wt =>
          if rdy = '1' then
            state <= dat;
          else
            state <= wt;
          end if;
        when dat =>
          datout <= datin;
          state <= d_av;
        when d_av =>
          state <= r_dy;
        when r_dy =>
          if rdy = '0' then
            state <= wt;
          else
            state <= r_dy;
          end if;
        end case;
      end if;
    end process handshake;
  end;
end;
  
```

```

architecture behavioral of fullrcv is
  type StateType is (w_dav, datav, r_rdy, wt_ndav);
  attribute enum_encoding of StateType: type is "00 01 11 10";
  signal state : StateType;
begin
  rdy <= '1' when (state = w_dav) or (state = datav) else '0';
  handshake : process(rclk)
  begin
    if rising_edge(rclk) then
      case state is
        when w_dav =>
          if dav = '1' then
            state <= datav;
          else
            state <= w_dav;
          end if;
        when datav =>
          datout <= datin;
          state <= r_rdy;
        when r_rdy =>
          state <= wt_ndav;
        when wt_ndav =>
          if dav = '0' then
            state <= w_dav;
          else
            state <= wt_ndav;
          end if;
        end case;
      end if;
    end process handshake;
  end;
end;
  
```

Here is an alternative way of writing an emulator for the '163 counter

- This is a register which can hold 4 bits
- Counts when P=T=1, holds when P*T=0
- Loads data when /LD = 0
- Clears data when /CL = 0
- All of these are synchronous: occur only on clock edges (positive edges)
- Daisy-chaining is possible: RCO connects to T of next most significant ctr
- RCO is T * Q3 * Q2 * Q1 * Q0

Here is an entity statement for this part

```

-- '163 emulator
library ieee;
use ieee.std_logic_1164.all;
use work.std_arith.all;

entity ctr is
  generic (size: integer := 4);
  port (n_clr, n_ld, p, t, clk : in std_logic;
        data: in std_logic_vector(size-1 downto 0);
        count: out std_logic_vector(size-1 downto 0);
        rco : out std_logic);
end ctr;
  
```

```

architecture behavioral of ctr is
  signal cnt_int : std_logic_vector(size - 1 downto 0);
  signal int_cnt : std_logic_vector(size - 1 downto 0); -- internal count
  signal all_ones : std_logic_vector(size downto 0);
begin -- behavioral
  all_ones <= (others => '1');
  rco <= '1' when (t & cnt_int) = all_ones else '0';
  count <= cnt_int;

  logical:process(p, t, n_clr, n_ld, cnt_int, data)
  begin
    if n_clr = '0' then
      int_cnt <= (others => '0');
    elsif n_ld = '0' then
      int_cnt <= data;
    elsif p = '0' or t = '0' then
      int_cnt <= cnt_int;
    else
      int_cnt <= cnt_int + 1;
    end if;
  end process logical;
  state_transition:process(clk)
  begin
    if rising_edge(clk) then
      cnt_int <= int_cnt;
    end if;
  end process state_transition;
end behavioral;

```

Note two processes here:
 One has the combinatorics associated with the logic in the part.
 The other has the state transition dynamics associated with the clock edge.

DESIGN EQUATIONS (11:32:34)

```

rco =
  t * count_2.Q * count_1.Q * count_0.Q * count_3.Q

count_3.D =
  t * count_2.Q * count_1.Q * count_0.Q * n_clr * n_ld * p *
  /count_3.Q
  + n_clr * n_ld * /p * count_3.Q
  + /count_0.Q * n_clr * n_ld * count_3.Q
  + /count_1.Q * n_clr * n_ld * count_3.Q
  + /count_2.Q * n_clr * n_ld * count_3.Q
  + /t * n_clr * n_ld * count_3.Q
  + n_clr * /n_ld * data_3

count_3.C =
  clk

count_2.D =
  t * /count_2.Q * count_1.Q * count_0.Q * n_clr * n_ld * p
  + count_2.Q * n_clr * n_ld * /p
  + count_2.Q * /count_0.Q * n_clr * n_ld
  + count_2.Q * /count_1.Q * n_clr * n_ld
  + /t * count_2.Q * n_clr * n_ld
  + n_clr * /n_ld * data_2

```

```

count_2.C =
  clk

count_1.D =
  t * /count_1.Q * count_0.Q * n_clr * n_ld * p
  + count_1.Q * n_clr * n_ld * /p
  + count_1.Q * /count_0.Q * n_clr * n_ld
  + /t * count_1.Q * n_clr * n_ld
  + n_clr * /n_ld * data_1

count_1.C =
  clk

count_0.D =
  t * /count_0.Q * n_clr * n_ld * p
  + count_0.Q * n_clr * n_ld * /p
  + /t * count_0.Q * n_clr * n_ld
  + n_clr * /n_ld * data_0

count_0.C =
  clk

```

These are just about what you would have expected.
 Note a lot of fluff has been optimized away.

RESOURCE ALLOCATION (11:32:34)

Information: Macrocell Utilization.

Description	Used	Max
Dedicated Inputs	8	8
Clock/Inputs	1	1
Enable/Inputs	0	1
Output Macrocells	5	8

14 / 18 = 77 %

Information: Output Logic Product Term Utilization.

Node#	Output Signal Name	Used	Max
12	count_3	7	8
13	count_2	6	8
14	count_1	5	8
15	count_0	4	8
16	rco	1	8
17	Unused	0	8
18	Unused	0	8
19	Unused	0	8

23 / 64 = 35 %

This was implemented on a 16V8
 Here are some numbers relating to how much of the resources of that part we used.

```

library ieee;
use ieee.std_logic_1164.all;
use work.std_arith.all; -- needed for integer + signal
entity test_tri is
  port(clk, oe, cnt_enb : in std_logic;
        counter : buffer std_logic_vector(3 downto 0);
        data : inout std_logic_vector(3 downto 0));
end test_tri;
architecture foo of test_tri is
-- signal counter : std_logic_vector(3 downto 0);
begin
  process (oe, counter)
  begin
    if (oe = '1') then data <= counter;
    else
      data <= "ZZZZ"; -- N.B. Z must be UPPERCASE!
    end if;
  end process;
  process (clk)
  begin
    if rising_edge(clk) then
      if (oe = '0') and (cnt_enb = '1') then
        counter <= counter + 1;
      end if;
    end if;
  end process;
end architecture foo;

```

Simulation of Tri-State as an Output

Note that data(3 downto 0) are white (meaning an input) when oe is low.



Now we are going to consider a strictly combinatoric circuit: an Arithmetic Logic Unit (ALU)

It takes 2 numbers (quite narrow in this case: 2 bits each)
(Plus a carry-in bit)

And can add, subtract and shift left

This can be done in more than one way. Consider addition:

1. a_int <= '0' & a
b_int <= '0' & b
if c_in = 0, c <= a_int + b_int
if c_in = 1, c <= a_int + b_int + 1
- 2.. a_int <= '0' & a & c_in
b_int <= '0' & b & c_in
c_int <= a_int + b_int
c <= c_int(width downto 1)

These have differences in the way they are implemented, and when we get to actual implementation of the full alu we will find yet another one

```

library ieee;
use ieee.std_logic_1164.all;
use work.std_arith.all; -- needed for integer + signal
entity alu is
  port(cin : in std_logic;
        a, b : in std_logic_vector(1 downto 0);
        alu_ctl : in std_logic_vector(1 downto 0);
        c : out std_logic_vector(2 downto 0));
end alu;

```

Here is one architecture for the adder

```

architecture justright of alu is
  signal a_int, b_int : std_logic_vector(2 downto 0);
  constant add : std_logic_vector(1 downto 0) := "00";
  constant sub : std_logic_vector(1 downto 0) := "01";
  constant shift: std_logic_vector(1 downto 0) := "10";
begin
  a_int <= '0' & a;
  b_int <= '0' & b;
  small_alu: process(a_int, b_int, cin, alu_ctl)
  begin
    case alu_ctl is
      when add => if cin = '0'
                    then c <= a_int + b_int;
                    else c <= a_int + b_int + 1;
                  end if;
      when sub => c <= a_int - b_int;
      when shift => if cin = '0'
                      then c <= a_int + a_int;
                      else c <= a_int + a_int + 1;
                    end if;
      when others => c <= (others => '-');
    end case;
  end process small_alu;
end architecture justright;

```

Note the carry bit is used to determine which expression to evaluate: in logic it is a kind of multiplexor. The 'opcode' is another multiplexor: in this case a 3:1. These have overhead.

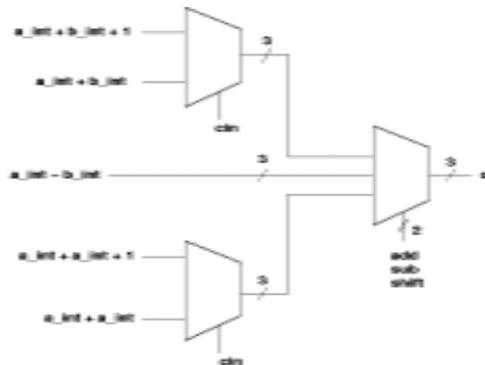
Here is the second architecture: does the same thing...

```
architecture justright of alu is
  signal a_int, b_int, c_int : std_logic_vector(3 downto 0);
  constant add : std_logic_vector(1 downto 0) := "00";
  constant sub : std_logic_vector(1 downto 0) := "01";
  constant shift: std_logic_vector(1 downto 0) := "10";
begin
  a_int <= '0' & a & cin;
  b_int <= '0' & b & cin;
  small_alu: process(a_int, b_int, cin, alu_ctl)
  begin
    case alu_ctl is
      when add => c_int <= a_int + b_int;
      when sub => c_int <= a_int - b_int;
      when shift => c_int <= a_int + a_int;
      when others => c_int <= (others => '-');
    end case;
  end process small_alu;
  c <= c_int(3 downto 1);
end architecture justright;
```

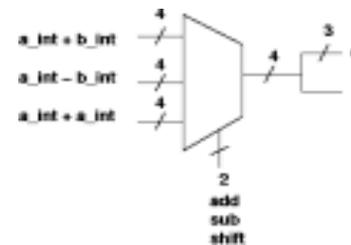
This may or may not have more overhead. Note that by adding Cin to both inputs (the first place after the binary point is of value) we save at least some notational overhead. It cancels on subtract. But we have to discard the rightmost bit at the end

```
architecture justright of alu is
  signal a_int, b_int, c_int : std_logic_vector(3 downto 0);
  signal a_1, n_b, upper, lower : std_logic_vector(3 downto 0);
  constant add : std_logic_vector(1 downto 0) := "00";
  constant sub : std_logic_vector(1 downto 0) := "01";
  constant shift: std_logic_vector(1 downto 0) := "10";
begin
  a_int <= '0' & a & cin;
  b_int <= '0' & b & cin;
  a_1 <= '0' & a & '1';
  n_b <= '1' & (not b) & '1';
  upper: process(a_int, a_1, alu_ctl)
  begin
    case alu_ctl is
      when add => upper <= a_int;
      when sub => upper <= a_1;
      when shift => upper <= a_int;
      when others => upper <= (others => '-');
    end case;
  end process upper;
  lower: process(a_int, b_int, n_b, alu_ctl)
  begin
    case alu_ctl is
      when add => lower <= b_int;
      when sub => lower <= n_b;
      when shift => lower <= a_int;
      when others => lower <= (others => '-');
    end case;
  end process lower;
  c_int <= upper + lower;
  c <= c_int(3 downto 1);
end architecture justright;
```

Here our final operation is just an addition, so we do more work in the earlier stages, such as doing the two's complement for the negative of b. We use the same trick for concatenating the carry in bit to both halves of the addition.



Here is a schematic of the way the first of these schemes is implemented. The final selection is a 9:3 MUX, while there are two 6:3 MUXes ahead of it. And ahead of that are some simple combinatoric circuits to generate the sums.

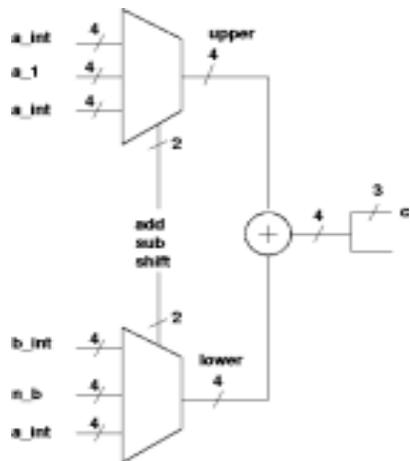


Here is the second of the two schemes

There still is a MUX: this time of 12:4, but it still has only two 'steering' bits.

We discard a bit in the final result

And we also have to construct the signals that come to this MUX from the left.



This is the third option: here we have two 12:4 MUXes and only a simple addition to reconstruct things.

Some logic is required to put together the signals at the left.

We discard a bit at the output

Here is the first of the three schemes
Information: Macrocell Utilization.

Description	Used	Max
Dedicated Inputs	1	1
Clock/Inputs	4	4
I/O Macrocells	7	64
Buried Macrocells	3	64
PIM Input Connects	12	312
27 / 445 = 6 %		

	Required	Max (Available)
CLOCK/LATCH ENABLE signals	0	4
Input REG/LATCH signals	0	69
Input PIN signals	5	5
Input PINS using I/O cells	2	2
Output PIN signals	5	62
Total PIN signals	12	69
Macrocells Used	8	128
Unique Product Terms	43	640

This is the second of the three schemes

Information: Macrocell Utilization.

Description	Used	Max
Dedicated Inputs	1	1
Clock/Inputs	4	4
I/O Macrocells	6	64
Buried Macrocells	2	64
PIM Input Connects	12	312
25 / 445 = 5 %		

	Required	Max (Available)
CLOCK/LATCH ENABLE signals	0	4
Input REG/LATCH signals	0	69
Input PIN signals	5	5
Input PINS using I/O cells	2	2
Output PIN signals	4	62
Total PIN signals	11	69
Macrocells Used	6	128
Unique Product Terms	28	640

This is the third of the three schemes

Information: Macrocell Utilization.

Description	Used	Max
Dedicated Inputs	1	1
Clock/Inputs	4	4
I/O Macrocells	5	64
Buried Macrocells	1	64
PIM Input Connects	8	312
19 / 445 = 4 %		

	Required	Max (Available)
CLOCK/LATCH ENABLE signals	0	4
Input REG/LATCH signals	0	69
Input PIN signals	5	5
Input PINS using I/O cells	2	2
Output PIN signals	3	62
Total PIN signals	10	69
Macrocells Used	4	128
Unique Product Terms	28	640