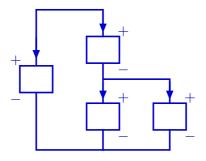
# 6.01: Introduction to EECS I

Op-Amps

#### Last Time: The Circuit Abstraction

**Circuits** represent systems as connections of elements

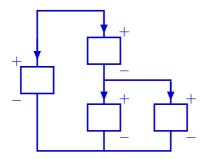
- through which currents (through variables) flow and
- across which voltages (across variables) develop.



## Last Time: Analyzing Circuits

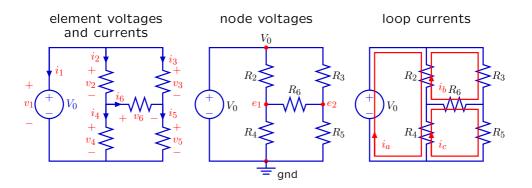
Circuits are analyzed by combining three types of equations.

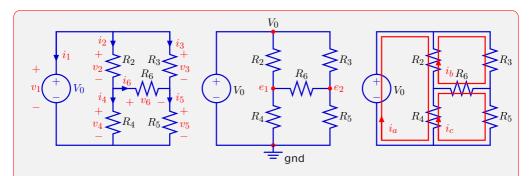
- KVL: sum of voltages around any closed path is zero.
- KCL: sum of currents out of any closed surface is zero.
- Element (constitutive) equations
  - resistor: V = IR
  - voltage source:  $V = V_0$
  - current source:  $I = I_0$



#### Last Time: Analyzing Circuits

Many KVL and KCL equations are redundant. We looked at three methods to systematically identify a linearly independent set.





#### How many of the following are true?

- 1.  $v_1 = v_2 + v_6 + v_5$
- 2.  $v_6 = e_1 e_2$
- 3.  $i_6 = (e_1 e_2)/R_6$
- 4.  $i_6 = i_b i_c$
- 5.  $v_6 = (i_b i_c)R_6$

How many of the following are true? 3

1. 
$$v_1 = v_2 + v_6 + v_5$$
  $\checkmark$ 
2.  $v_6 = e_1 - e_2$   $\checkmark$ 
3.  $i_6 = (e_1 - e_2)/R_6$   $\checkmark$ 

4. 
$$i_6 = i_b - i_c$$
  $\times$   $i_6 = i_c - i_b$   
5.  $v_6 = (i_b - i_c)R_6$   $\times$   $v_6 = (i_c - i_b)R_6$ 

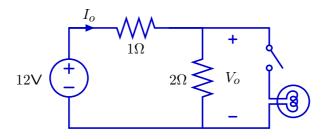
#### **Node Voltages with Component Currents**

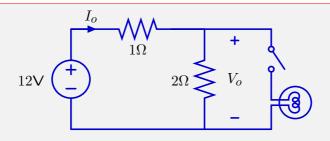
We will study a variation of the node method (NVCC) in software lab today.

#### **Interaction of Circuit Elements**

Circuit design is complicated by interactions among the elements. Adding an element changes voltages & currents **throughout** circuit.

Example: closing a switch is equivalent to adding a new element.

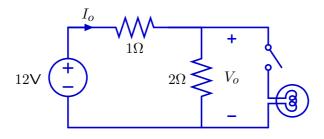




How does closing the switch affect  $V_o$  and  $I_o$ ?

- 1.  $V_o$  decreases,  $I_o$  decreases
- 2.  $V_o$  decreases,  $I_o$  increases
- 3.  $V_o$  increases,  $I_o$  decreases
- 4.  $V_o$  increases,  $I_o$  increases
- 5. could be any of above, depending on bulb resistance

Start by computing  $V_o$  and  $I_o$  when the switch is open.



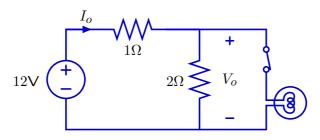
Calculate  $V_o$  using voltage divider relation:

$$V_o = \frac{2\Omega}{1\Omega + 2\Omega} \, 12 \mathsf{V} = 8 \mathsf{V}$$

Calculate  $I_o$  by lumping resistors into series equivalent:

$$I_o = \frac{12\mathsf{V}}{1\Omega + 2\Omega} = 4\mathsf{A}$$

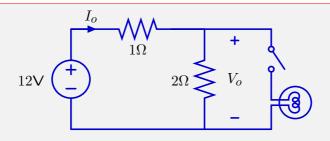
Now compute  $V_o$  and  $I_o$  when the switch is closed.



Assume the light bulb can be represented by a resistor R ( $0 < R < \infty$ ). Then R is in parallel with the  $2\Omega$  resistor.

$$V_o = \frac{2\Omega||R|}{1\Omega + 2\Omega||R|} 12V = \frac{\frac{2\Omega \times R}{2\Omega + R}}{1\Omega + \frac{2\Omega \times R}{2\Omega + R}} 12V = \frac{2R}{2\Omega + 3R} 12V \le 8V$$

$$I_o = \frac{12\mathsf{V}}{1\Omega + 2\Omega||R} = \frac{12\mathsf{V}}{1\Omega + \frac{2\Omega \times R}{2\Omega + R}} = \frac{2\Omega + R}{2\Omega + 3R} \cdot 12\mathsf{A} \ge 4\mathsf{A}$$

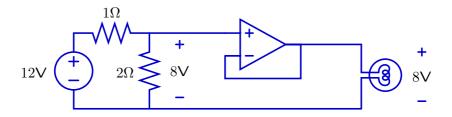


How does closing the switch affect  $V_o$  and  $I_o$ ? 2

- 1.  $V_o$  decreases,  $I_o$  decreases
- 2.  $V_o$  decreases,  $I_o$  increases
- 3.  $V_o$  increases,  $I_o$  decreases
- 4.  $V_o$  increases,  $I_o$  increases
- 5. could be any of above, depending on bulb resistance

## **Buffering with Op-Amps**

Interactions between elements can be reduced (or eliminated) by using an op-amp as a **buffer**.



This op-amp circuit produces an output voltage equal to its input voltage (8V) while having no effect on the left part of the circuit.

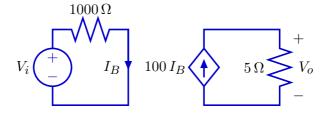
Today: how to analyze and design op-amp circuits

#### **Dependent Sources**

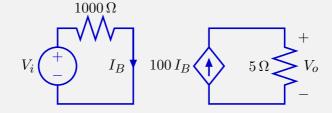
To analyze op-amps, we must introduce a new kind of element: a dependent source.

A dependent source generates a voltage or current whose value depends on another voltage or current.

Example: current-controlled current source

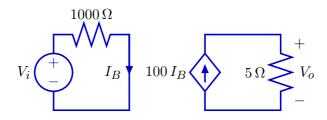


Find  $\frac{V_o}{V_i}$ .



- 1. 500
- 2.  $\frac{1}{20}$
- 3. 1
  - 4.  $\frac{1}{2}$
- 5. none of the above

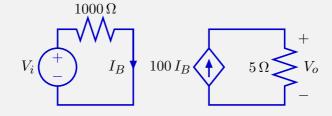
Find  $\frac{V_o}{V_i}$ .



$$I_B = \frac{V_i}{1000\,\Omega}$$

$$V_o = 100 I_B \times 5 \Omega = 100 \frac{V_i}{1000 \Omega} \times 5 \Omega = \frac{1}{2} V_i$$

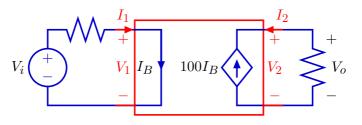
Find  $\frac{V_o}{V}$ .



- 1. 500
- 2.  $\frac{1}{20}$
- 3. 1
  - 4.  $\frac{1}{2}$
- 5. none of the above

## **Dependent Sources**

Dependent sources are two-ports: characterized by two equations.

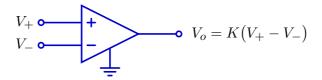


Here  $V_1 = 0$  and  $I_2 = -100 I_1$ .

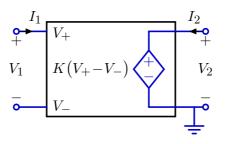
By contrast, one-ports (resistors, voltage sources, current sources) are characterized by a single equation.

#### **Op-Amp**

An op-amp (operational amplifier) can be represented by a voltagecontrolled voltage source.



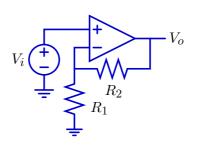
A voltage-controlled voltage source is a two-port.



 $I_1=0$  and  $V_2=KV_1$  where K is large (typically  $K>10^5$ ).

## **Op-Amp: Analysis**

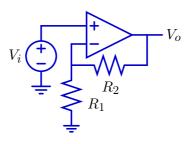
Example. Find  $\frac{V_o}{V_i}$  for the following circuit.



$$\begin{split} V_{+} &= V_{i} \\ V_{-} &= \frac{R_{1}}{R_{1} + R_{2}} \, V_{o} \\ V_{o} &= K \big( V_{+} - V_{-} \big) = K \big( V_{i} - \frac{R_{1}}{R_{1} + R_{2}} \, V_{o} \big) \\ &\frac{V_{o}}{V_{i}} = \frac{K}{1 + \frac{KR_{1}}{R_{1} + R_{2}}} = \frac{K \big( R_{1} + R_{2} \big)}{R_{1} + R_{2} + KR_{1}} \, \approx \, \frac{R_{1} + R_{2}}{R_{1}} \quad \text{(if $K$ is large)} \end{split}$$

## Non-inverting Amplifier

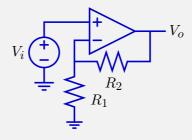
For large K, this circuit implements a non-inverting amplifier.



$$\frac{V_o}{V_i} = \frac{R_1 + R_2}{R_1} \ge 1$$

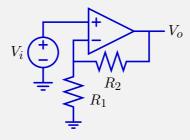
$$V_o \ge V_i$$

For which value(s) of  $R_1$  and/or  $R_2$  is  $V_o = V_i$ .



- 1.  $R_1 \to \infty$
- 2.  $R_2 = 0$
- 3.  $R_1 \rightarrow \infty$  and  $R_2 = 0$
- 4. all of the above
- 5. none of the above

For which value(s) of  $R_1$  and/or  $R_2$  is  $V_0 = V_i$ .



- 1.  $R_1 \to \infty$
- 2.  $R_2 = 0$
- 3.  $R_1 \rightarrow \infty$  and  $R_2 = 0$
- 5. none of the above

4. all of the above all are unity buffers

## The "Ideal" Op-Amp

As  $K \to \infty$ , the difference between  $V_+$  and  $V_-$  goes to zero.

Example:

$$V_i$$

$$V_{o} = K (V_{+} - V_{-}) = K (V_{i} - V_{o})$$

$$V_{o} = \frac{K}{1 + K} V_{i}$$

$$V_{+} - V_{-} = V_{i} - V_{o} = V_{i} - \frac{K}{1 + K} V_{i} = \frac{1}{1 + K} V_{i} = \frac{1}{K} V_{o}$$

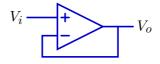
$$\lim_{K \to \infty} (V_{+} - V_{-}) = 0$$

If the difference between V+ and  $V_-$  did not go to zero as  $K\to\infty$  then  $V_o=K$   $(V_+-V_-)$  could not be finite.

## The "Ideal" Op-Amp

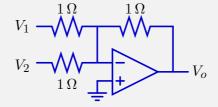
The approximation that  $V_+=V_-$  is referred to as the "ideal" op-amp approximation. It greatly simplifies analysis.

Example.



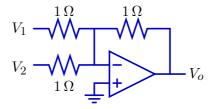
If  $V_+ = V_-$  then  $V_o = V_i$ !

Determine the output of the following circuit.



- 1.  $V_o = V_1 + V_2$
- 2.  $V_0 = V_1 V_2$
- 3.  $V_o = -V_1 V_2$
- 4.  $V_o = -V_1 + V_2$
- 5 none of the above

Determine the output of the following circuit.



Ideal op-amp approximation:

$$V_- = V_+ = 0$$

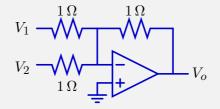
KCL at  $V_-$ :

$$\frac{V_1 - 0}{1} + \frac{V_2 - 0}{1} + \frac{V_o - 0}{1} = 0$$

Solving:

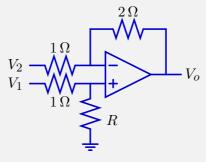
$$V_0 = -V_1 - V_2$$

Determine the output of the following circuit. 3



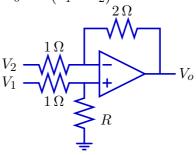
- 1.  $V_o = V_1 + V_2$
- 2.  $V_0 = V_1 V_2$
- 3.  $V_o = -V_1 V_2$  an inverting summer
- 4.  $V_o = -V_1 + V_2$
- 5 none of the above

Determine R so that  $V_o = 2(V_1 - V_2)$ .



- 1. R = 0
- 2. R = 1
- 3. R = 2
- 4.  $R \rightarrow \infty$
- 5. none of the above

Determine R so that  $V_o = 2(V_1 - V_2)$ .



No current in positive or negative inputs:

$$V_{+} = \frac{R}{1+R} V_{1}$$

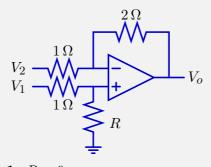
$$V_{-} = V_{2} + \frac{1}{1+2} (V_{o} - V_{2}) = \frac{2}{3} V_{2} + \frac{1}{3} V_{o}$$

Ideal op-amp:

$$V_{+} = V_{-} = \frac{R}{1+R} V_{1} = \frac{2}{3} V_{2} + \frac{1}{3} V_{o}$$

$$V_o = \frac{3R}{1+R} V_1 - 2V_2$$
  $\rightarrow$   $\frac{3R}{1+R} = 2$   $\rightarrow$   $R = 2\Omega$ 

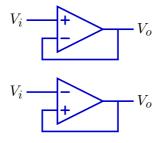
Determine R so that  $V_o = 2(V_1 - V_2)$ . 3



- 1. R = 0
- 2. R = 1
- 3. R = 2
- 4.  $R \rightarrow \infty$
- 5. none of the above

## The "Ideal" Op-Amp

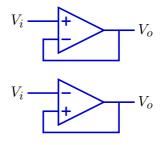
The ideal op-amp approximation implies that both of these circuits function identically.



$$V_+ = V_- \rightarrow V_o = V_i !$$

## The "Ideal" Op-Amp

The ideal op-amp approximation implies that both of these circuits function identically.

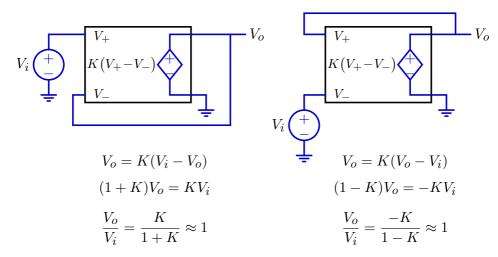


$$V_{+} = V_{-} \rightarrow V_{o} = V_{i}$$
!

This sounds a bit implausible!

#### **Paradox**

Try analyzing the voltage-controlled voltage source model.



These circuits seem to have identical responses if K is large.

Something is wrong!

# "Thinking" like an op-amp

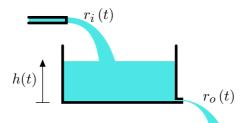
This reasoning is wrong because it ignores a critical property of circuits.

For a voltage to change, charged particles must flow.

To understand flow, we need to understand continuity.

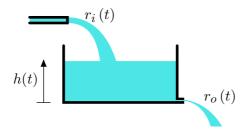
## Flows and Continuity

If a quantity is conserved, then the difference between what comes in and what goes out must accumulate.



## Flows and Continuity

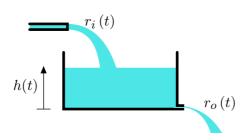
If a quantity is conserved, then the difference between what comes in and what goes out must accumulate.



If water is conserved then  $\frac{dh(t)}{dt} \propto r_i(t) - r_o(t)$ .

# **Leaky Tanks and Capacitors**

Water accumulates in a leaky tank.



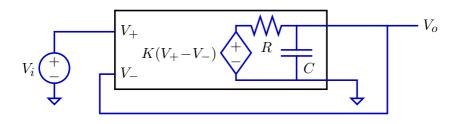
Charge accumulates in a capacitor.

$$\begin{array}{c|c}
 & i_0 \\
 & + \\
 & v \\
 & - \\
\end{array}$$

$$rac{dv}{dt} = rac{i_i - i_o}{C} \propto i_i - i_o$$
 analogous to  $rac{dh}{dt} \propto r_i - r_i$ 

## Charge Accumulation in an Op-Amp

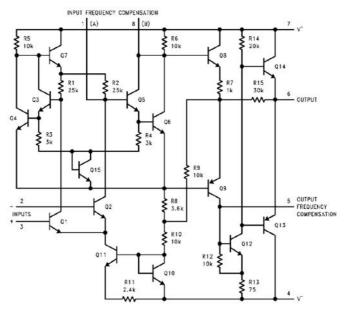
We can add a resistor and capacitor to "model" the accumulation of charge in an op-amp.



This is not an accurate representation of what is inside an op-amp.

### **Op-Amp Model**

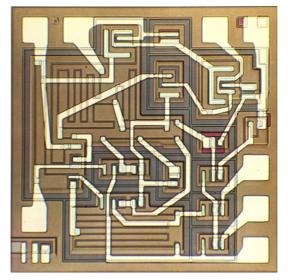
Here is a more accurate circuit model of a  $\mu$ A709 op-amp.



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### **Op-Amp**

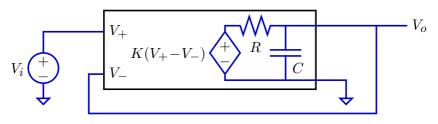
This artwork shows the physical structure of a  $\mu$ A709 op-amp.



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## Charge Accumulation in an Op-Amp

We can add a resistor and capacitor to "model" the accumulation of charge in an op-amp.



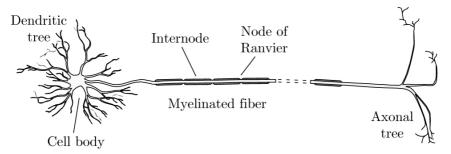
This is not an accurate representation of what is inside an op-amp.

This is a **model** of how the op-amp works.

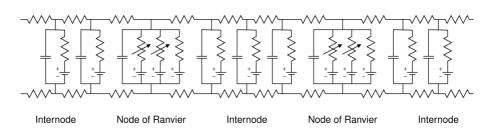
This is an example of using circuits as a tool for modeling.

#### Circuits as Models

Circuits as models of complex systems: myelinated neuron.



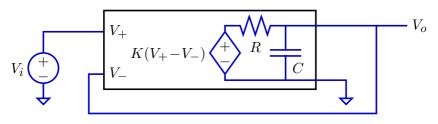
Model of myelinated nerve fiber



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## Charge Accumulation in an Op-Amp

We can add a resistor and capacitor to "model" the accumulation of charge in an op-amp.

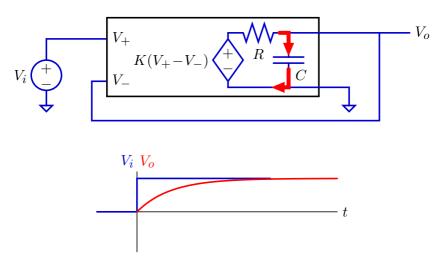


This is not an accurate representation of what is inside an op-amp.

This is a **model** of how the op-amp works.

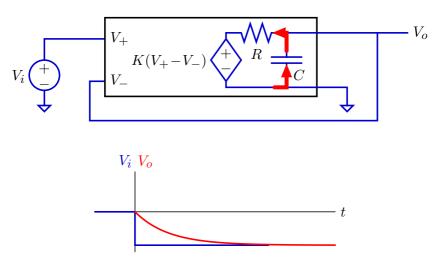
This is an example of using circuits as a tool for modeling.

If the input voltage to this circuit suddenly increases, then current will flow into the capacitor and gradually increase  $V_o$ .



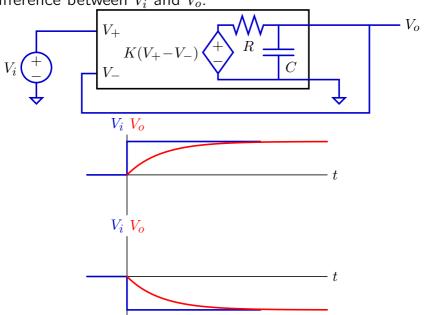
As  $V_o$  increases, the difference  $V_+ - V_-$  decreases, less current flows, and  $V_o$  approaches a final value equal to  $V_i$ .

If the input voltage to this circuit suddenly decreases, then current will flow out of the capacitor and decrease  $V_o$ .

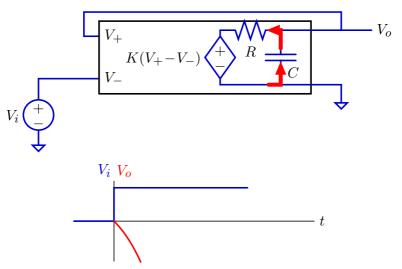


As  $V_o$  decreases, the  $|V_+-V_-|$  decreases, the magnitude of the current decreases, and  $V_o$  approaches a final value equal to  $V_i$ .

Regardless of how  $V_i$  changes,  $V_o$  changes in a direction to reduce the difference between  $V_i$  and  $V_o$ .

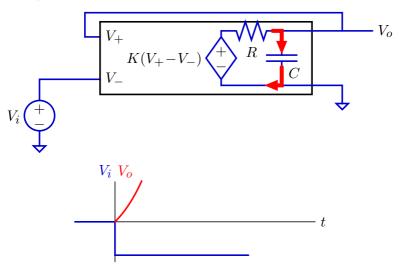


Switching the plus and minus inputs flips these relations. Now if the input increases, current will flow out of the capacitor and decrease  $V_o$ .



This makes the difference between input and output even bigger!

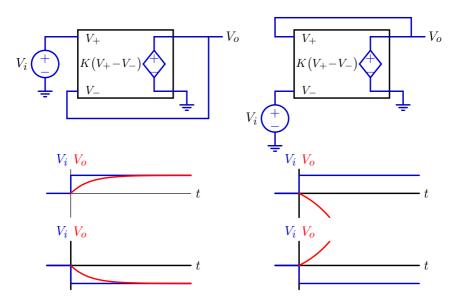
Similarly, if the input decreases, current will flow into the capacitor a increase  $V_o$ .



As the output diverges from the input, the magnitude of the capacitor current increases, and the rate of divergence increases!

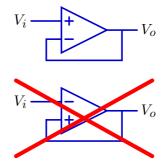
## **Positive and Negative Feedback**

Negative feedback (left) drives the output **toward** the input. Positive feedback (right) drives the output **away from** the input.



### **Paradox Resolved**

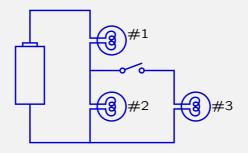
Although both circuits have solutions with  $V_o = V_i$  (large K), only the first is stable to changes in  $V_i$ .



Feedback to the positive input of an op-amp is unstable.

Use negative feedback to get a stable result.

What happens if we add third light bulb?



Closing the switch will make

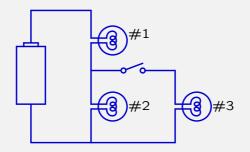
1. bulb 1 brighter

2. bulb 2 dimmer

3. 1 and 2

- 4. bulbs 1, 2, & 3 equally bright
- 5. none of the above

What happens if we add third light bulb? 3



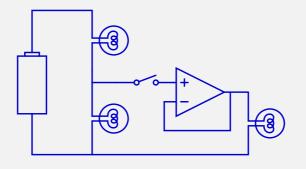
Closing the switch will make

- 1. bulb 1 brighter 2. bulb 2 dimmer

3. 1 and 2

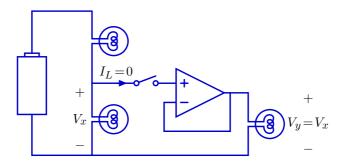
- 4. bulbs 1, 2, & 3 equally bright
- 5. none of the above

What will happen when the switch is closed?



- 1. top bulb is brightest 2. right bulb is brightest
- right bulb is dimmest
   all 3 bulbs equally bright
  - 5. none of the above

What will happen when the switch is closed?



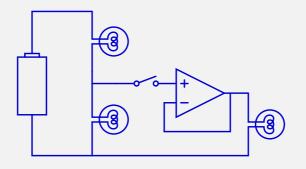
Closing the switch will have no effect on the left bulbs because no current will flow ( $I_L=0$ ) when switch is open OR closed.

 $\rightarrow$  This is half of the buffer idea: no input current!

When the switch is closed,  $V_y = V_x$ .

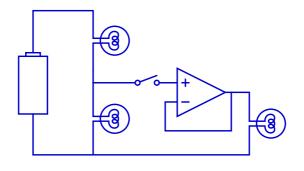
→ This is the other half: output voltage = input voltage!

What will happen when the switch is closed? 4



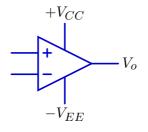
- 1. top bulb is brightest 2. right bulb is brightest
- 3. right bulb is dimmest 4. all 3 bulbs equally bright
  - 5. none of the above

The battery provides the power to illuminate the left bulbs. Where does the power come from to illuminate the right bulb?



### **Power Rails**

Op-amps derive power from connections to a power supply.



Typically, the output voltage of an op-amp is constrained by the power supply:

$$-V_{EE} < V_o < V_{CC}$$
.

### **Summary**

An op-amp can be represented as a voltage-dependent voltage source.

The "ideal" op-amp approximation is  $V_+ = V_-$ .

The ideal op-amp approximation only makes sense when the op-amp is connected with negative feedback.



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