

Lecture 23

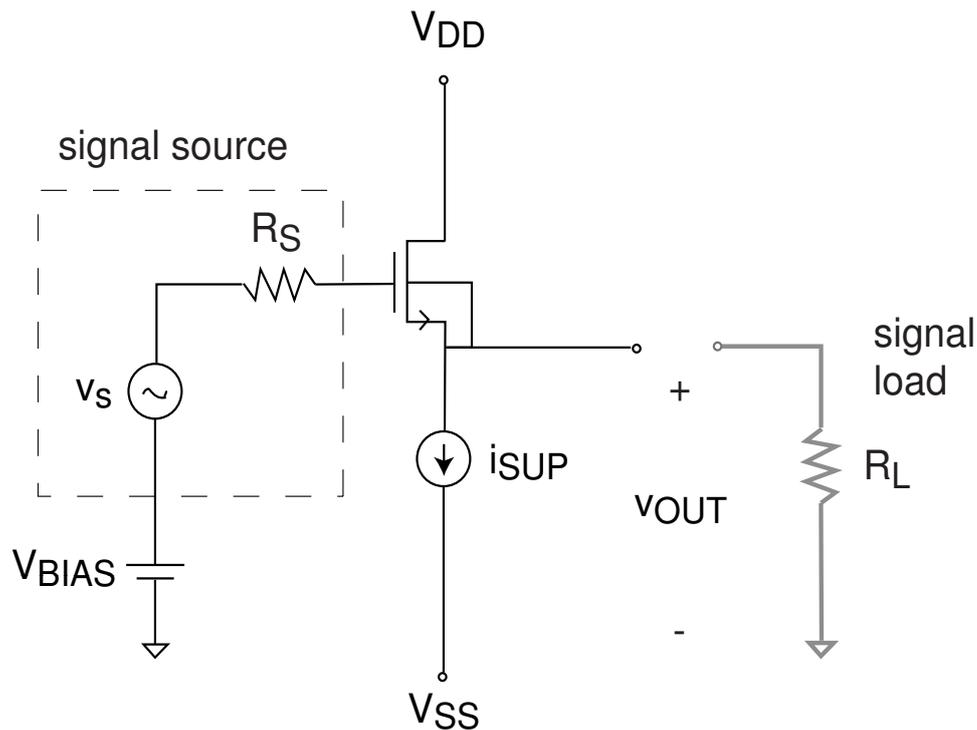
Frequency Response of Amplifiers (III)

OTHER AMPLIFIER STAGES

Outline

1. Frequency Response of the Common-Drain Amplifier
2. Frequency Response of the Common-Gate Amplifier

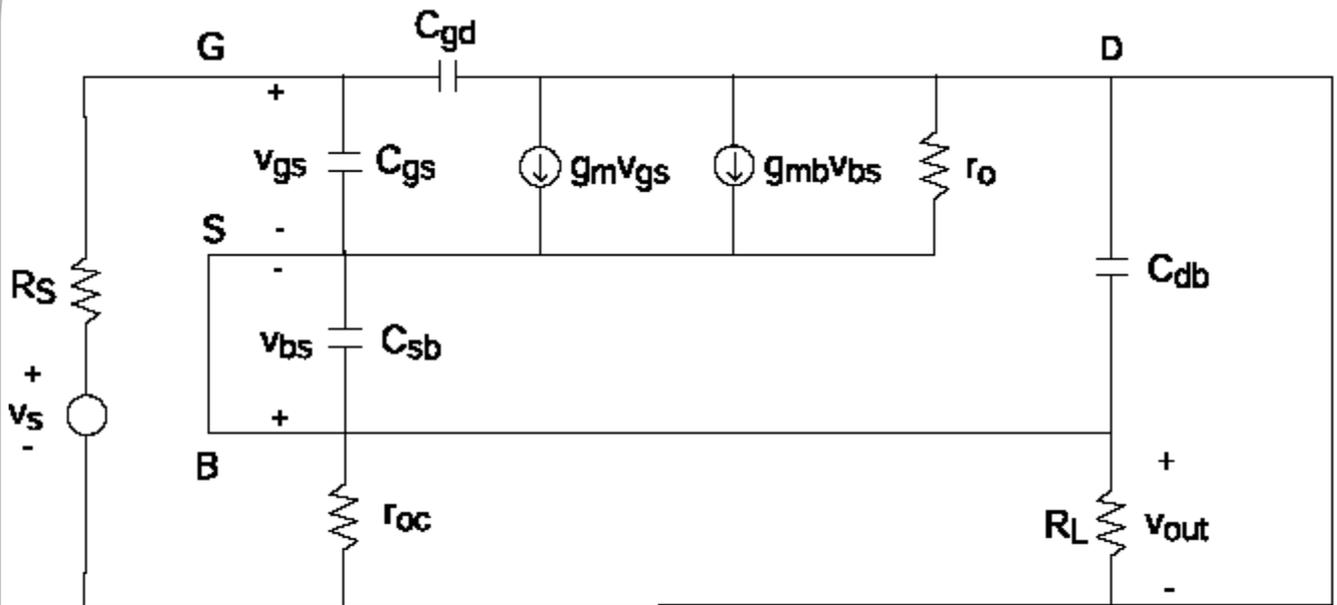
1. Frequency Response of the Common-Drain Amplifier



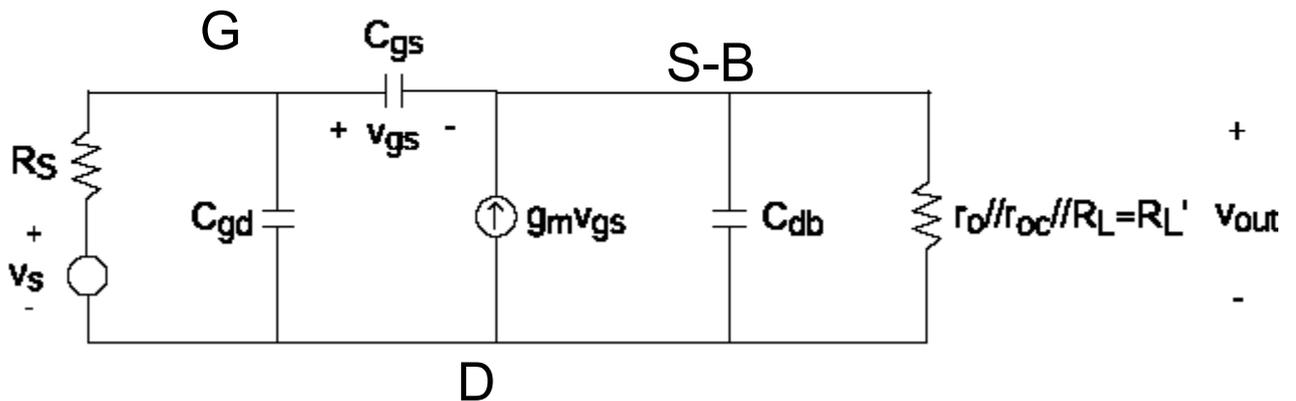
Characteristics of CD Amplifier:

- Voltage gain ≈ 1
- High input resistance
- Low output resistance
- \Rightarrow Good voltage buffer

High-frequency small-signal model



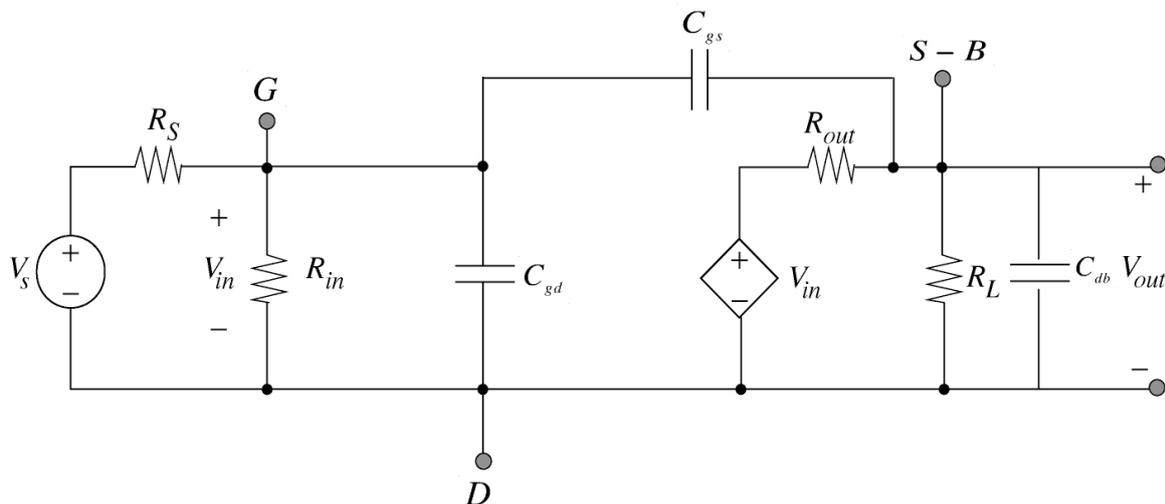
$v_{bs}=0$



Could use OTC to solve for bandwidth.

To estimate bandwidth it is easier to use the 2-port models.

Low Frequency Analysis Using 2-Port Model



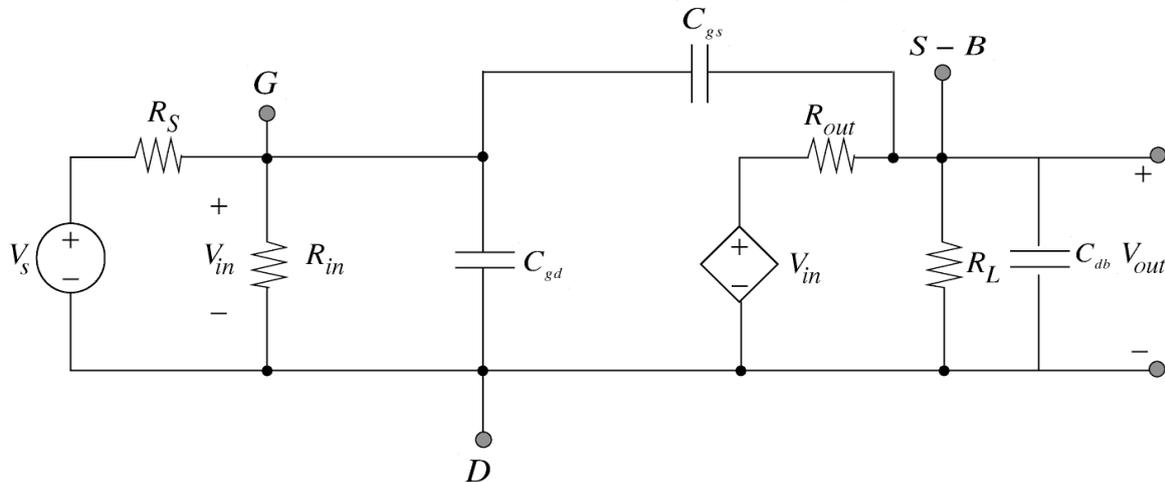
Capacitors -open circuit

$$A_{vo} = \left(\frac{R_{in}}{R_S + R_{in}} \right) (1) \left(\frac{R_L}{R_L + R_{out}} \right)$$

$$\approx \frac{g_m R_L}{1 + g_m R_L} \leq 1$$

In the calculation of the intrinsic voltage gain we assume that $r_o \parallel r_{oc}$ was large. That is why we do not have $R_L \parallel r_o \parallel r_{oc}$

High Frequency Analysis Using 2-Port Model - Add capacitors



Use Miller Approximation

$$A_v C_{gs} = \frac{R_L}{R_{out} + R_L} = \frac{g_m R_L}{1 + g_m R_L}$$

$$C_M = C_{gs} \left(1 - A_v C_{gs} \right) = C_{gs} \left(1 - \frac{g_m R_L}{1 + g_m R_L} \right) = \frac{C_{gs}}{1 + g_m R_L}$$

Total capacitance at input

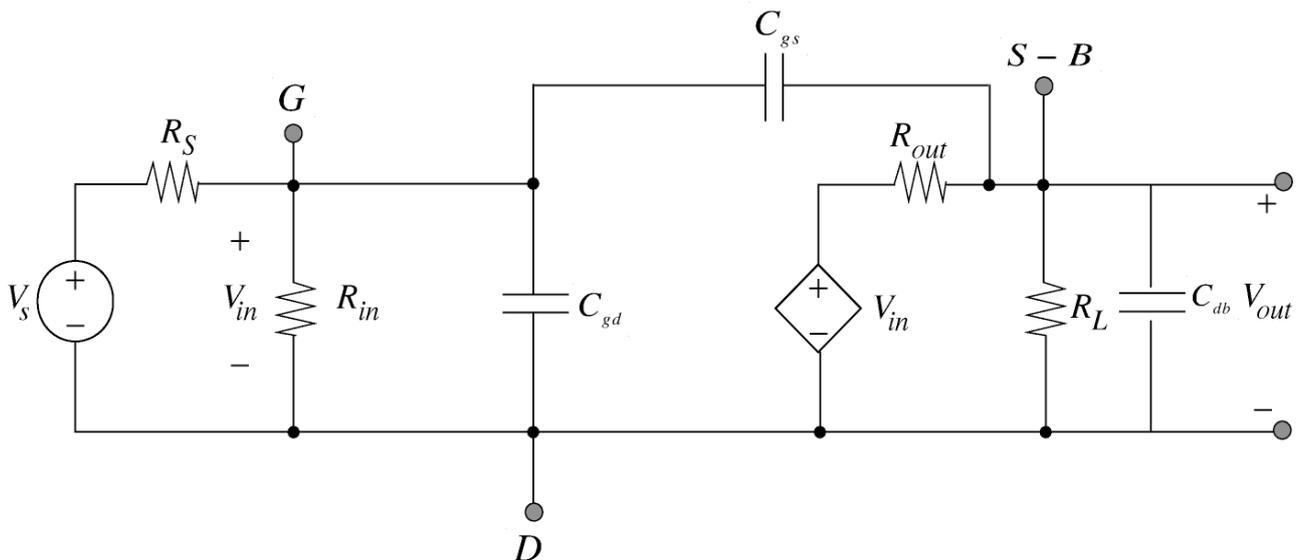
$$C_T = C_{gs} \left(1 - \frac{g_m R_L}{1 + g_m R_L} \right) + C_{gd}$$

$$R_T = R_S \parallel R_{in} = R_S$$

Add time constant due to C_{db} capacitance at output

$$R_{C_{db}} = R_{out} \parallel R_L = \frac{R_{out} R_L}{R_{out} + R_L} = \frac{R_L}{1 + g_m R_L}$$

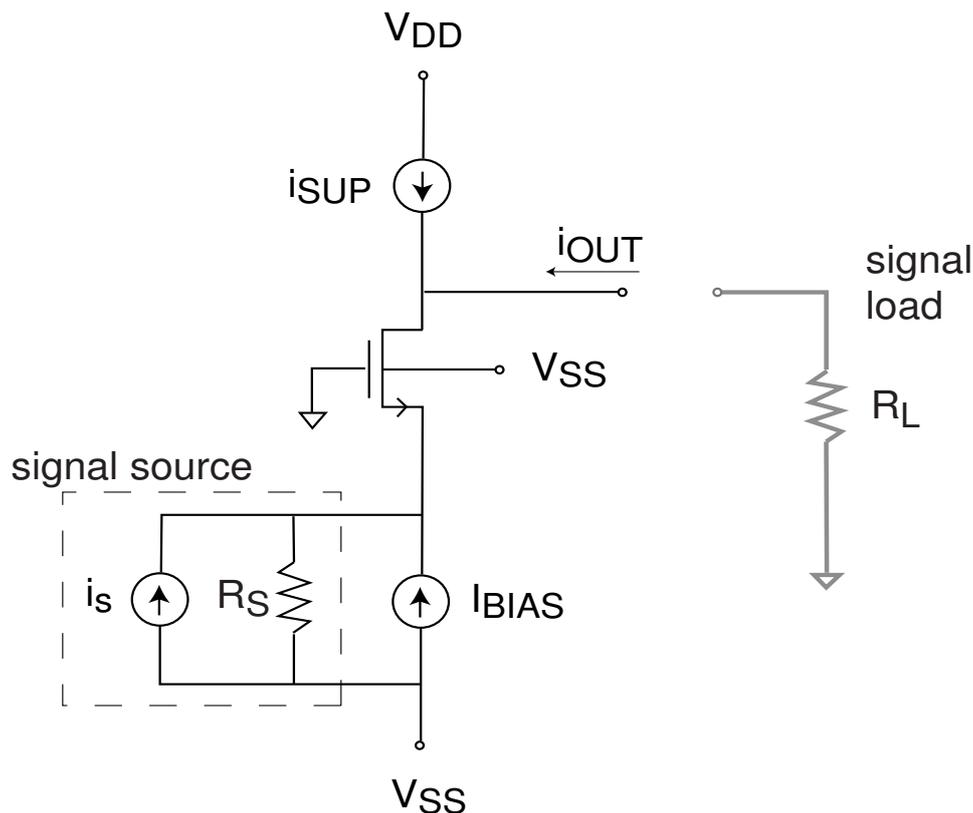
Frequency Response of Common-Drain Amplifier



$$\omega_{3dB} \approx \frac{1}{R_S \left(\frac{C_{gs}}{1 + g_m R_L} + C_{gd} \right) + C_{db} \frac{R_L}{1 + g_m R_L}}$$

If R_S is not too high, bandwidth can be rather high and approach ω_T .

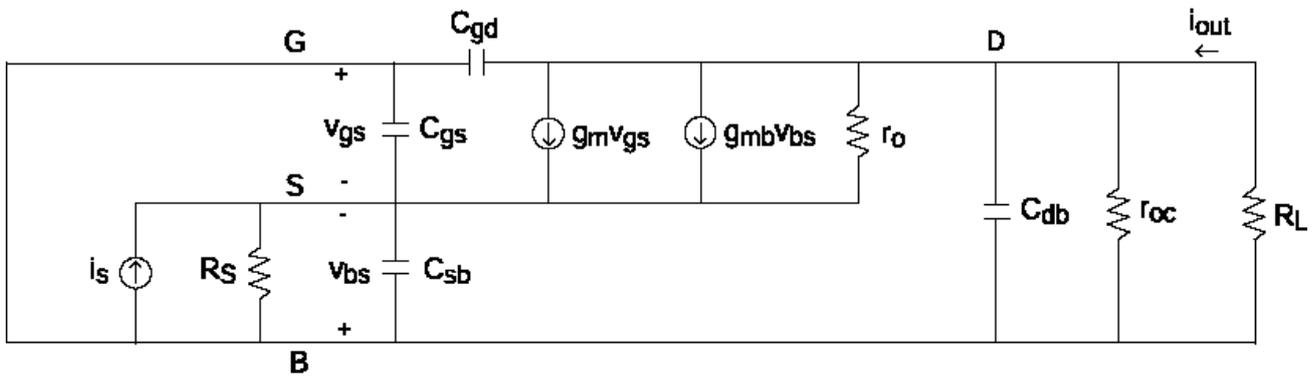
2. Frequency Response of the Common-Gate Amplifier



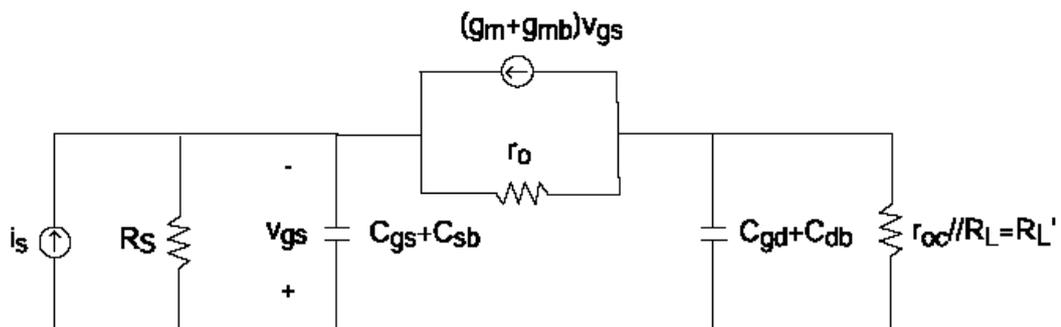
Characteristics of CG Amplifier:

- Current gain ≈ 1
- Low input resistance
- High output resistance
- \Rightarrow Good current buffer

High Frequency Small Signal Model



↓ $v_{gs} = v_{bs}$

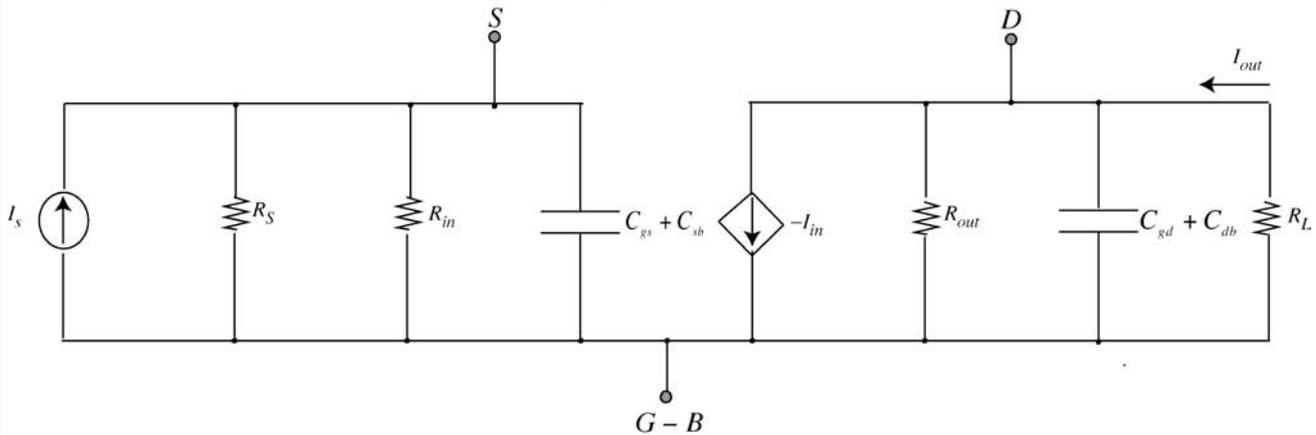


Could use OTC to solve for bandwidth.

To estimate bandwidth it is easier to use the 2-port models.

High-frequency small-signal 2-port model

Assume backgate is shorted to source



Low frequency transfer function:

$$A_{io} = \frac{i_{out}}{i_s} = \left(\frac{R_S}{R_{in} + R_S} \right) (1) \left(\frac{R_{out}}{R_L + R_{out}} \right)$$

Use OTC to find ω_{3dB} :

Thevenin resistance across $C_{gs} +$

C_{sb}

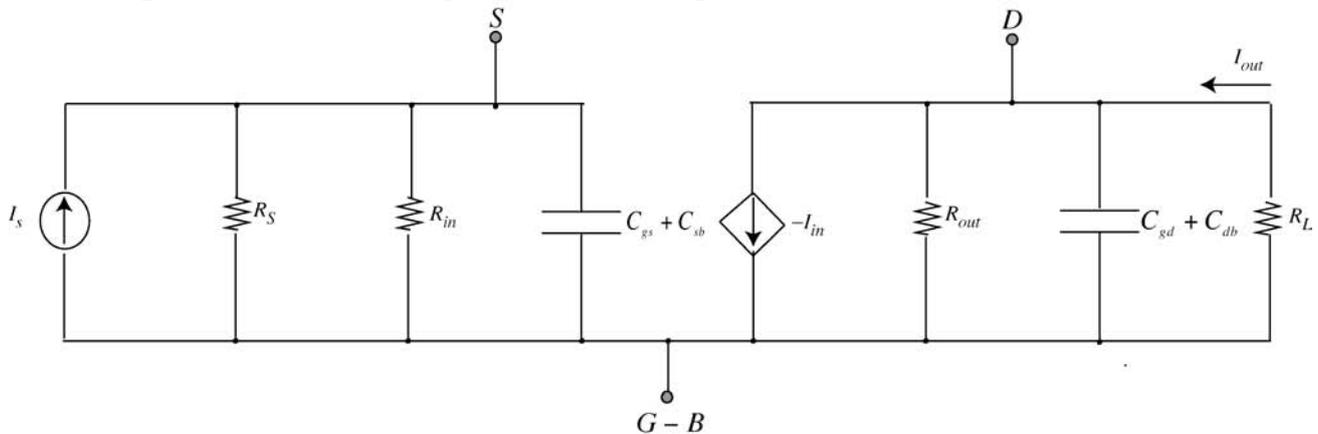
$$R_{TC_{gs}} = R_S \parallel R_{in} = R_S \parallel \left(1 / (g_m + g_{mb}) \right)$$

Thevenin resistance across $C_{gd} +$

C_{db}

$$R_{TC_{gd}} = R_{out} \parallel R_L = ((r_o + g_m r_o R_S) \parallel r_{oc}) \parallel R_L$$

High-frequency small-signal 2-port model con't



Open circuit time constants:

$$\tau_{C_{gs}+C_{sb}} = (R_S \parallel (1/g_m + g_{mb}))(C_{gs} + C_{sb})$$

$$\tau_{C_{gd}+C_{db}} = (((r_o + g_m r_o R_S) \parallel r_{oc}) \parallel R_L)(C_{gd} + C_{db})$$

Summing the open circuit time constants:

$$\omega_{3dB} = 1 / \left[(R_S \parallel (1/g_m + g_{mb}))(C_{gs} + C_{sb}) + (((r_o + g_m r_o R_S) \parallel r_{oc}) \parallel R_L)(C_{gd} + C_{db}) \right]$$

If R_L is not too high, bandwidth can be rather high and approach ω_T .

What did we learn today?

Summary of Key Concepts

- Common-drain amplifier:
 - Voltage gain ≈ 1 , *Miller Effect* nearly completely eliminates the effect of C_{gs}
 - If R_S is not too high, CD amplifier has high bandwidth
- Common-gate amplifier
 - No Miller Effect because there is no feedback capacitor
 - If R_L is not too high, CG amplifier has high bandwidth
- R_S , R_L can affect bandwidth of amplifiers.

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