

# Lecture 19

## Transistor Amplifiers (I)

### Common-Source Amplifier

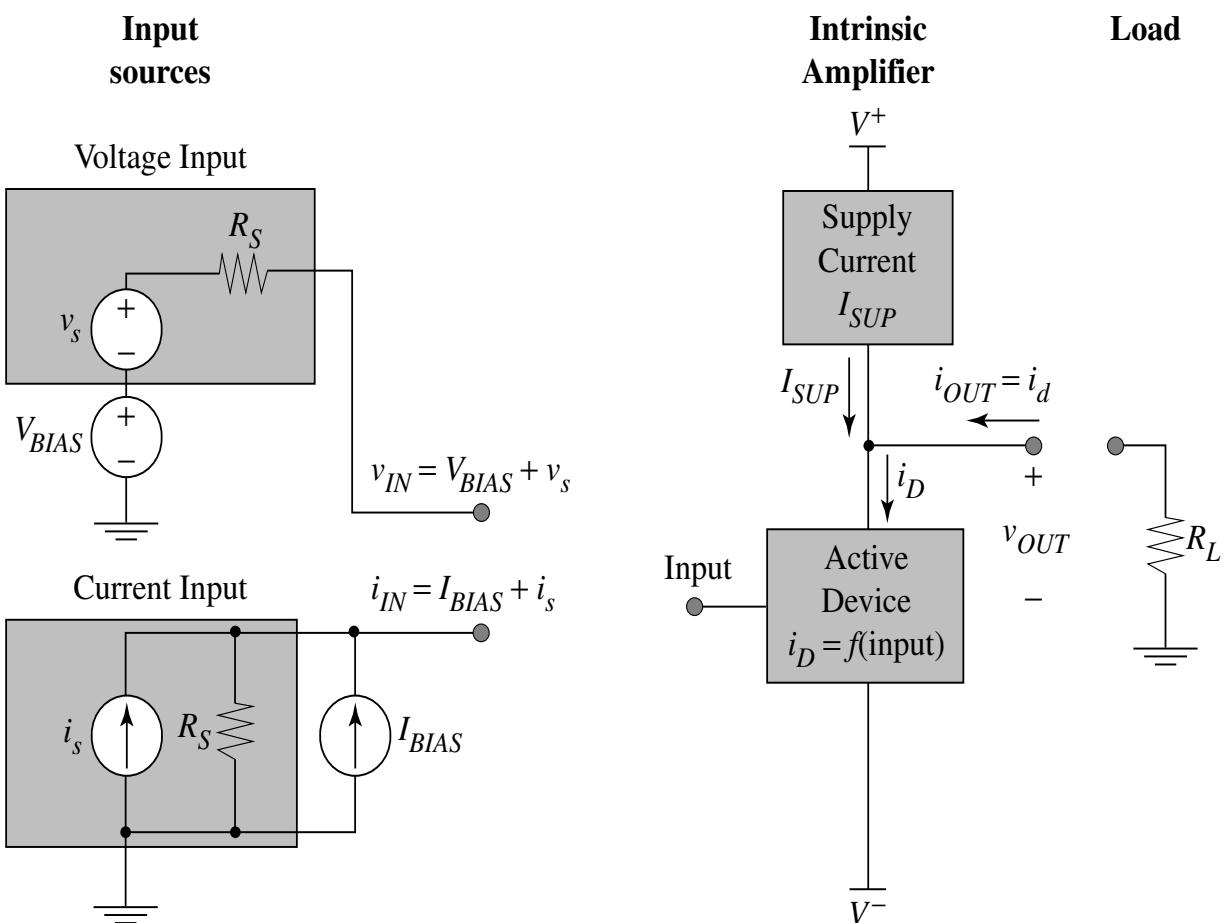
#### Outline

- Amplifier fundamentals
- Common-source amplifier
- Common-source amplifier with current-source supply

**Reading Assignment:**  
Howe and Sodini; Chapter 8, Sections 8.1-8.4

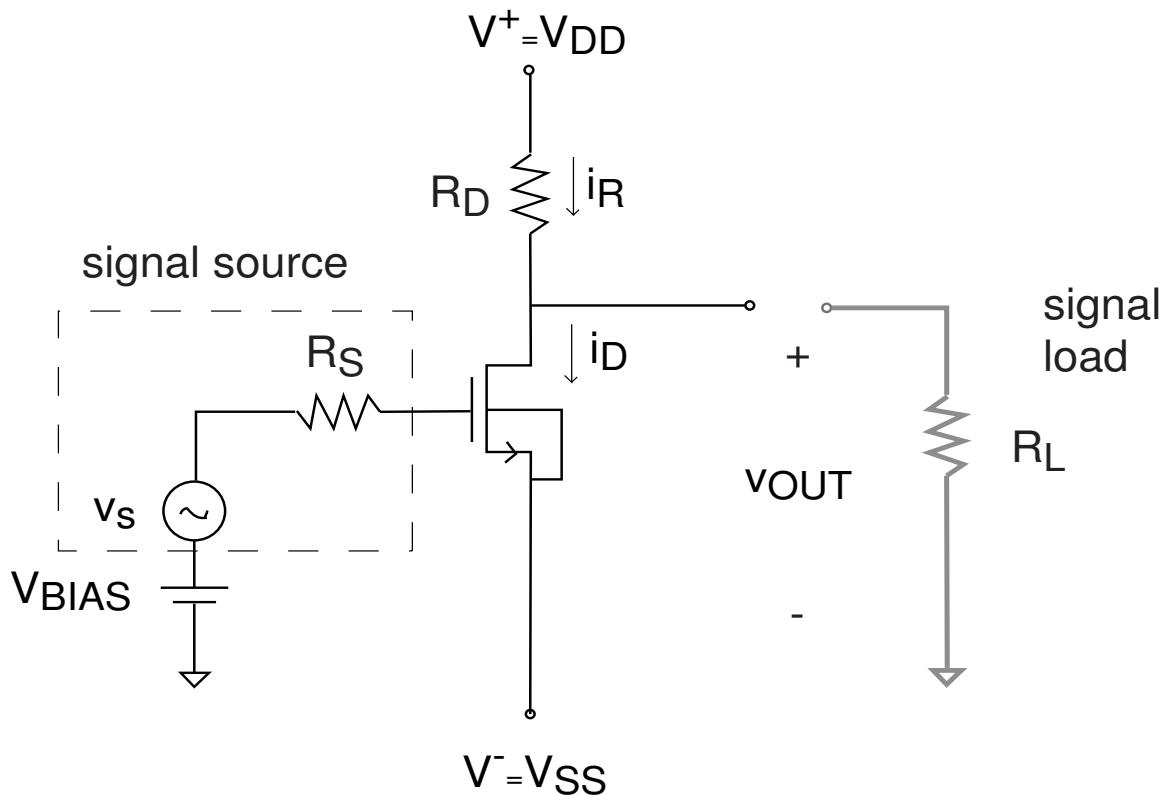
# Amplifier Fundamentals

- Source resistance  $R_S$  is associated *only* with small signal sources
- Choose  $I_D = I_{SUP}$
- DC output current
  - $I_{OUT} = 0$
  - $V_{OUT} = 0$



## 2. Common-Source Amplifier:

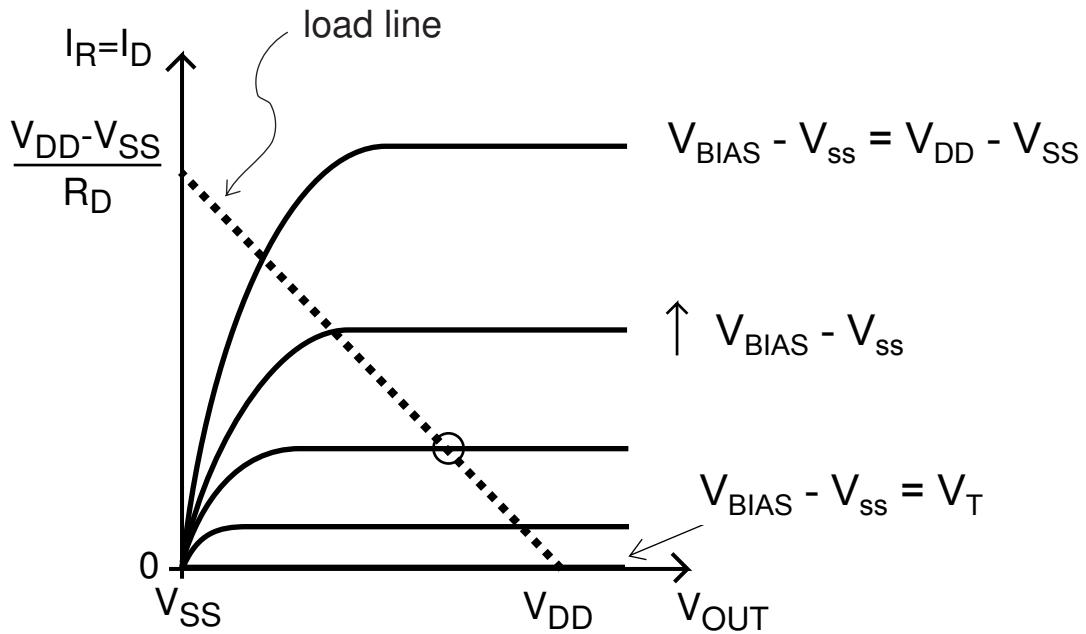
Consider the following circuit:



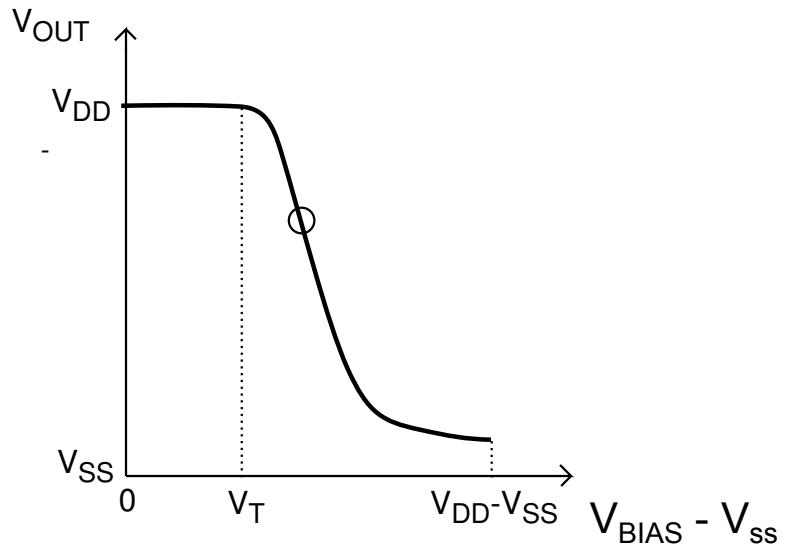
- Consider intrinsic voltage amplifier - no loading
  - $R_S = 0$
  - $R_L \rightarrow \infty$
  - $V_{GS} = V_{BIAS} - V_{SS}$
- $V_{BIAS}$ ,  $R_D$  and W/L of MOSFET selected to bias transistor in saturation and obtain desired output bias point (i.e.  $V_{OUT} = 0$ ).

Watch notation:  $v_{OUT}(t) = V_{OUT} + v_{out}(t)$

## Load line view of amplifier:



## Transfer characteristics of amplifier:



Want:

- Bias point calculation;
- Limits to signal swing
- Small-signal gain;
- Frequency response [in a few days]

**Bias point:** choice of  $V_{BIAS}$ , W/L, and  $R_D$  to keep transistor in saturation and to get proper quiescent  $V_{OUT}$ .

Assume MOSFET is in saturation:

$$I_D = \frac{W}{2L} \mu_n C_{ox} (V_{BIAS} - V_{SS} - V_T)^2$$

$$I_R = \frac{V_{DD} - V_{OUT}}{R_D}$$

If we select  $V_{OUT}=0$ :

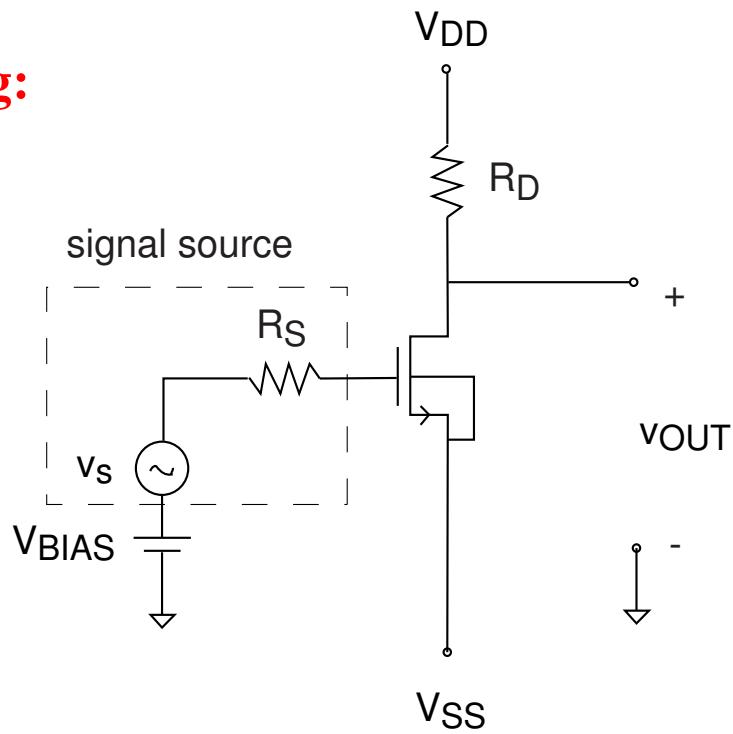
$$I_D = I_R = \frac{W}{2L} \mu_n C_{ox} (V_{BIAS} - V_{SS} - V_T)^2 = \frac{V_{DD}}{R_D}$$

Then:

$$V_{BIAS} = \sqrt{\frac{2I_D}{\frac{W}{L} \mu_n C_{ox}}} + V_{SS} + V_T$$

Equation that allows us to compute needed  $V_{BIAS}$  given  $R_D$  and W/L.

## Signal swing:



- Upswing: limited by MOSFET going into cut-off.

$$v_{out,max} = V_{DD}$$

- Downswing: limited by MOSFET leaving saturation.

$$V_{DS,sat} = V_{GS} - V_T = \sqrt{\frac{2I_D}{\frac{W}{L} \mu_n C_{ox}}}$$

or

$$v_{out,min} - V_{SS} = V_{BIAS} - V_{SS} - V_T$$

Then:

$$v_{out,min} = V_{BIAS} - V_T$$

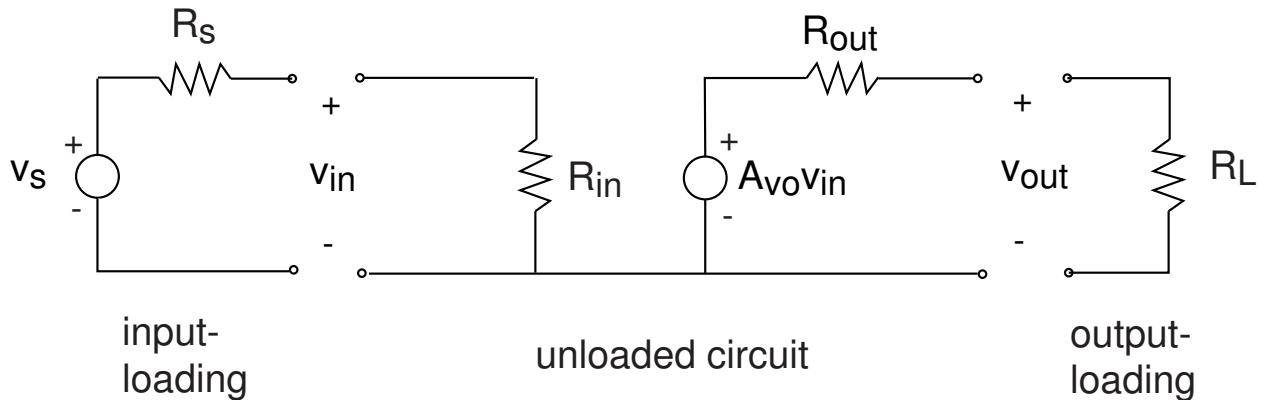
## Generic view of the effect of loading on small-signal operation

Two-port network view of small-signal equivalent circuit model of a voltage amplifier:

$R_{in}$  is *input resistance*

$R_{out}$  is *output resistance*

$A_{vo}$  is *unloaded voltage gain*



Voltage divider at input:

$$v_{in} = R_{in} \frac{v_s}{R_{in} + R_s}$$

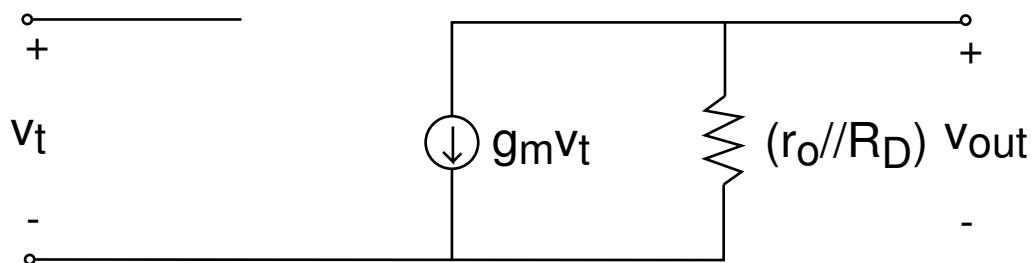
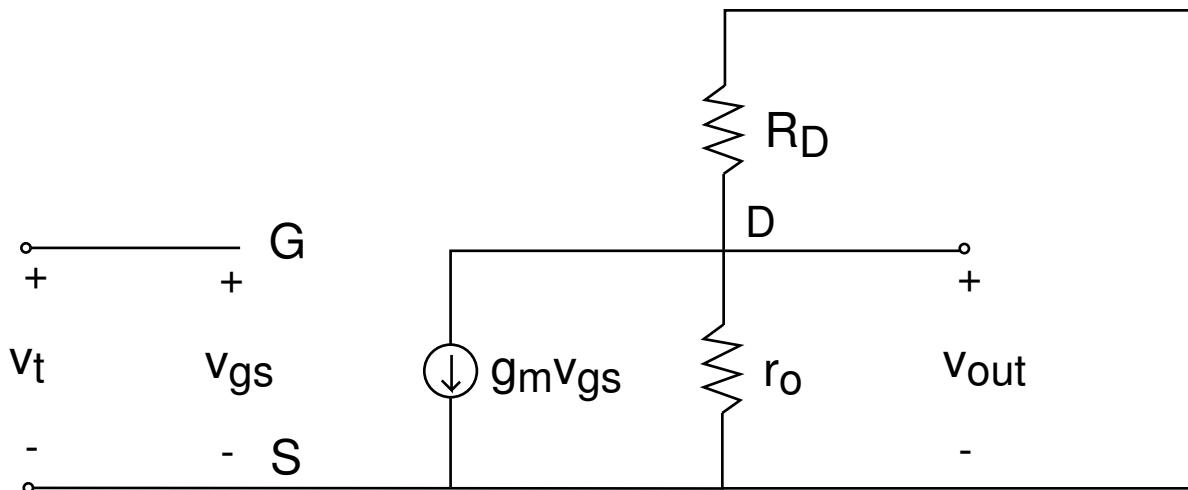
Voltage divider at output:

$$v_{out} = R_L \frac{A_{vo}v_{in}}{R_{out} + R_L}$$

Loaded voltage gain:

$$\frac{v_{out}}{v_s} = \frac{R_{in}}{R_{in} + R_s} A_{vo} \frac{R_L}{R_L + R_{out}}$$

**Small-signal voltage gain  $A_{vo}$ :** draw small-signal equivalent circuit model: Remove  $R_L$  and  $R_S$



$$v_{out} = -g_m v_t (r_o // R_D)$$

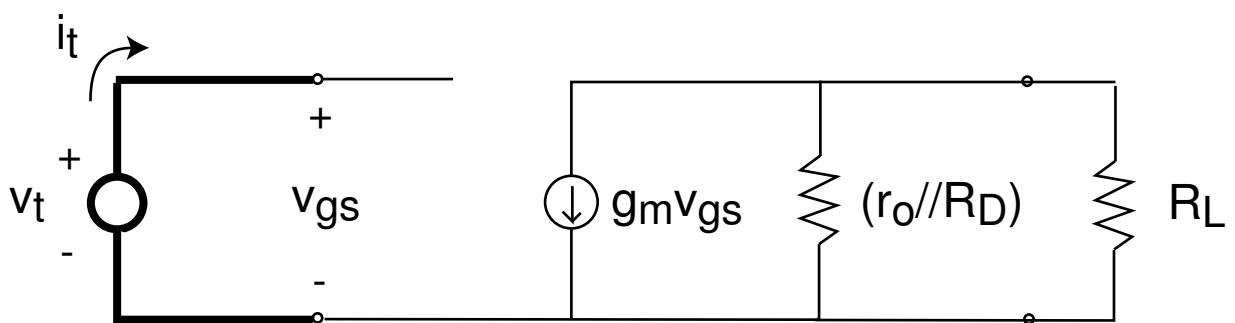
Then unloaded voltage gain:

$$A_{vo} = \frac{v_{out}}{v_t} = -g_m (r_o // R_D)$$

# Input Resistance

- Calculation of input resistance,  $R_{in}$ :
  - Load amplifier with  $R_L$
  - Apply test voltage (or current) at input, measure test current (or voltage).

For common-source amplifier:



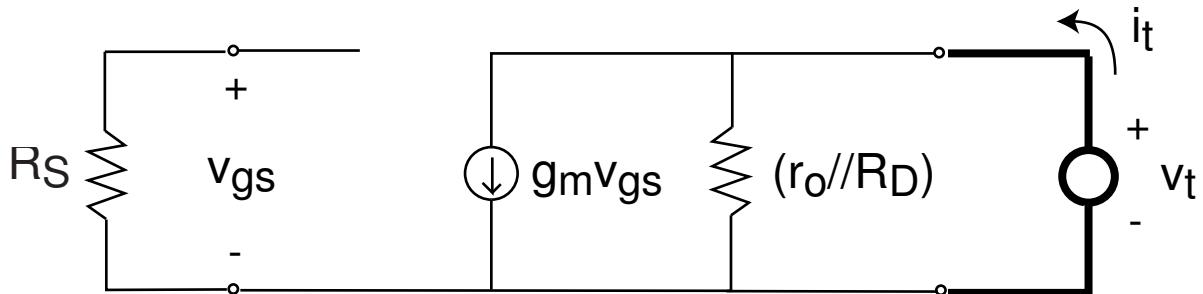
$$i_t = 0 \Rightarrow R_{in} = \frac{v_t}{i_t} = \infty$$

No effect of loading at input.

# Output Resistance

- Calculation of output resistance,  $R_{out}$ :
  - Load amplifier with  $R_S$
  - Apply test voltage (or current) at output, measure test current (or voltage).
  - Set input source equal zero

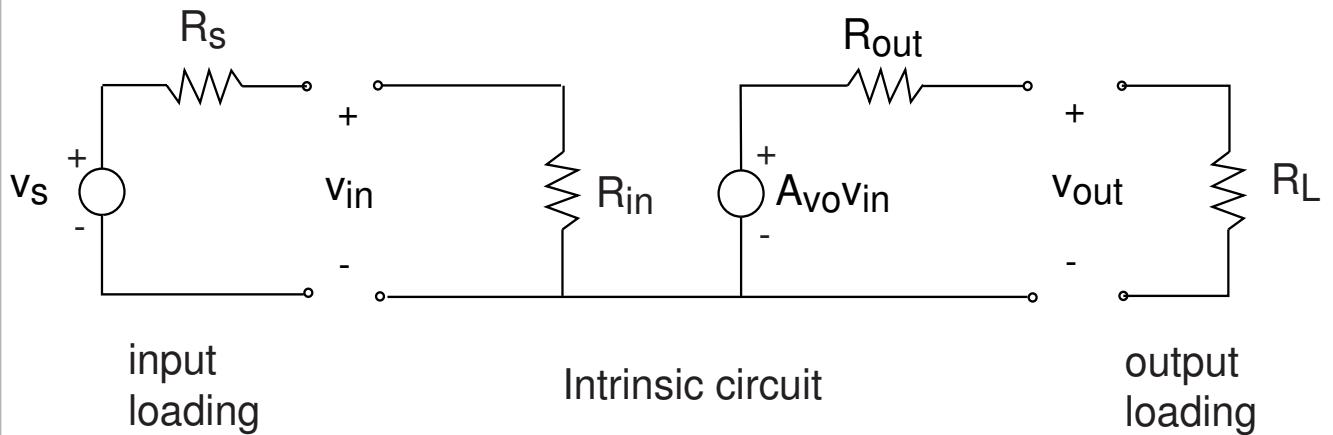
For common-source amplifier:



$$v_{gs} = 0 \Rightarrow g_m v_{gs} = 0 \Rightarrow v_t = i_t (r_o // R_D)$$

$$R_{out} = \frac{v_t}{i_t} = r_o // R_D$$

## Two-port network view of common-source amplifier Voltage Amplifier

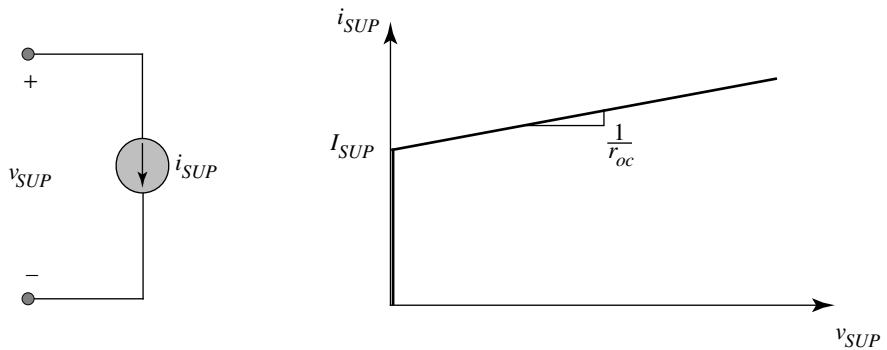


$$\frac{v_{out}}{v_s} = \frac{R_{in}}{R_{in} + R_S} A_{vo} \frac{R_L}{R_L + R_{out}}$$

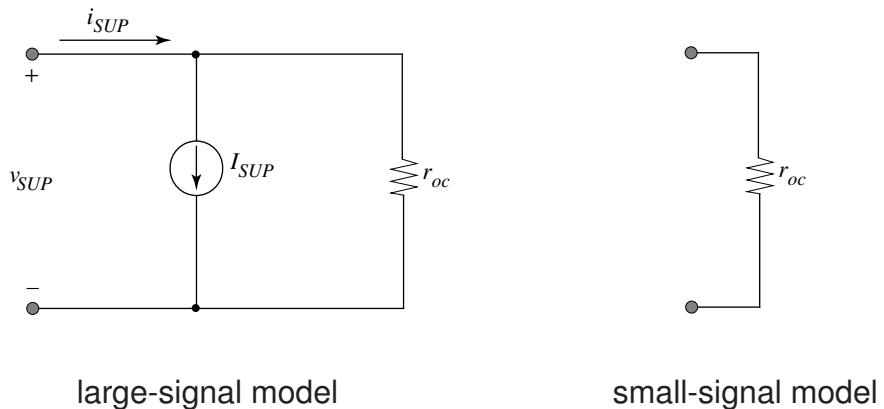
$$\frac{v_{out}}{v_s} = -g_m(r_o // R_D) \frac{R_L}{R_L + r_o // R_D} = -g_m(r_o // R_D // R_L)$$

# Current Source Supply

I—V characteristics of current source:

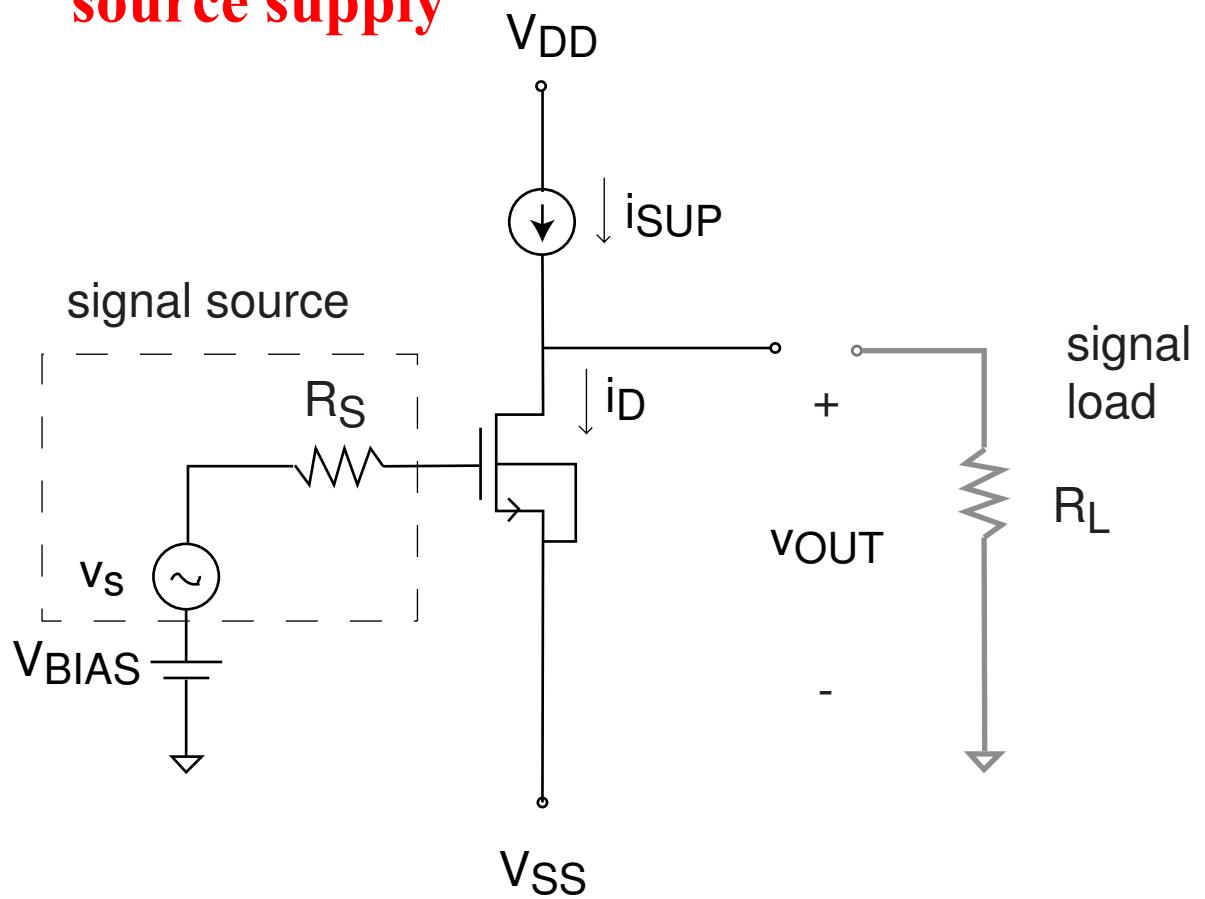


Equivalent circuit models :

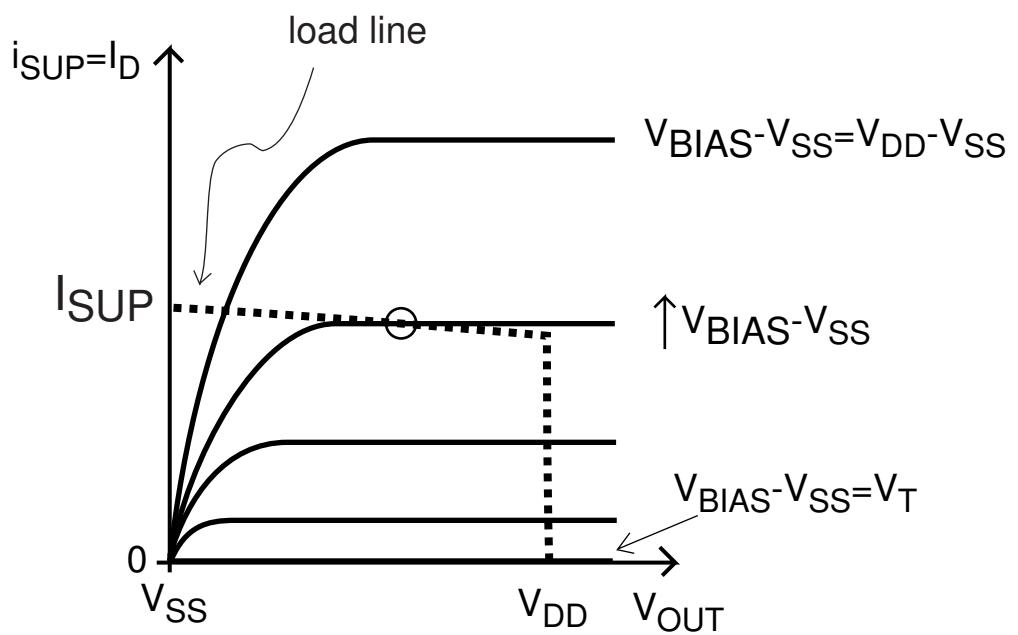


- $i_{SUP} = 0$  for  $v_{SUP} \leq 0$
- $i_{SUP} = I_{SUP} + v_{SUP}/r_{oc}$  for  $v_{SUP} > 0$
- High small-signal resistance  $r_{oc}$ .

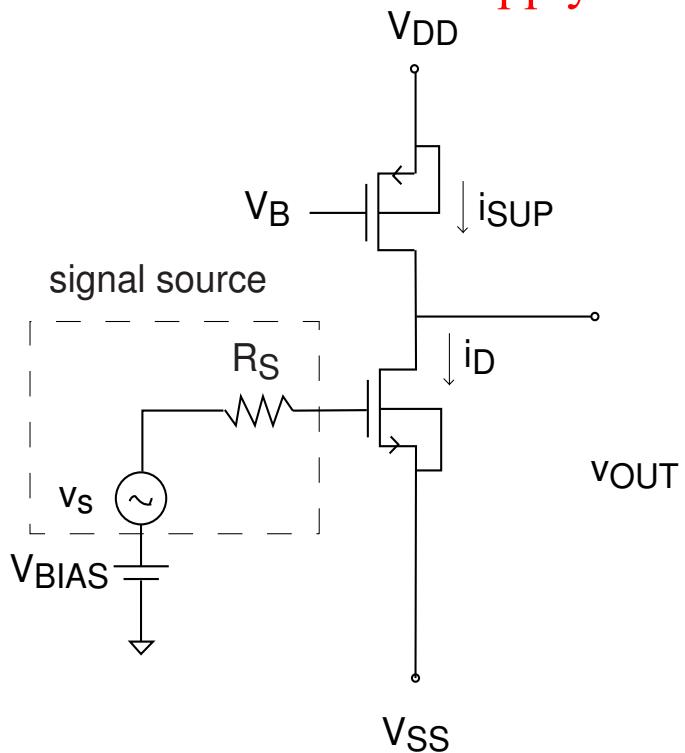
### 3. Common-source amplifier with current-source supply



Loadline View



## Use PMOS for current source supply



Bias point: Assume both transistors in saturation

$V_{OUT} = 0$ . Choose  $I_{SUP}$  and determine  $V_B$ .

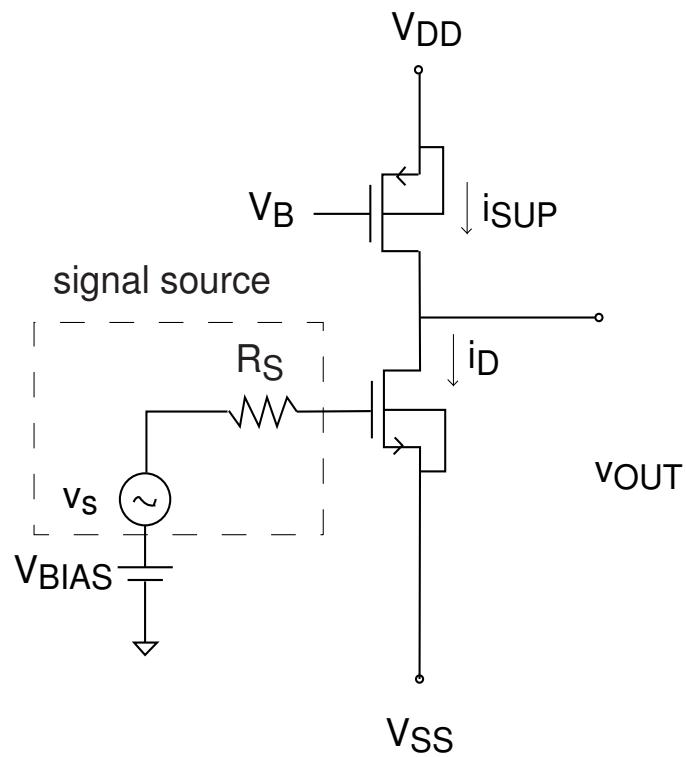
$$I_{SUP} = -I_{Dp} = \left( \frac{W}{2L} \right)_p \mu_p C_{ox} (V_{DD} - V_B + V_{Tp})^2$$

Set  $-I_{Dp} = I_{Dn}$  for  $V_{OUT} \sim 0$

$$I_{SUP} = I_{Dn} = \left( \frac{W}{2L} \right)_n \mu_n C_{ox} (V_{BIAS} - V_{SS} - V_{Tn})^2$$

$$V_{BIAS} = \sqrt{\frac{2I_{SUP}}{\left( \frac{W}{L} \right)_n \mu_n C_{ox}}} + V_{SS} + V_{Tn}$$

## Signal swing:



- Upswing: limited by PMOS leaving saturation.

$$V_{SD,sat} = V_{SG} + V_{Tp} = V_{DD} - V_B + V_{Tp}$$

$$V_{DD} - v_{out,max} = V_{DD} - V_B + V_{Tp}$$

$$v_{out,max} = V_B - V_{Tp}$$

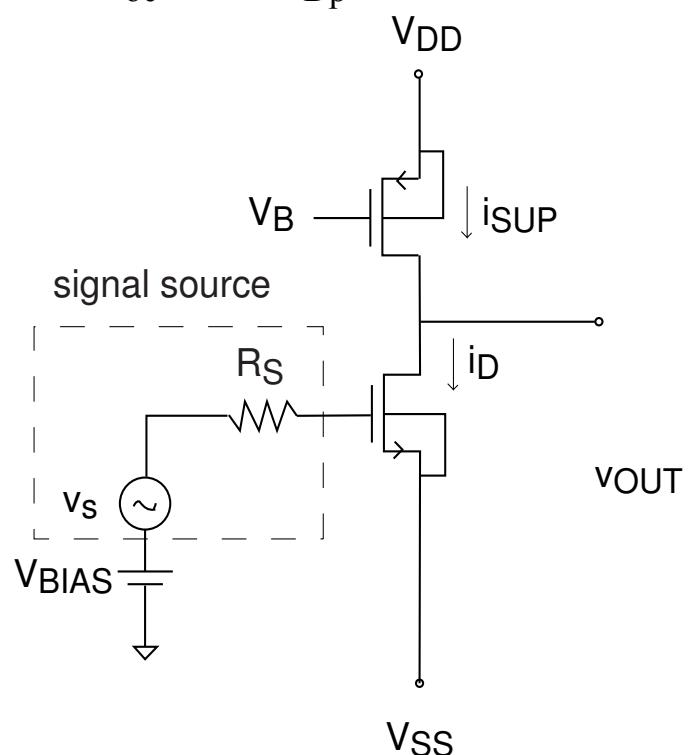
- Downswing: limited by NMOS leaving saturation.
- Same result as with resistive supply current.

$$v_{out,min} = V_{BIAS} - V_T$$

### 3. Common-source amplifier with current-source supply (contd.)

Current source characterized by high output resistance:  
 $r_{oc}$ . Significantly higher than amplifier with resistive supply.

p-channel MOSFET:  $r_{oc} = 1/\lambda I_{Dp}$



- Voltage gain:  $A_{vo} = -g_m (r_o // r_{oc})$ .
- Input resistance :  $R_{in} = \infty$
- Output resistance:  $R_{out} = r_o // r_{oc}$ .

## Relationship between circuit figures of merit and device parameters

Remember:

$$g_m = \sqrt{2I_D \frac{W}{L} \mu_n C_{ox}}$$

$$r_o \approx \frac{1}{\lambda_n I_D} \propto \frac{L}{I_D}$$

Then:

Device* Parameters	Circuit Parameters		
	$ A_{vo} $	$R_{in}$	$R_{out}$
	$g_m(r_o//r_{oc})$	$\propto$	$r_o//r_{oc}$
$I_{SUP} \uparrow$	$\downarrow$	-	$\downarrow$
$W \uparrow$	$\uparrow$	-	-
$L \uparrow$	$\uparrow$	-	$\uparrow$

\* adjustments are made to  $V_{BIAS}$  so that none of the other parameters change

CS amplifier with current source supply is a good voltage amplifier ( $R_{in}$  high and  $|A_{vo}|$  high), but  $R_{out}$  high too  $\Rightarrow$  voltage gain degraded if  $R_L \ll r_o//r_{oc}$ .

# What did we learn today?

## Summary of Key Concepts for CS amplifier

- Bias Calculations
- Signal Swing
- Small Signal Circuit Parameters
  - Voltage Gain -  $A_{VO}$
  - Input Resistance -  $R_{in}$
  - Output Resistance -  $R_{out}$
- Relationship between small signal circuit and device parameters

MIT OpenCourseWare  
<http://ocw.mit.edu>

6.012 Microelectronic Devices and Circuits  
Spring 2009

For information about citing these materials or our Terms of Use, visit: <http://ocw.mit.edu/terms>.