

## Lecture 10 - MOSFET (II)

### MOSFET I-V CHARACTERISTICS (*cont.*)

October 13, 2005

#### Contents:

1. The saturation regime
2. Backgate characteristics

#### Reading assignment:

Howe and Sodini, Ch. 4, §4.4

#### Announcements:

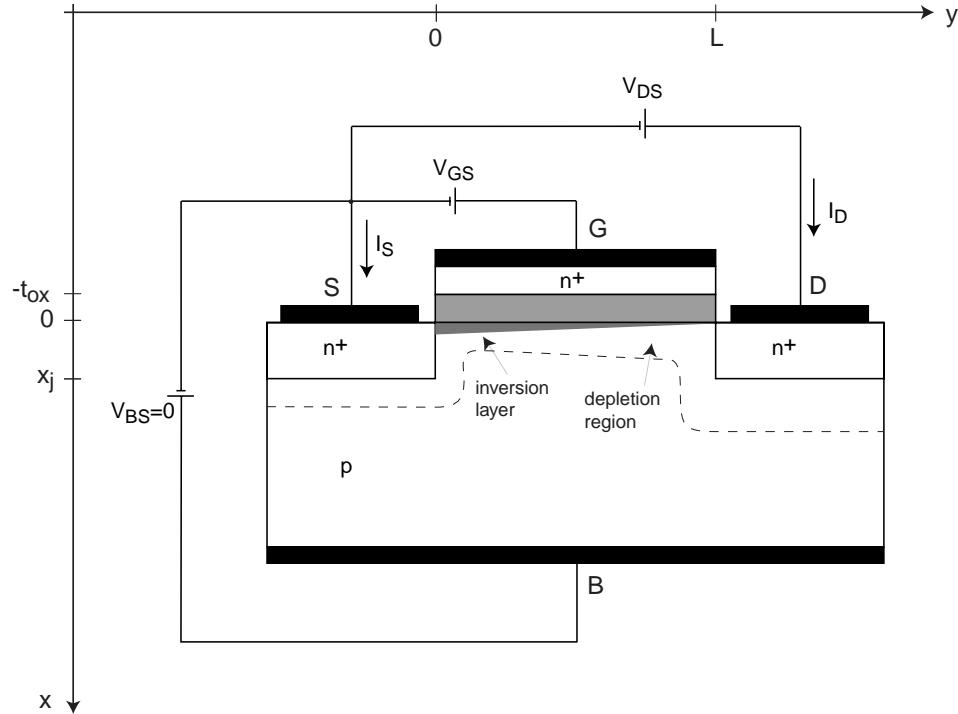
Quiz 1: 10/13, 7:30-9:30 PM, (lectures #1-9); open book; must have calculator.

## Key questions

- How does the MOSFET work in saturation?
- Does the pinch-off point represent a block to current flow?
- How come the MOSFET current still increases a bit with  $V_{DS}$  in saturation?
- How does the application of a back bias affect the MOSFET I-V characteristics?

# 1. The saturation regime

Geometry of problem:



Regimes of operation so far ( $V_{BS} = 0$ ):

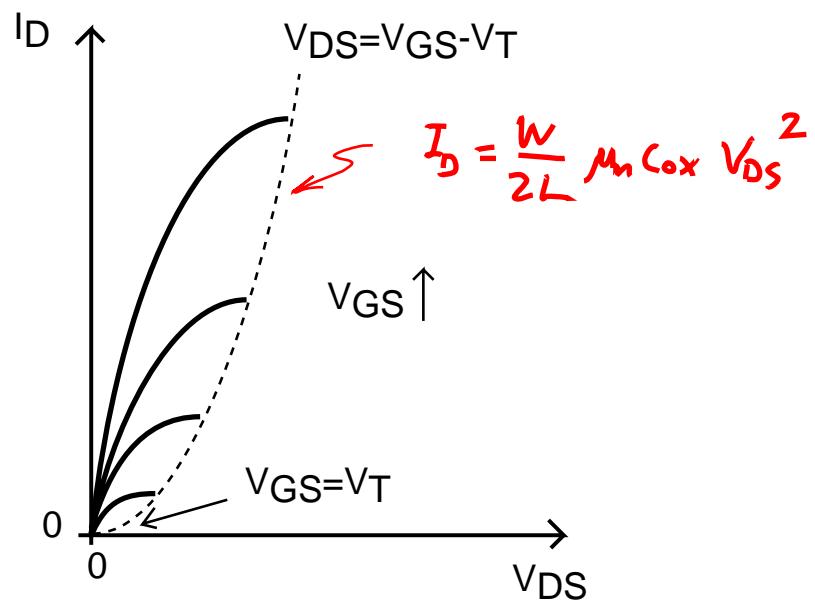
- *Cut-off*:  $V_{GS} < V_T$ ,  $V_{GD} < V_T$ :  
no inversion layer anywhere underneath gate

$$I_D = 0$$

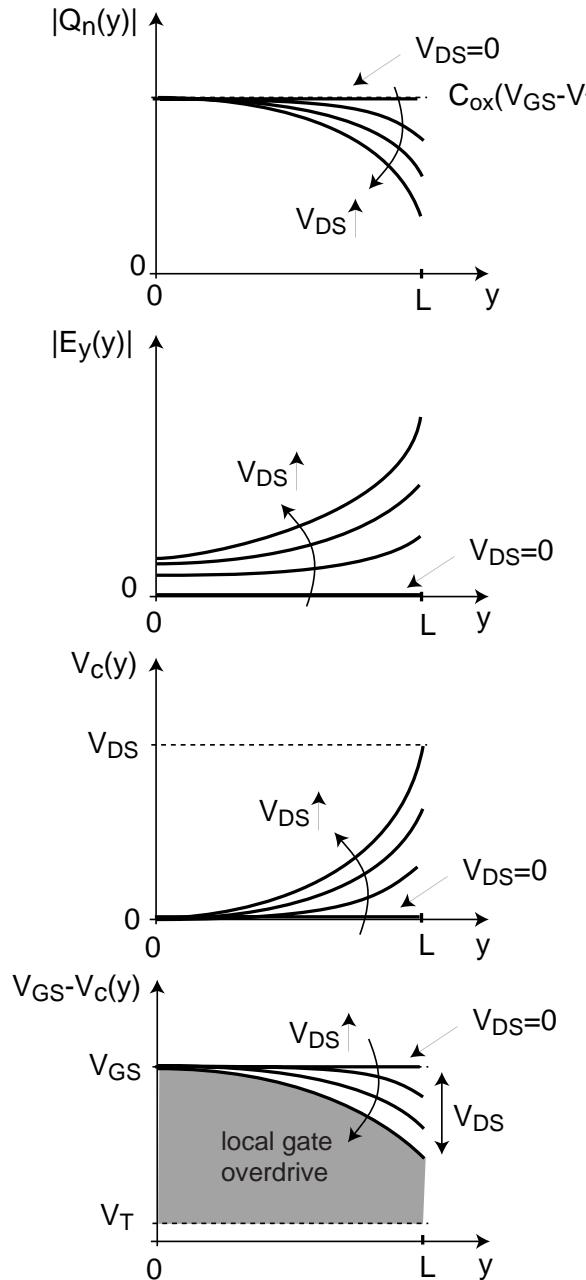
- *Linear*:  $V_{GS} > V_T$ ,  $V_{GD} > V_T$  (with  $V_{DS} > 0$ ):  
inversion layer everywhere underneath gate

$$I_D = \frac{W}{L} \mu_n C_{ox} \left( V_{GS} - \frac{V_{DS}}{2} - V_T \right) V_{DS}$$

Output characteristics:



- Review of  $Q_n$ ,  $E_y$ ,  $V_c$ , and  $V_{GS} - V_c(y)$  in linear regime as  $V_{DS}$  increases:



Ohmic drop along channel debiases inversion layer  
 $\Rightarrow I_D$  rises more slowly with  $V_{DS}$

$\Rightarrow I_D$  saturates as  $V_{DS}$  approaches  $V_{GS} - V_T$

□ Drain current saturation

As  $V_{DS}$  approaches:

$$V_{DSsat} = V_{GS} - V_T$$

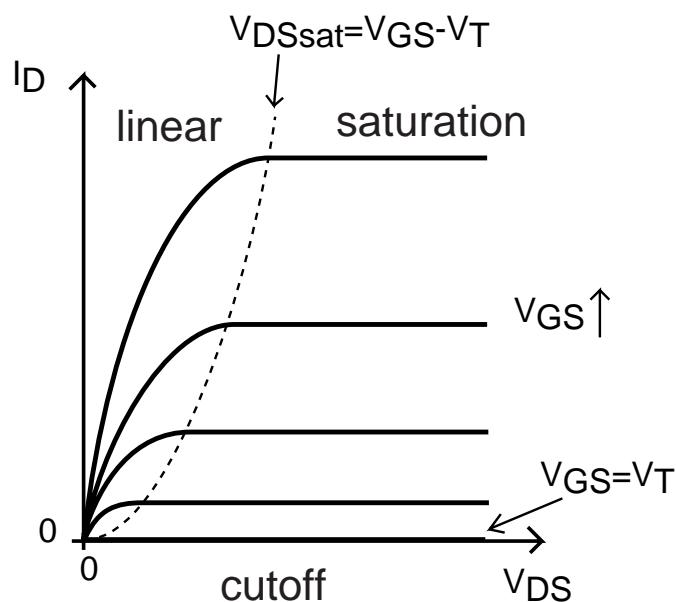
increase in  $|E_y|$  compensated by decrease in  $|Q_n|$   
 $\Rightarrow I_D$  saturates to:

$$I_{Dsat} = I_{Dlin}(V_{DS} = V_{DSsat} = V_{GS} - V_T)$$

Then:

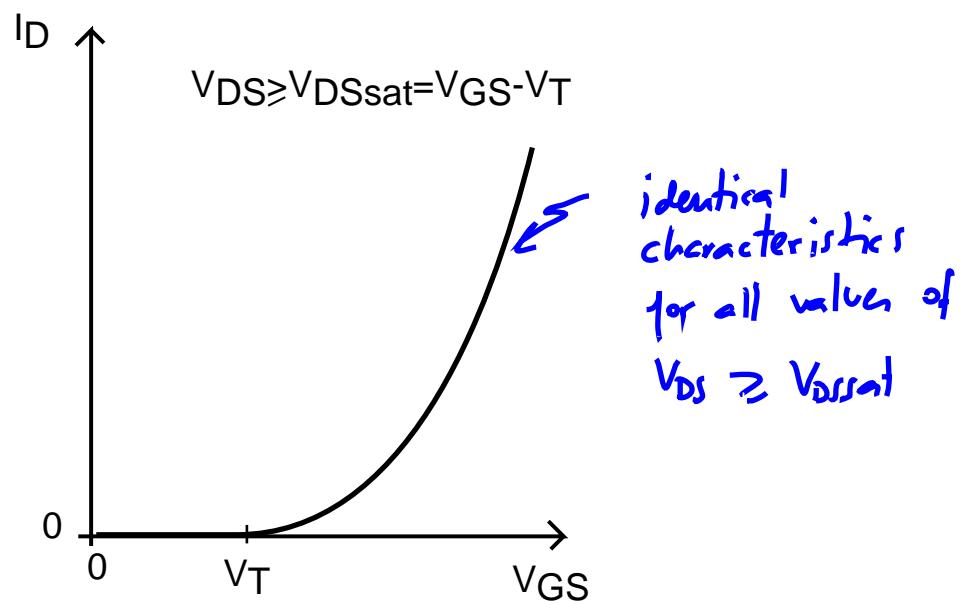
$$I_{Dsat} = \frac{W}{2L} \mu_n C_{ox} (V_{GS} - V_T)^2$$

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$$I_{Dsat} = \frac{W}{2L} \mu_n C_{ox} (V_{GS} - V_T)^2$$

Transfer characteristics in saturation:



- What happens when  $V_{DS} = V_{GS} - V_T$ ?

Charge control relation at drain-end of channel:

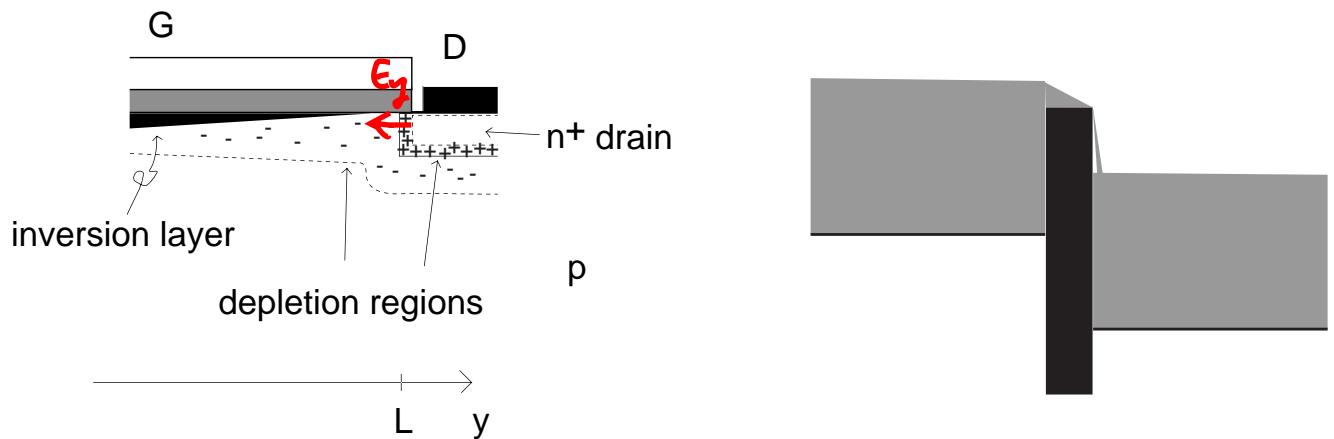
$$Q_n(L) = -C_{ox}(V_{GS} - V_{DS} - V_T) = 0$$

*$\curvearrowleft V_c(y=L)$*

No inversion layer at end of channel??!!  $\Rightarrow$  **Pinch-off**

At pinch-off:

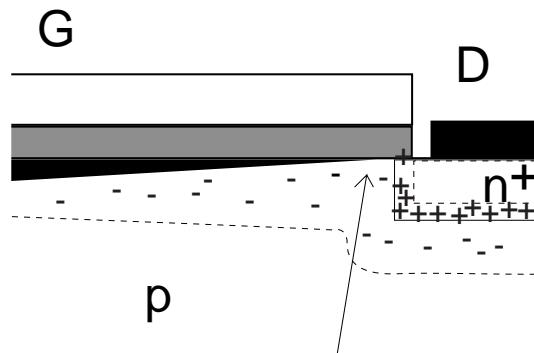
- charge control equation inaccurate around  $V_T$
- electron concentration small but not zero
- electrons move fast because electric field is very high
- dominant electrostatic feature: acceptor charge
- there is no barrier to electron flow (on the contrary!)



- Key dependencies of  $I_{Dsat}$

- $I_{Dsat} \propto (V_{GS} - V_T)^2$

Voltage at pinch-off point ( $V_c = 0$  at source):



$$V_c(L) = V_{DSsat} = V_{GS} - V_T$$

Drain current at pinch-off:

$$\propto \text{lateral electric field} \propto V_{DSsat} = V_{GS} - V_T$$

$$\propto \text{electron concentration} \propto V_{GS} - V_T$$

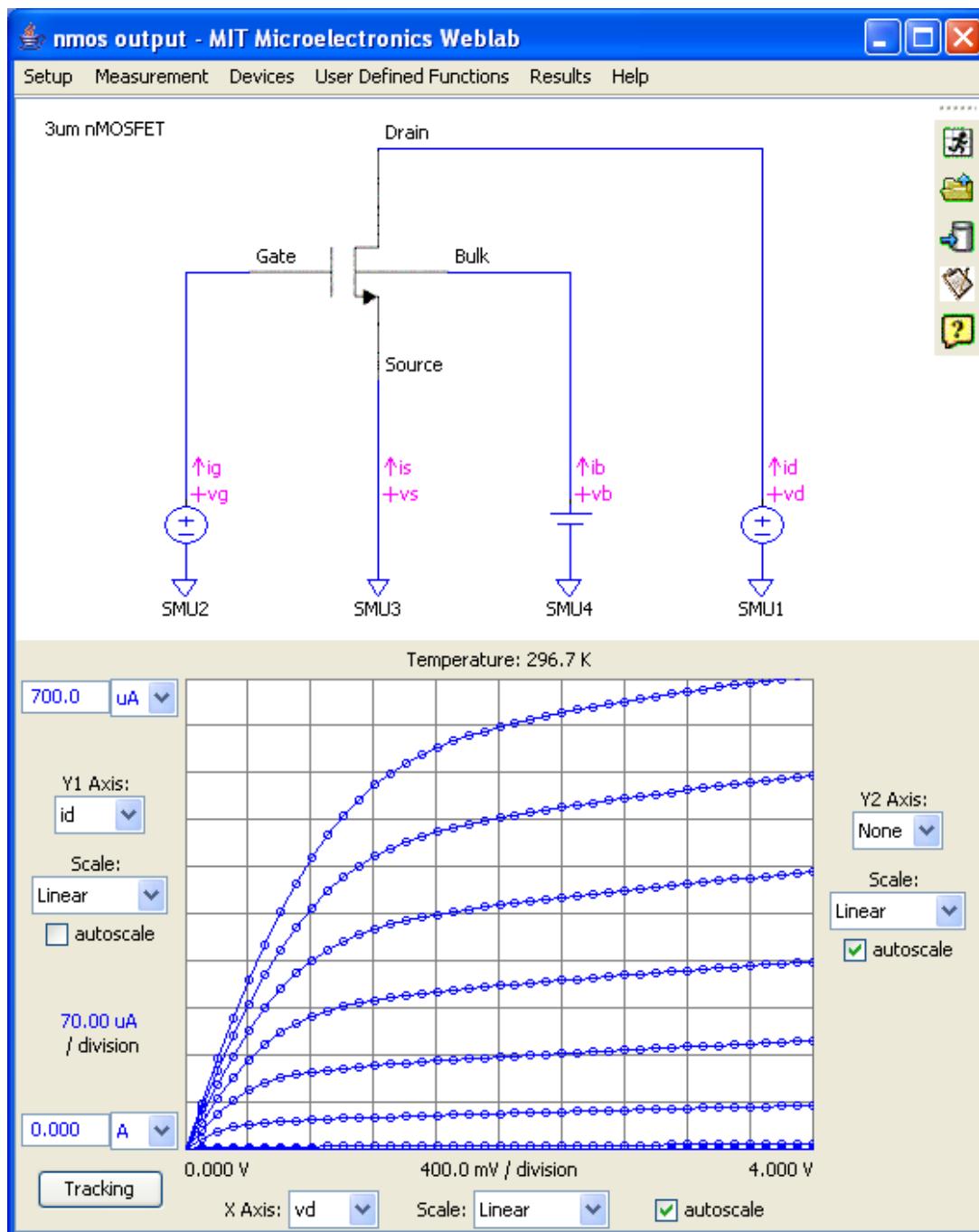
$$\Rightarrow I_{Dsat} \propto (V_{GS} - V_T)^2$$

- $I_{Dsat} \propto \frac{1}{L}$

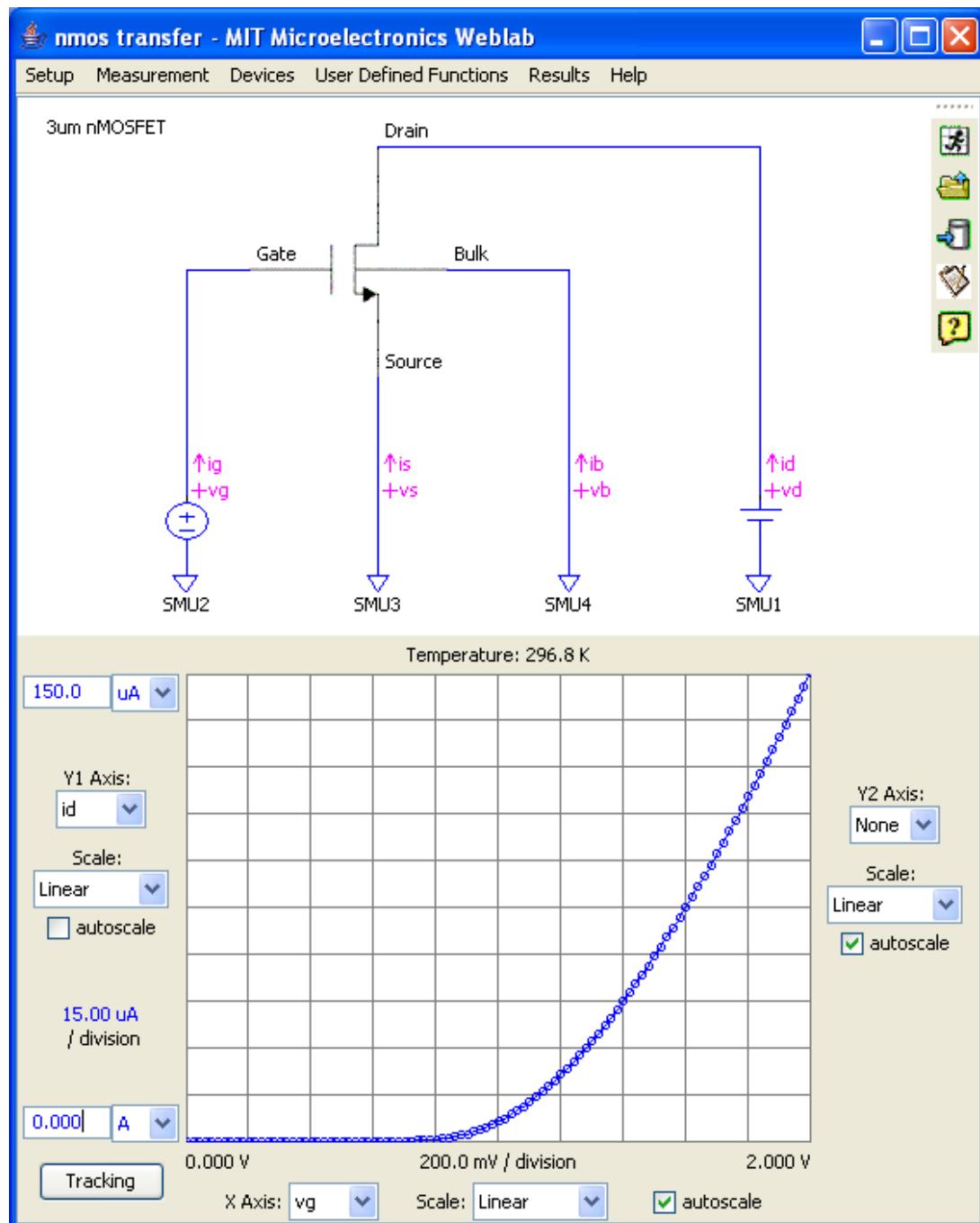
$$L \downarrow \rightarrow |E_y| \uparrow$$

## $3\mu m$ n-channel MOSFET

Output characteristics ( $V_{GS} = 0 - 4 V$ ,  $\Delta V_{GS} = 0.5 V$ ):

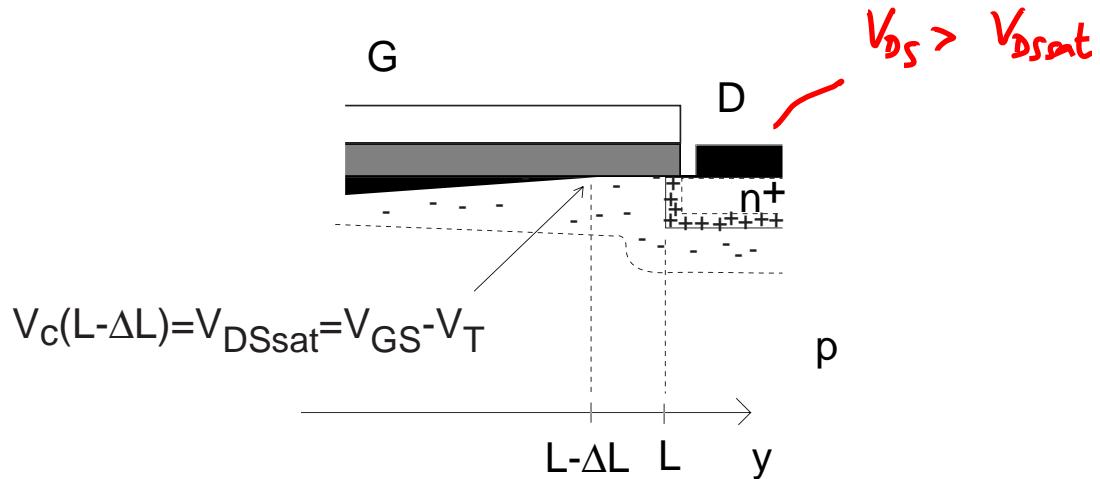


Transfer characteristics in saturation ( $V_{DS} = 3\text{ V}$ ):



- What happens if  $V_{DS} > V_{GS} - V_T$ ?

Depletion region separating pinch-off point and drain widens  
(just like in reverse-biased pn junction)



To first order,  $I_D$  does not increase past pinchoff:

$$I_D = I_{Dsat} = \frac{W}{2L} \mu_n C_{ox} (V_{GS} - V_T)^2$$

To second order, electrical channel length affected  
("*channel-length modulation*"):  $V_{DS} \uparrow \Rightarrow L_{channel} \downarrow \Rightarrow I_D \uparrow$

$$I_D \propto \frac{1}{L - \Delta L} \simeq \frac{1}{L} \left(1 + \frac{\Delta L}{L}\right)$$

↑ small
↑ Taylor series expansion

Experimental finding:

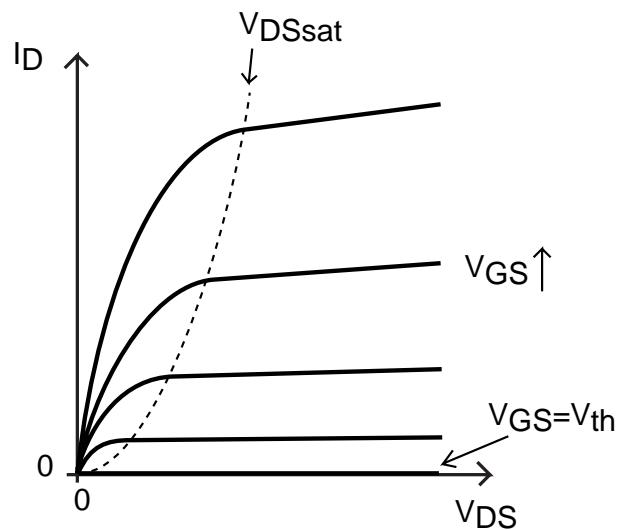
$$\Delta L \propto V_{DS} - V_{DSsat}$$

Hence:

$$\frac{\Delta L}{L} = \lambda(V_{DS} - V_{DSsat})$$

Improved model in saturation:

$$I_{Dsat} = \frac{W}{2L} \mu_n C_{ox} (V_{GS} - V_T)^2 [1 + \lambda(V_{DS} - V_{DSsat})]$$



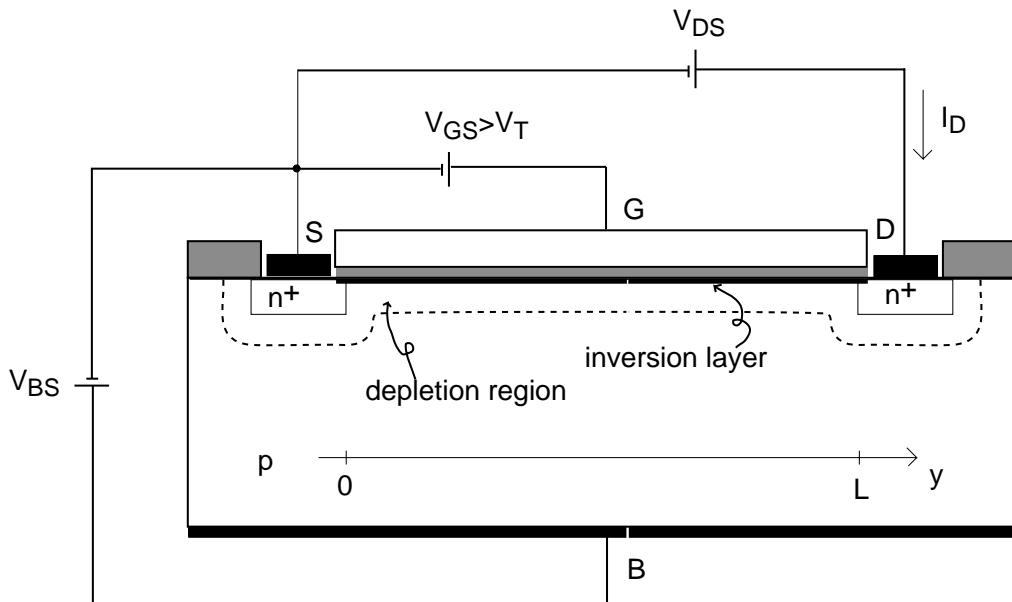
Also, experimental finding:

$$\lambda \propto \frac{1}{L}$$

## 2. Backgate characteristics

There is a fourth terminal in a MOSFET: the *body*.

What does the body do?



Body contact allows application of bias to body with respect to inversion layer,  $V_{BS}$ .

Only interested in  $V_{BS} < 0$  (pn diode in reverse bias).

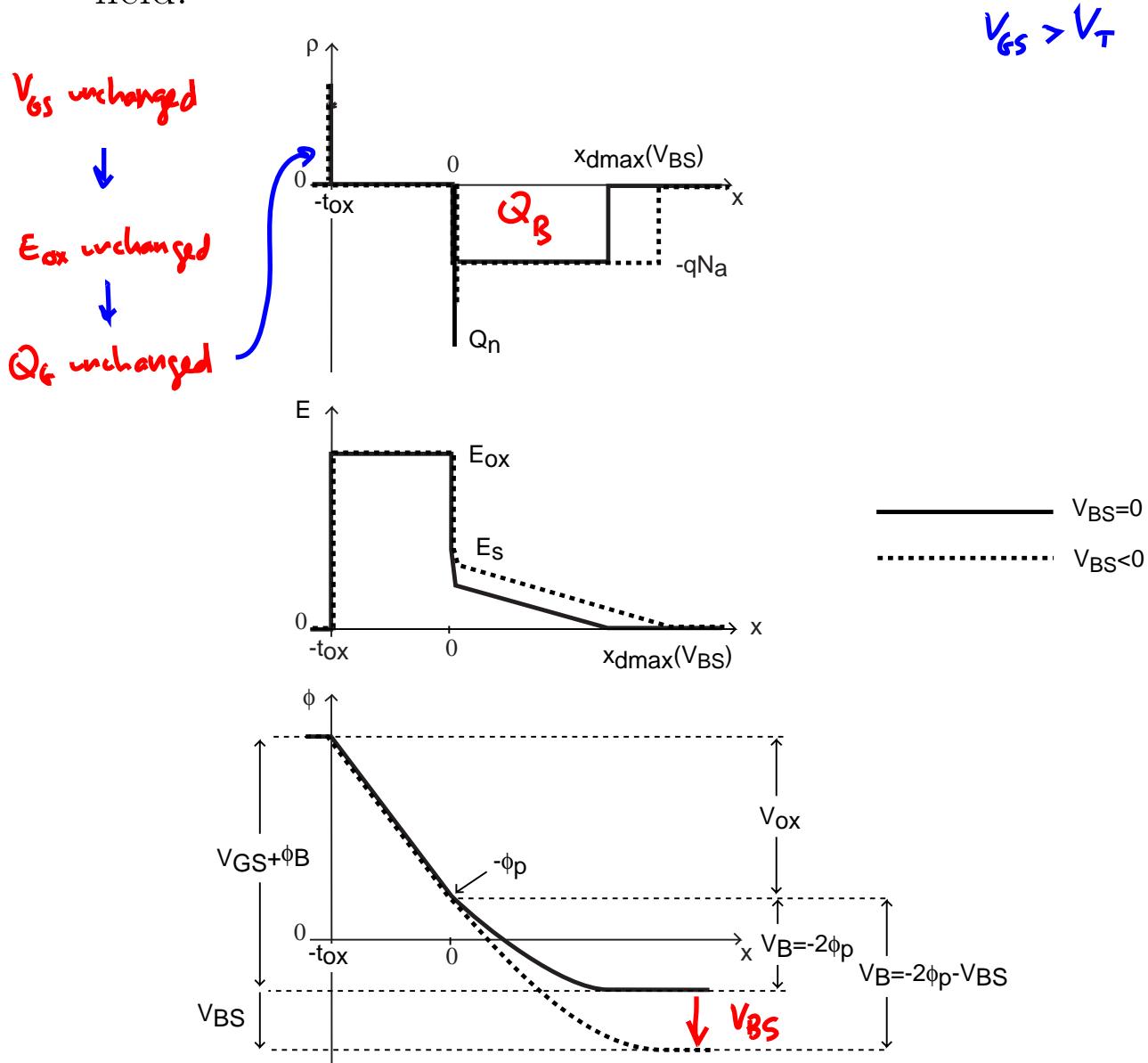
Interested in effect on inversion layer

$\Rightarrow$  examine for  $V_{GS} > V_T$  (keep  $V_{GS}$  constant).

Application of  $V_{BS} < 0$  increases potential build-up across semiconductor:

$$-2\phi_p \Rightarrow -2\phi_p - V_{BS}$$

Depletion region must widen to produce required extra field:



Consequences of application of  $V_{BS} < 0$ :

- $-2\phi_p \Rightarrow -2\phi_p - V_{BS}$
- $|Q_B| \uparrow \Rightarrow x_{dmax} \uparrow$
- since  $V_{GS}$  constant,  $V_{ox}$  unchanged  
 $\Rightarrow E_{ox}$  unchanged  
 $\Rightarrow |Q_s| = |Q_G|$  unchanged
- $|Q_s| = |Q_n| + |Q_B|$  unchanged, but  $|Q_B| \uparrow \Rightarrow |Q_n| \downarrow$   
 $\Rightarrow$  inversion layer harge is reduced!

Application of  $V_{BS} < 0$  with constant  $V_{GS}$  reduces electron concentration in inversion layer  $\Rightarrow V_T \uparrow$

How does  $V_T$  change with  $V_{BS}$ ?

In  $V_T$  formula change  $-2\phi_p$  to  $-2\phi_p - V_{BS}$ :

$$V_T^{GB}(V_{BS}) = V_{FB} - 2\phi_p - V_{BS} + \gamma\sqrt{(-2\phi_p - V_{BS})}$$

In MOSFETs, interested in  $V_T$  between gate and source:

$$V_{GB} = V_{GS} - V_{BS} \Rightarrow V_T^{GB} = V_T^{GS} - V_{BS}$$

Then:

$$V_T^{GS} = V_T^{GB} + V_{BS}$$

And:

$$V_T^{GS}(V_{BS}) = V_{FB} - 2\phi_p + \gamma\sqrt{(-2\phi_p - V_{BS})} \equiv V_T(V_{BS})$$

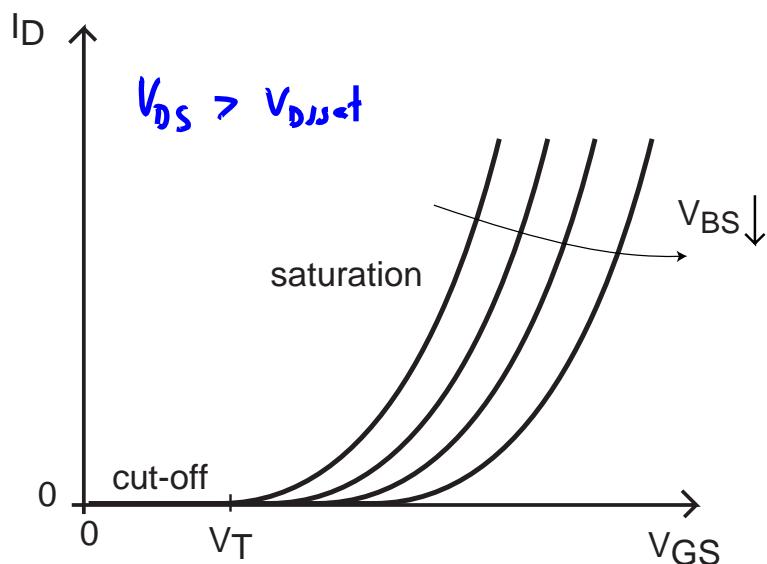
In the context of the MOSFET,  $V_T$  is always defined in terms of *gate-to-source voltage*.

Define:

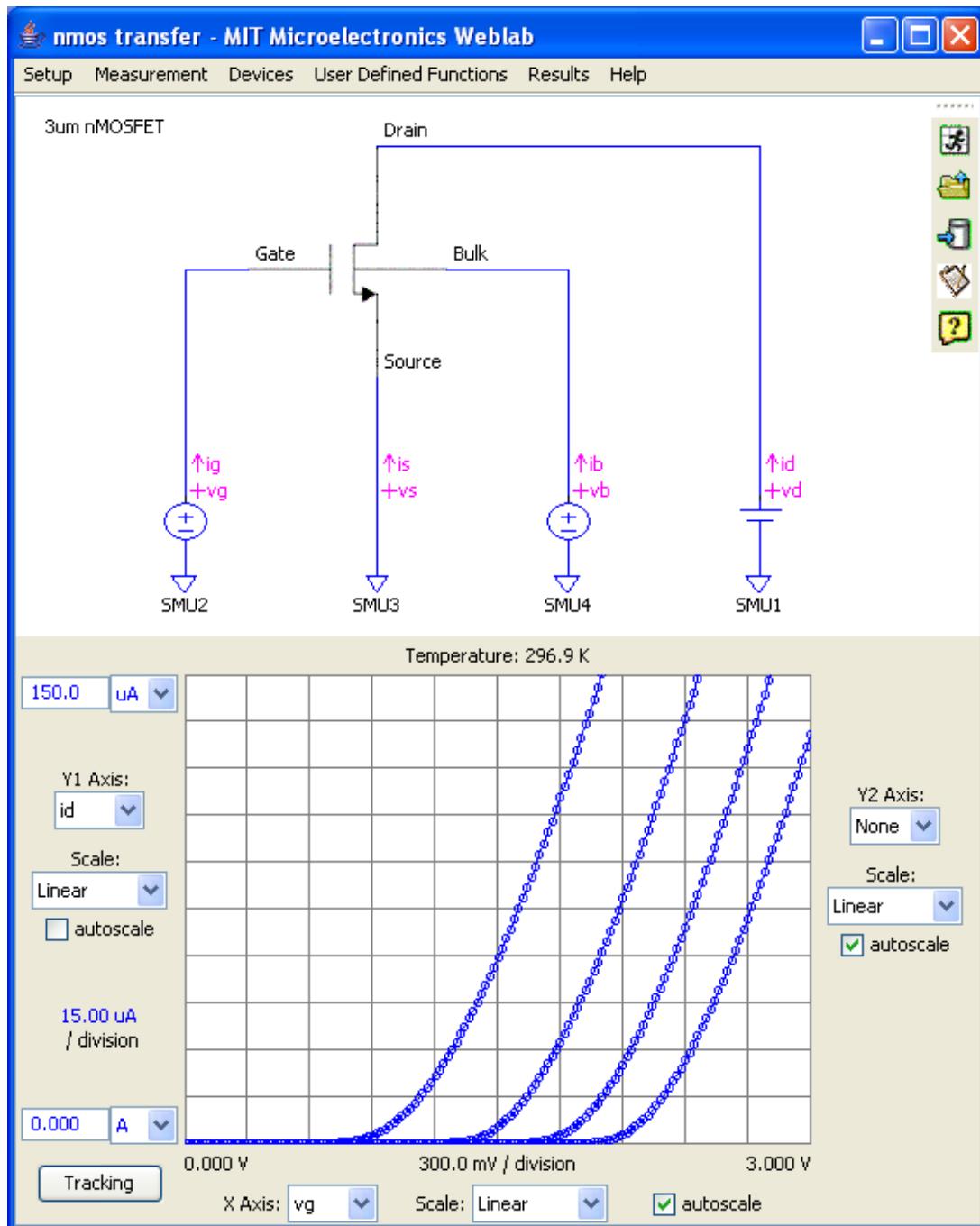
$$V_{To} = V_T(V_{BS} = 0)$$

Then:

$$V_T(V_{BS}) = V_{To} + \gamma(\sqrt{-2\phi_p - V_{BS}} - \sqrt{-2\phi_p})$$



Backgate characteristics ( $V_{BS} = 0, -1, -2, -3 \text{ V}$ ,  $V_{DS} = 3 \text{ V}$ ):



## Key conclusions

- MOSFET in saturation ( $V_{DS} \geq V_{DSsat}$ ): *pinch-off* point at drain-end of channel
  - electron concentration small, but
  - electrons move very fast;
  - pinch-off point does not represent a barrier to electron flow
- In saturation,  $I_D$  saturates:

$$I_{Dsat} = \frac{W}{2L} \mu_n C_{ox} (V_{GS} - V_T)^2$$

- But..., due to *channel-length modulation*,  $I_{Dsat}$  increases slightly with  $V_{DS}$
- Application of back bias shifts  $V_T$  (*back-gate effect*)