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16.323 Principles of Optimal Control
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Topic #14

16.31 Feedback Control Systems

MIMO Systems

- Singular Value Decomposition

Multivariable Frequency Response

- In the MIMO case, the system $G(s)$ is described by a $p \times m$ **transfer function matrix** (TFM)
 - Still have that $G(s) = C(sI - A)^{-1}B + D$
 - But $G(s) \rightarrow A, B, C, D$ MUCH less obvious than in SISO case.
 - Also seen that the discussion of poles and zeros of MIMO systems is much more complicated.

- In SISO case we use the Bode plot to develop a measure of the system “size”.

- Given $z = Gw$, where $G(j\omega) = |G(j\omega)|e^{j\phi(\omega)}$

- Then $w = |w|e^{j(\omega_1 t + \psi)}$ applied to $|G(j\omega)|e^{j\phi(\omega)}$ yields

$$|w||G(j\omega_1)|e^{j(\omega_1 t + \psi + \phi(\omega_1))} = |z|e^{j(\omega_1 t + \psi_0)} \equiv z$$

- Amplification and phase shift of the input signal obvious in the SISO case.

- MIMO extension?

- Is the response of the system large or small?

$$G(s) = \begin{bmatrix} 10^3/s & 0 \\ 0 & 10^{-3}/s \end{bmatrix}$$

- For MIMO systems, cannot just plot all of the g_{ij} elements of G
 - Ignores the coupling that might exist between them.
 - So not enlightening.
- **Basic MIMO frequency response:**
 - Restrict all inputs to be at the same frequency
 - Determine how the system responds at that frequency
 - See how this response changes with frequency
- So inputs are $\mathbf{w} = \mathbf{w}_c e^{j\omega t}$, where $\mathbf{w}_c \in \mathbb{C}^m$
 - Then we get $\mathbf{z} = G(s)|_{s=j\omega} \mathbf{w}$, $\Rightarrow \mathbf{z} = \mathbf{z}_c e^{j\omega t}$ and $\mathbf{z}_c \in \mathbb{C}^p$
 - We need only analyze $\mathbf{z}_c = G(j\omega) \mathbf{w}_c$
- As in the SISO case, we need a way to establish if the system response is **large** or **small**.
 - How much amplification we can get with a bounded input.
- Consider $\mathbf{z}_c = G(j\omega) \mathbf{w}_c$ and set $\|\mathbf{w}_c\|_2 = \sqrt{\mathbf{w}_c^H \mathbf{w}_c} \leq 1$. What can we say about the $\|\mathbf{z}_c\|_2$?
 - Answer depends on ω and on the **direction** of the input \mathbf{w}_c
 - Best found using **singular values**.

- Must perform **SVD** of the matrix $G(s)$ at each frequency $s = j\omega$

$$G(j\omega) \in \mathbb{C}^{p \times m} \quad U \in \mathbb{C}^{p \times p} \quad \Sigma \in \mathbb{R}^{p \times m} \quad V \in \mathbb{C}^{m \times m}$$

$$G = U\Sigma V^H$$

– $U^H U = I$, $U U^H = I$, $V^H V = I$, $V V^H = I$, and Σ is diagonal.

– Diagonal elements $\sigma_k \geq 0$ of Σ are the **singular values** of G .

$$\sigma_i = \sqrt{\lambda_i(G^H G)} \quad \text{or} \quad \sigma_i = \sqrt{\lambda_i(G G^H)}$$

the positive ones are the same from both formulas.

– Columns of matrices U and V (u_i and v_j) are the associated eigenvectors

$$G^H G v_j = \sigma_j^2 v_j$$

$$G G^H u_i = \sigma_i^2 u_i$$

$$G v_i = \sigma_i u_i$$

- If the $\text{rank}(G) = r \leq \min(p, m)$, then
 - $\sigma_k > 0$, $k = 1, \dots, r$
 - $\sigma_k = 0$, $k = r + 1, \dots, \min(p, m)$
 - Singular values are sorted so that $\sigma_1 \geq \sigma_2 \geq \dots \geq \sigma_r$
- An SVD gives a **very detailed description** of how a matrix (the system G) acts on a vector (the input w) at a particular frequency.

- So how can we use this result?
 - Fix the size $\|\mathbf{w}_c\|_2 = 1$ of the input, and see how large we can make the output.
 - Since we are working at a single frequency, we just analyze the relation

$$\mathbf{z}_c = G_w \mathbf{w}_c, \quad G_w \equiv G(s = \mathbf{j}\omega)$$

- Define the maximum and minimum amplifications as:

$$\bar{\sigma} \equiv \max_{\|\mathbf{w}_c\|_2=1} \|\mathbf{z}_c\|_2$$

$$\underline{\sigma} \equiv \min_{\|\mathbf{w}_c\|_2=1} \|\mathbf{z}_c\|_2$$

- Then we have that (let $q = \min(p, m)$)

$$\bar{\sigma} = \sigma_1$$

$$\underline{\sigma} = \begin{cases} \sigma_q & p \geq m \quad \text{“tall”} \\ 0 & p < m \quad \text{“wide”} \end{cases}$$

- Can use $\bar{\sigma}$ and $\underline{\sigma}$ to determine the possible amplification and attenuation of the input signals.
- Since $G(s)$ changes with frequency, so will $\bar{\sigma}$ and $\underline{\sigma}$

- Consider (wide case)

$$G_w = \begin{bmatrix} 5 & 0 & 0 \\ 0 & 0.5 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix} \begin{bmatrix} 5 & 0 & 0 \\ 0 & 0.5 & 0 \end{bmatrix} \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$$

$$= U\Sigma V^H$$

so that $\sigma_1 = 5$ and $\sigma_2 = 0.5$

$$\bar{\sigma} \equiv \max_{\|\mathbf{w}_c\|_2=1} \|G_w \mathbf{w}_c\|_2 = 5 = \sigma_1$$

$$\underline{\sigma} \equiv \min_{\|\mathbf{w}_c\|_2=1} \|G_w \mathbf{w}_c\|_2 = 0 \neq \sigma_2$$

- But now consider (tall case)

$$\tilde{G}_w = \begin{bmatrix} 5 & 0 \\ 0 & 0.5 \\ 0 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix} \begin{bmatrix} 5 & 0 \\ 0 & 0.5 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}$$

$$= U\Sigma V^H$$

so that $\sigma_1 = 5$ and $\sigma_2 = 0.5$ still.

$$\bar{\sigma} \equiv \max_{\|\mathbf{w}_c\|_2=1} \|G_w \mathbf{w}_c\|_2 = 5 = \sigma_1$$

$$\underline{\sigma} \equiv \min_{\|\mathbf{w}_c\|_2=1} \|G_w \mathbf{w}_c\|_2 = 0.5 = \sigma_2$$

- For MIMO systems, the gains (or σ 's) are only part of the story, as we must also consider the **input direction**.

- To analyze this point further, note that we can rewrite

$$G_w = U\Sigma V^H = \begin{bmatrix} u_1 & \dots & u_p \end{bmatrix} \begin{bmatrix} \sigma_1 & & \\ & \ddots & \\ & & \sigma_m \\ & & & 0 \end{bmatrix} \begin{bmatrix} v_1^H \\ \vdots \\ v_m^H \end{bmatrix}$$

$$= \sum_{i=1}^m \sigma_i u_i v_i^H$$

– Assumed tall case for simplicity, so $p > m$ and $q = m$

- Can now analyze impact of **various alternatives for the input**
 - Only looking at one frequency, so the basic signal is harmonic.
 - But, we are free to pick the relative **sizes** and **phases** of each of the components of the input vector \mathbf{w}_c .
 - ◊ These define the **input direction**

- For example, we could pick $\mathbf{w}_c = v_1$, then

$$\mathbf{z}_c = G_w \mathbf{w}_c = \left(\sum_{i=1}^m \sigma_i u_i v_i^H \right) v_1 = \sigma_1 u_1$$

since $v_i^H v_j = \delta_{ij}$.

- Output amplified by σ_1 . The relative **sizes** and **phases** of each of the components of the output are given by the vector \mathbf{z}_c .

- By selecting other input directions (at the same frequency), we can get quite different amplifications of the input signal

$$\underline{\sigma} \leq \frac{\|G_w \mathbf{w}_c\|_2}{\|\mathbf{w}_c\|_2} \leq \bar{\sigma}$$

- Thus we say that

- G_w is large if $\underline{\sigma}(G_w) \gg 1$

- G_w is small if $\bar{\sigma}(G_w) \ll 1$

- **MIMO frequency response** are plots of $\bar{\sigma}(j\omega)$ and $\underline{\sigma}(j\omega)$.

- Then use the singular value vectors to analyze the response at a particular frequency.