

# 16.06 Principles of Automatic Control

## Lecture 34

Zeros:  $z = -0.9867.. \rightarrow W = -30,000$   
 $z = \infty \rightarrow W = +200$   
 Poles:  $z = 0.9802 \rightarrow W = -2$

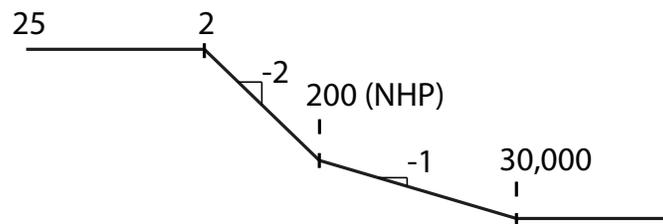
Bode Gain:  $G_d(1) = 25$ .

Also, need to map crossover frequency:

$$\nu_c = \frac{\tan(\omega_c T/2)}{T/2} = 51.068$$

There's not much warping at  $\omega T = 0.5$  (2%). At  $\omega T = 1.0$ , there is 9% warping; at  $\omega T = 1.5$ , 24%. So for most problems, may not need to prewarp.

Bode plot of G:



To meet specs, need lead around  $\nu_c = 51$ , lag below  $\nu = 5.1$ .

$$\angle G(j51) = -189.7^\circ$$

So need phase from lead compensator

$$\begin{aligned} \phi_{\text{lead}} &= -180^\circ - \angle G(j51) + 6^\circ + \text{PM} \\ &= 65.7^\circ \end{aligned}$$

where  $6^\circ$  anticipates lag compensation.

$$\begin{aligned} \Rightarrow \sqrt{\frac{b}{a}} &= 4.65 \\ \Rightarrow b &= 51 \cdot 4.65 = 237 \\ a &= 51/4.65 = 11.0 \\ \Rightarrow K_{\text{lead}} &= 5.44 \frac{1 + W/11}{1 + W/237} \end{aligned}$$

For this compensator,  $K_p = 136$ . So need lag ratio of 36.76.

$$\Rightarrow K_{\text{lag}} = \frac{W + 5}{W + 0.136}$$

Therefore, the compensator is

$$K(W) = 5.44 \frac{W + 5}{W + 0.136} \frac{1 + W/11}{1 + W/237}$$

The discrete time compensator is the Tustin transform, yielding

$$K_d(z) = 57.97 \frac{(z - 0.9512)(z - 0.8957)}{(z - 0.9986)(z + 0.0847)}$$

Remarks:

1. Compensator is almost identical to what would be found if we used emulation, *if* we include effective  $T/2$  delay of ZOH.
2. W-Transform approach guarantees stability of discrete-time system. Not really an issue for  $\omega_c T = 0.5$ , but might be for faster crossover.
3. Note RHP zero at  $wW = +\frac{2}{T} = 200$ . This zero limits achievable bandwidth of controller, just like time delay.

### Example

Same plant as above. Make  $\omega_c$  ( $\nu_c$ ) as high as practical and still have  $50^\circ$  phase margin.

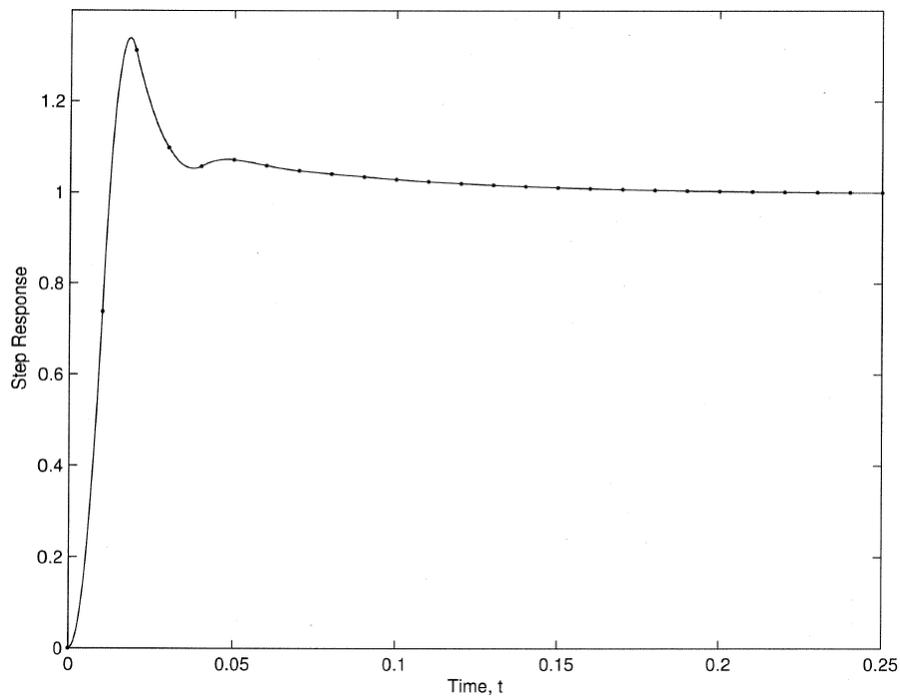
Solution: Let's pick crossover frequency  $\nu_c$  to be factor of only 2 below NMP zero.

$$\nu_c = 100$$

Use lead compensator to get desired PM and crossover:

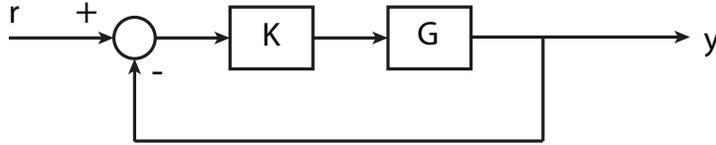
$$\begin{aligned} \angle G_w(j\omega_0) &= -204.1^\circ \\ \phi_{\text{lead}} &= -180 - \angle G + \text{PM} = 74.1^\circ \\ \Rightarrow \sqrt{\frac{b}{a}} &= 7.15 \\ b &= 100 \cdot 7.15 = 715 \\ a &= 100/7.15 = 14 \\ \Rightarrow K_N(W) &= 12.53 \frac{1 + W/14}{1 + W/715} \\ \Rightarrow K_d(z) &= 149.7 \frac{z - 0.8692}{z + 0.5628} \end{aligned}$$

See step response below:



## Direct Design

Suppose we have the usual unity feedback control structure:



(The system might be continuous or discrete). Suppose we want the closed loop transfer function

$$H = \frac{KG}{1 + KG}$$

to have a specific form, e.g., have a particular rise time, settling time, etc. Why not just *solve* for desired  $K$  in terms of  $G, H$ ?

$$\begin{aligned} H(1 + KG) &= KG \\ H &= -KGH + KG \\ K &= \frac{1}{G} \frac{H}{1 - H} \end{aligned}$$

So

$$K = \frac{1}{G} \frac{H}{1 - H}$$

Note that  $K$  essentially cancels  $G$  with the factor  $1/G$ , so makes the loop gain

$$K = \frac{H}{1 - H}$$

exactly what is needed to have the desired closed loop transfer function.

But we can't choose any  $H$  desired!

### Constraints on H:

**Stability** In order that  $K$  not cancel on unstable pole or NMP zero, we must have that

1.  $H$  must have as zeros all the zeros of  $G$  outside the unit circle.
2.  $1 - H$  must have as zeros all the unstable poles of  $G$ .

**Causality** In order that  $K$  be causal, we must have:

3. The relative degree of  $H$  is at least as large as the relative degree of  $G$ .

## Example

$$G(s) = \frac{25}{(1 + s/2)^2}$$
$$G_d(z) = 0.004934 \frac{z + 0.9867}{(z - 0.9802)^2}$$
$$T = 0.01$$

do a direct design such that

1. The system is Type 1:  $\Rightarrow H(1) = 1$ .
2. The system is deadbeat (all poles of  $H$  at  $z = 0$ ).

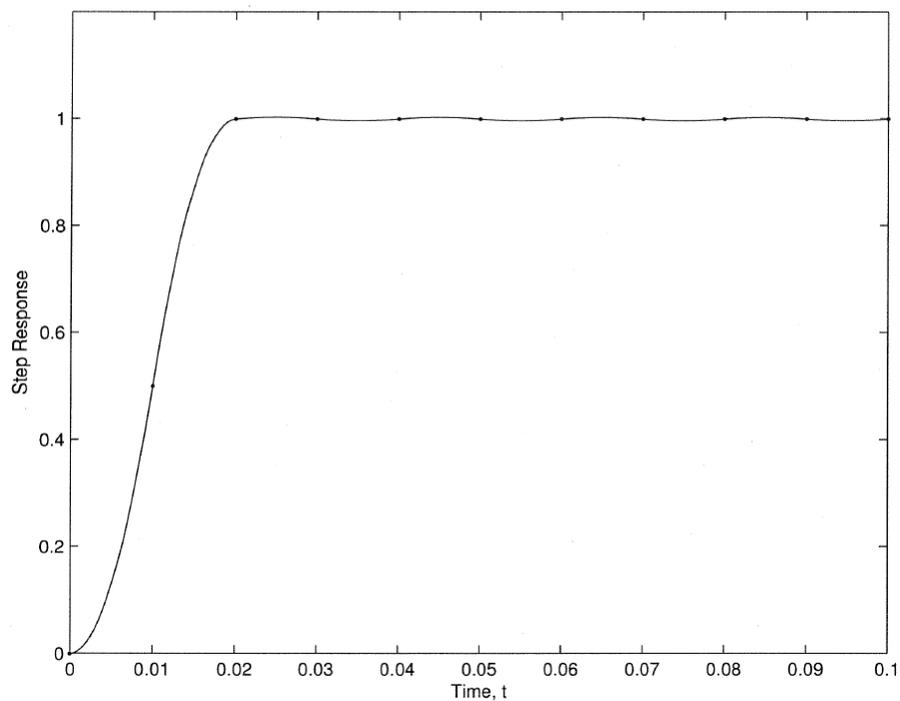
Therefore, might select

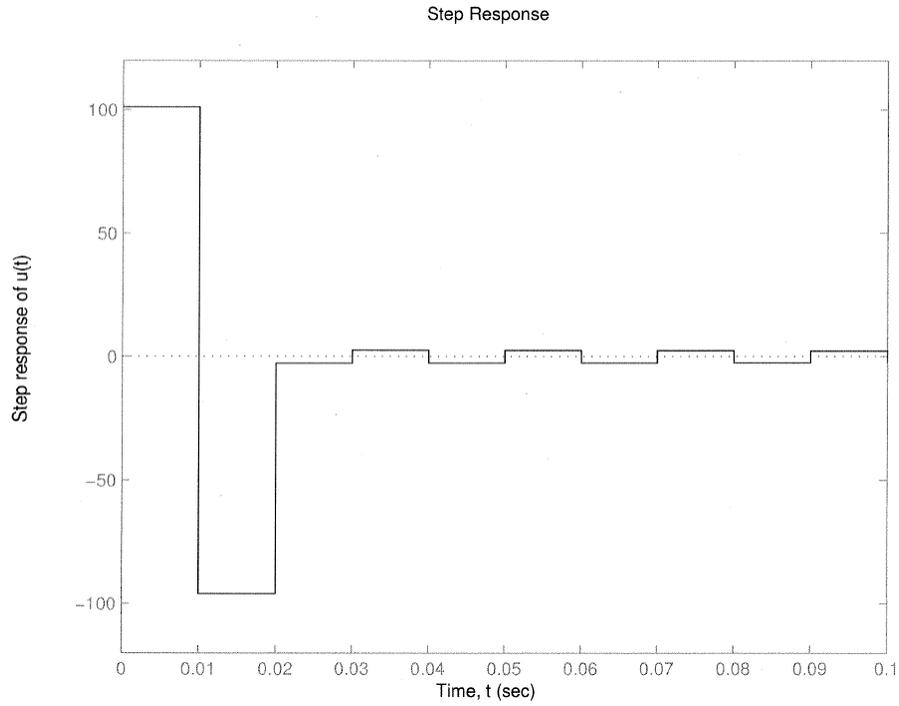
$$H(z) = \frac{1}{2} \frac{z + 1}{z^2}$$

Then

$$K_d(z) = \frac{1}{G_d(z)} \cdot \frac{1}{2} \cdot \frac{z + 1}{z^2 - 0.5z - 0.5}$$

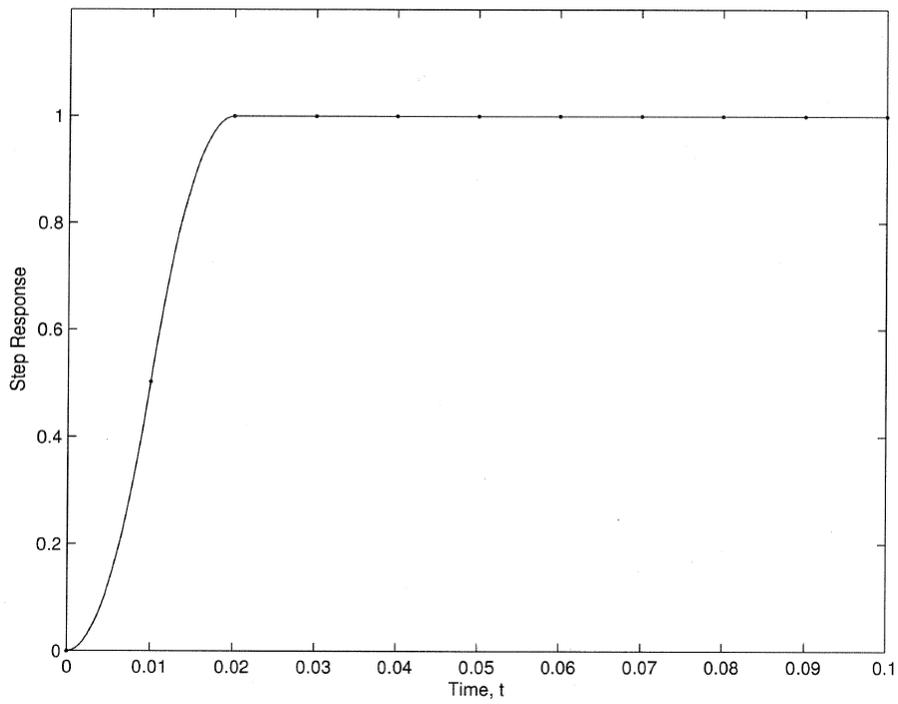
See responses on next two pages. Note the “ringing” in  $u[k]$ . To eliminate, put zero of  $H$  at  $-0.9867$ .



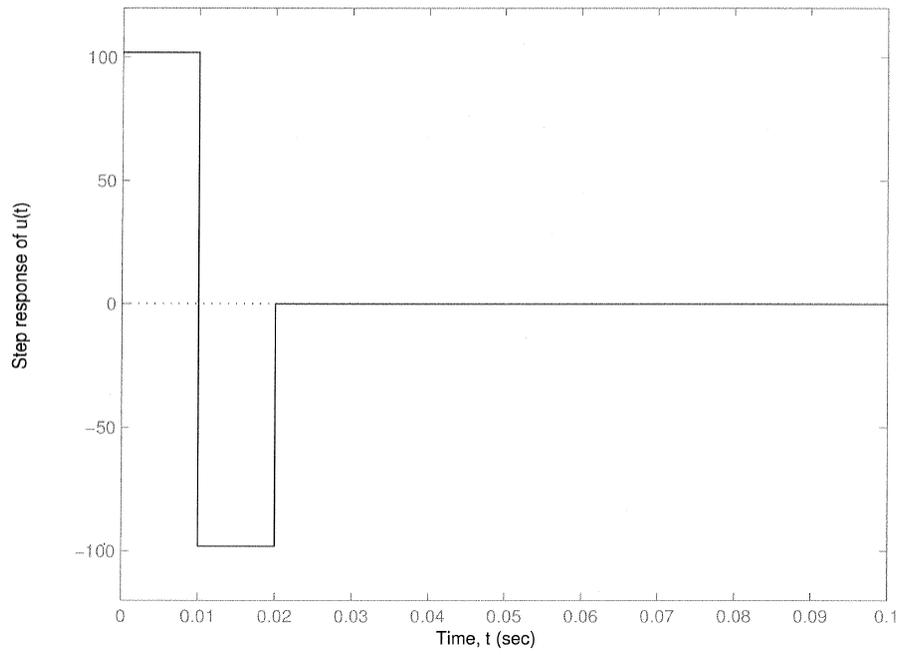


To eliminate ringing, choose

$$H(z) = \frac{1}{1.9867} \frac{z + 0.9867}{z^2}$$



Step Response



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