

# 16.06 Principles of Automatic Control

## Lecture 13

### Root Locus Rules (cont-d)

- **Rule 5** The locus crosses the  $j\omega$  axis at points where the Routh criterion shows a transition in the number of unstable roots.

**Example:**

$$L(s) = \frac{1}{(s+1)^3}$$

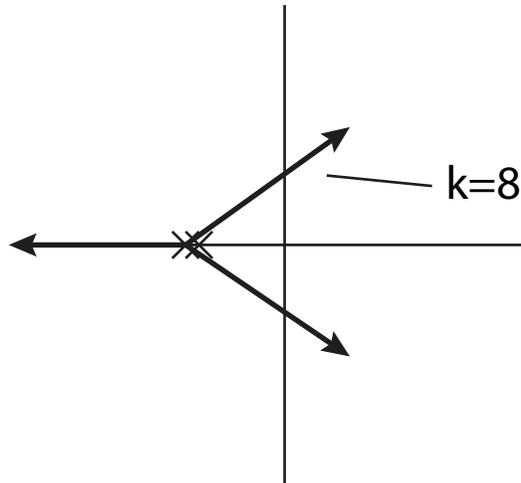
The characteristic equation is

$$s^3 + 3s^2 + 3s + 1 + k = 0$$

The Routh array is

$$\begin{array}{cc} 1 & 3 \\ 3 & 1+k \\ \frac{8-k}{3} & 0 \\ 1+k & 0 \end{array}$$

So the transitions occur at  $k = 8, -1$ . Look at locus:



For  $k = 8$ , the characteristic equation is

$$s^3 + 3s^2 + 3s + 9 = 0$$

which has roots at

$$s = -3, \underbrace{\pm\sqrt{3}j}$$



$j\omega$  axis crossing

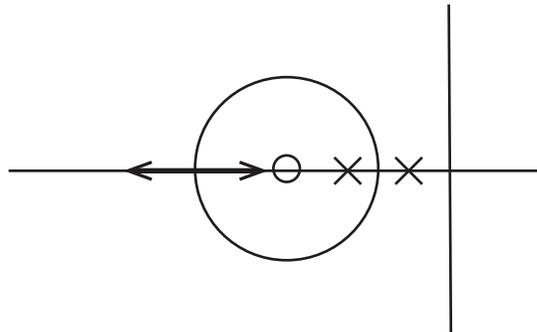
- **Rule 6** The locus will have multiple roots at points on the locus where

$$n(s)\frac{dd(s)}{ds} - d(s)\frac{dn(s)}{ds} = 0$$

(see FPE for details)

**Example:**

$$L(s) = \frac{s + 3}{(s + 1)(s + 2)}$$



where does locus depart/arrive real axis?

$$\begin{aligned}n(s) &= s + 3 \\d(s) &= s^2 + 3s + 2\end{aligned}$$

$$\begin{aligned}n'(s) &= 1 \\d'(s) &= 2s + 3\end{aligned}$$

$$n(s)d'(s) - d(s)n'(s)$$

$$\begin{aligned}&= (s + 3)(2s + 3) - (s^2 + 3s + 2) \\&= 2s^2 + 9s + 9 - (s^2 + 3s + 2) \\&= s^2 + 6s + 7 = 0 \\&\Rightarrow s = -\frac{6}{2} \pm \frac{\sqrt{36 - 28}}{2} \\& s = -3 \pm \sqrt{2}\end{aligned}$$

as in recitation!

## Lead Compensator Example

For the plant

$$G(s) = \frac{2}{(s + 1)(s + 2)}$$

find a unity feedback controller with compensator  $K(s)$  such that

$$\begin{aligned}t_r &\lesssim 0.5 \text{ sec} \\M_p &\lesssim 10\%\end{aligned}$$

If the closed loop system is second order, the poles would need to have

$$M_p = 0.10 = e^{-\pi \tan \theta} \Rightarrow \theta = 0.6325 \Rightarrow \zeta = \sin \theta = 0.5912$$

So set  $\boxed{\zeta = 0.707}$  to allow some margin.

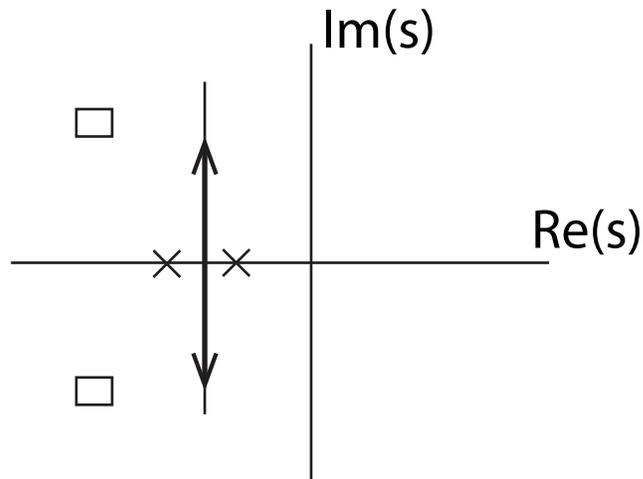
$t_r = \frac{1.8}{\omega_n} \Rightarrow \omega_n \approx 3.6 \text{ rad/sec}$   
 This would place poles at

$$s = -2.6 \pm 2.6j$$

To simplify, want poles at

$$s = -3 \pm 3j$$

Look at locus with gain only:

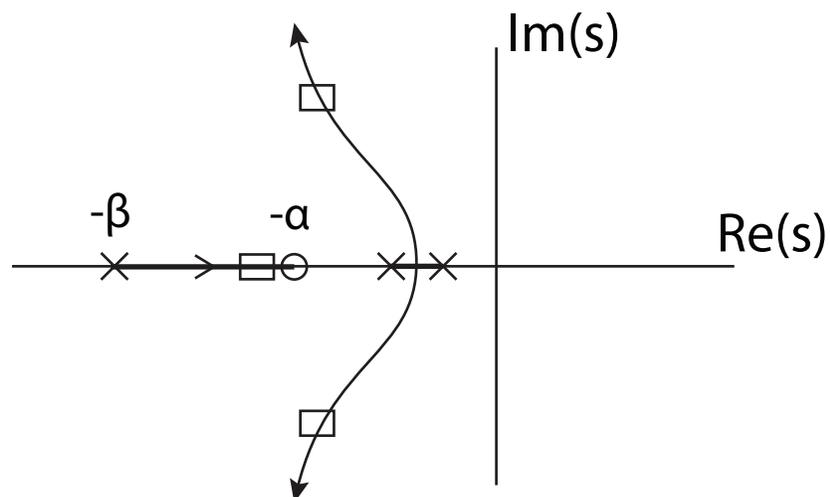


So gain only doesn't work - must add *lead compensation*:

$$K(s) = k \frac{(s + \alpha)}{(s + \beta)}$$

where  $\alpha < \beta$ .

Then rough locus will be

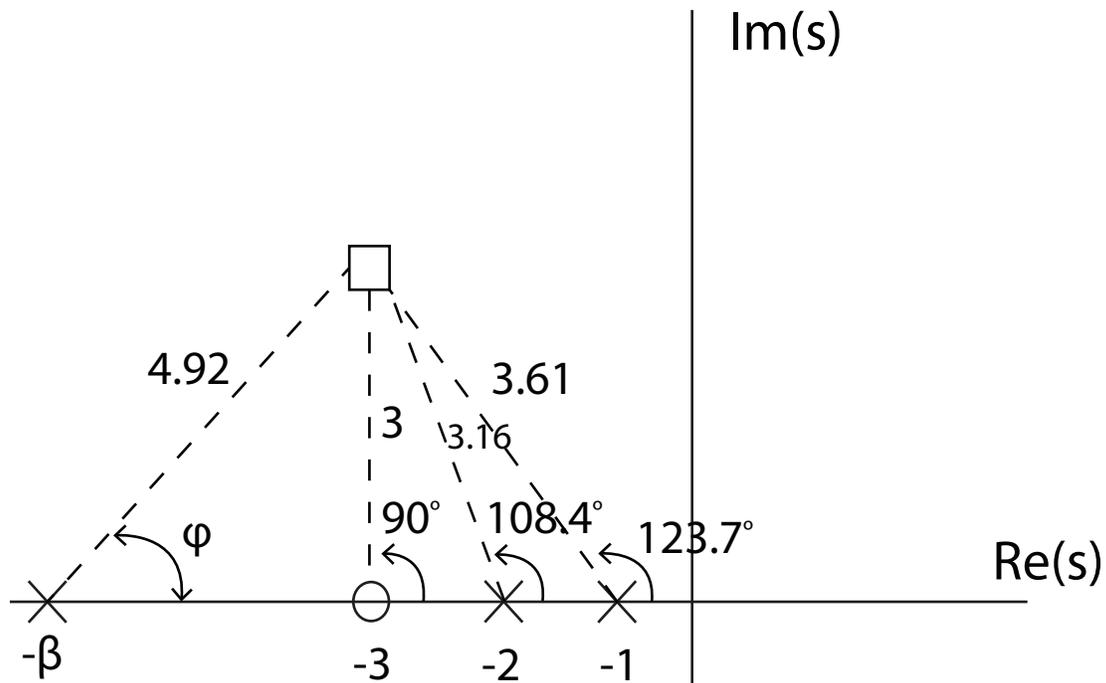


Must choose  $\alpha, \beta, k$  to make this work.

We have multiple degrees of freedom, so answer is not unique. Let's fix

$$\alpha = 3$$

to guarantee the real closed-loop pole settles faster than complex poles. Then  $\beta$  must be selected to achieve desired angle condition.



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